

FEMTOCLOCKS[™] CRYSTAL-TO-3.3V, 2.5V LVPECL CLOCK GENERATOR

ICS843022I-02

General Description

The ICS843022I-02 is a Gigabit Ethernet Clock Generator. The ICS843022I-02 uses a 25MHz crystal to synthesize 125MHz or 62.5MHz. The ICS843022I-02 has excellent phase jitter performance, over the 12kHz – 20MHz integration range. The ICS843022I-02 is packaged in a small 16-pin VFQFN, making it ideal for use in systems with limited board space.

FEATURES

- One differential 3.3V or 2.5V LVPECL output
- Crystal oscillator interface designed for 25MHz, 18pF parallel resonant crystal
- Output frequencies: 125MHz or 62.5MHz (selectable)
- RMS phase jitter @ 125MHz, using a 25MHz crystal (12kHz 20MHz): 0.57ps (typical)
- Full 3.3V or 2.5V operating supply
- -40°C to 85°C ambient operating temperature
- Available in lead-free (RoHS 6) packaging

FUNCTION TABLE

Inputs	Output Frequencies		
FREQ_SEL	(with a 25MHz crystal)		
0	125MHz		
1	62.5MHz		

BLOCK DIAGRAM



PIN ASSIGNMENT



TABLE 1. PIN DESCRIPTIONS

Number	Name	Ту	ре	Description
1	PWR_DN	Input	Pullup	Output state control pin. See Table 3. LVCMOS/LVTTL interface levels.
2, 3	XTAL_OUT, XTAL_IN	Input		Crystal oscillator interface. XTAL_in is the input, XTAL_OUT is the output.
4, 5, 8	V _{EE}	Power		Negative supply pins.
6, 13, 14, 15	nc	Unused		No connect.
7	FREQ_SEL	Input	Pulldown	Frequency select pin. LVCMOS/LVTTL interface levels.
9, 16	V _{cc}	Power		Power supply pins.
10	V _{cco}	Power		Output supply pin.
11, 12	nQ, Q	Output		Differential clock outputs. LVPECL interface levels.

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C _{IN}	Input Capacitance			4		pF
R _{PULLDOWN}	Input Pulldown Resistor			51		kΩ
R _{PULLUP}	Input Pullup Resistor			51		kΩ

TABLE 3. PWR_DN FUNCTION TABLE

PWR_DWN Input	Description
0	Output in high impedance
1	Output in normal operation

ABSOLUTE MAXIMUM RATINGS

Supply Voltage, V _{cc}	4.6V
Inputs, V _I	-0.5V to V_{cc} + 0.5V
Outputs, I _o Continuous Current Surge Current	50mA 100mA
Package Thermal Impedance, $\boldsymbol{\theta}_{_{JA}}$	74.9°C/W (0 mps)
Storage Temperature, T _{STG}	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Table 4A. Power Supply DC Characteristics, $V_{cc} = V_{cco} = 3.3V \pm 5\%$, $V_{ee} = 0V$, Ta = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{cc}	Power Supply Voltage		3.135	3.3	3.465	V
V _{cco}	Output Supply Voltage		3.135	3.3	3.465	V
		PWR_DN = 1			86	mA
I _{EE} Power Supply Current	PWR_DN = 0			<1	mA	

Table 4B. Power Supply DC Characteristics, $V_{\rm CC}$ = $V_{\rm CCO}$ = 2.5V±5%, $V_{\rm EE}$ = 0V, Ta = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{cc}	Power Supply Voltage		2.375	2.5	2.625	V
V _{cco}	Output Supply Voltage		2.375	2.5	2.625	V
		PWR_DN = 1			77	mA
I _{EE} Power Supply Current	$PWR_DN = 0$			<1	mA	

TABLE 4C. LVCMOS/LVTTL DC CHARACTERISTICS,	$V_{\rm CC} = V_{\rm CCO} = 3.3V \pm 5\%$ or 2.5V $\pm 5\%$, $V_{\rm EE} = 0$ V, Ta = -40°C t	го 85°С
--	--	----------------

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V	Input High Voltage		V _{cc} = 3.465V	2		V _{cc} + 0.3	V
V _{IH}	Input High Voltage		$V_{cc} = 2.625V$	1.7		V _{cc} + 0.3	V
V			$V_{cc} = 3.465V$	-0.3		0.8	V
V _{IL}	Input Low Voltage		$V_{cc} = 2.625V$	-0.3		0.7	V
	I Input High Current 🛏	FREQ_SEL	$V_{\rm CC} = V_{\rm IN} = 3.465 V \text{ or } 2.625 V$			150	μA
I _{IH}		PWR_DN	$V_{\rm CC} = V_{\rm IN} = 3.465 V \text{ or } 2.625 V$			5	μA
	Input Low Current	FREQ_SEL	$V_{\rm CC} = 3.465 V \text{ or } 2.625 V, V_{\rm IN} = 0 V$	-5			μA
۱ _{IL}		PWR_DN	$V^{}_{\rm CC}$ = 3.465V or 2.625V, $V^{}_{\rm IN}$ = 0V	-150			μA

3

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V _{OH}	Output High Voltage; NOTE 1		V _{cco} - 1.4		V _{cco} - 0.9	V
V _{OL}	Output Low Voltage; NOTE 1		V _{cco} - 2.0		V _{cco} - 1.7	V
V _{SWING}	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

TABLE 4D. LVPECL DC Characteristics, $V_{CC} = V_{CCO} = 3.3V \pm 5\%$ or $2.5V \pm 5\%$, $V_{EE} = 0V$, Ta = -40°C to 85° C

NOTE 1: Outputs terminated with 50 $\!\Omega$ to V $_{_{CCO}}$ - 2V.

TABLE 5. CRYSTAL CHARACTERISTICS

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		F	undamenta	I	
Frequency			25		MHz
Equivalent Series Resistance (ESR)				50	Ω
Shunt Capacitance				7	pF

NOTE: It is not recommended to overdrive the crystal input with an external clock.

TABLE 6A. AC CHARACTERISTICS, $V_{CC} = V_{CCO} = 3.3V \pm 5\%$, $V_{EE} = 0V$, TA = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
f		F_SEL = 0		125		MHz
IOUT	Output Frequency	F_SEL = 1		62.5		MHz
fiit((())	RMS Phase Jitter;	125MHz, Integration Range: 12kHz - 20MHz		0.57		ps
<i>t</i> jit(Ø)	NOTE 1	62.5MHz, Integration Range: 12kHz - 10MHz		0.52		ps
t _R / t _F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle		47		53	%

NOTE 1: Please refer to the Phase Noise Plot.

Table 6B. AC Characteristics, $V_{_{\rm CC}}$ = $V_{_{\rm CCO}}$ = 2.5V±5%, $V_{_{\rm EE}}$ = 0V, Ta = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
£	Output Frequency	F_SEL = 0		125		MHz
OUT		F_SEL = 1		62.5		MHz
<i>t</i> jit(Ø)	RMS Phase Jitter; NOTE 1	125MHz, Integration Range: 12kHz - 20MHz		0.62		ps
		62.5MHz, Integration Range: 12kHz - 10MHz		0.60		ps
t _R / t _F	Output Rise/Fall Time	20% to 80%	200		700	ps
odc	Output Duty Cycle		47		53	%

NOTE 1: Please refer to the Phase Noise Plot.









APPLICATION INFORMATION

CRYSTAL INPUT INTERFACE

The ICS843022I-02 has been characterized with 18pF parallel resonant crystals. The capacitor values, C1 and C2, shown in *Figure 1* below were determined using a 25MHz, 18pF parallel

resonant crystal and were chosen to minimize the ppm error. The optimum C1 and C2 values can be slightly adjusted for different board layouts.





 \mathcal{O}

VFQFN EPAD THERMAL RELEASE PATH

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 3*. The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes")

are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/ slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/ Electrically Enhance Leadfame Base Package, Amkor Technology.



FIGURE 3. P.C.ASSEMBLY FOR EXPOSED PAD THERMAL RELEASE PATH -SIDE VIEW (DRAWING NOT TO SCALE)

RECOMMENDATIONS FOR UNUSED INPUT PINS

INPUTS:

LVCMOS CONTROL PINS

All control pins have internal pull-ups or pull-downs; additional resistance is not required but can be added for additional protection. A $1k\Omega$ resistor can be used.

TERMINATION FOR 3.3V LVPECL OUTPUTS

The clock layout topology shown below is typical for IA64/32 platforms. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω



FIGURE 4A. LVPECL OUTPUT TERMINATION

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.



FIGURE 4B. LVPECL OUTPUT TERMINATION

9

TERMINATION FOR 2.5V LVPECL OUTPUTS

Figure 5A and *Figure 5B* show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to V_{cc} - 2V. For V_{cc} = 2.5V, the V_{cc} - 2V is very close to ground



FIGURE 5A. 2.5V LVPECL DRIVER TERMINATION EXAMPLE



FIGURE 5C. 2.5V LVPECL TERMINATION EXAMPLE

level. The R3 in Figure 5B can be eliminated and the termination is shown in *Figure 5C*.



FIGURE 5B. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the ICS843022I-02. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the ICS843022I-02 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{cc} = 3.3V + 5\% = 3.465V$, which gives worst case results. **NOTE:** Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = V_{CC MAX} * I_{EE MAX} = 3.465V * 86mA = 298mW
- Power (outputs)_{MAX} = 30mW/Loaded Output pair

Total Power (3.465V, with all outputs switching) = 298mW + 30mW = 328mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the boost wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature 125°C.

The equation for Tj is as follows: $Tj = \theta_{JA} * Pd_{total} + T_{A}$

Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 74.9°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is: 85°C + 0.328W * 74.9°C/W = 109.6°C. This is well below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

TABLE 7. THERMAL RESISTANCE θ_{JA} FOR 16-PIN VFQFN, FORCED CONVECTION

θ_{JA} by Velocity (Meters per Second)					
	0	1	2.5		
Multi-Layer PCB, JEDEC Standard Test Boards	74.9°C/W	65.5°C/W	58.8°C/W		

3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load. LVPECL output driver circuit and termination are shown in *Figure 6.*



FIGURE 6. LVPECL DRIVER CIRCUIT AND TERMINATION

To calculate worst case power dissipation into the load, use the following equations which assume a 50 Ω load, and a termination voltage of V $_{cc}$ - 2V.

• For logic high, $V_{OUT} = V_{OH MAX} = V_{CC MAX} - 0.9V$

$$(V_{CC_MAX} - V_{OH_MAX}) = 0.9V$$

• For logic low, $V_{OUT} = V_{OL_MAX} = V_{CC_MAX} - 1.7V$

$$(V_{CC_MAX} - V_{OL_MAX}) = 1.7V$$

Pd_H is power dissipation when the output drives high. Pd_L is the power dissipation when the output drives low.

 $\begin{aligned} \mathsf{Pd}_{\mathsf{H}} &= [(\mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - 2\mathsf{V}))/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}))/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}))/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}))/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}})]/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}})]/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OH}_{\mathsf{MAX}}}}) = [(\mathsf{V} - \mathsf{V}_{\mathsf{CC}_{\mathsf{MAX}}} - \mathsf{V}_{\mathsf{OH}_{\mathsf{MAX}}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{C}}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{C}}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{\mathsf{CC}_{\mathsf{MAX}}} - \mathsf{V}_{\mathsf{C}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{\mathsf{CC}_{\mathsf{MAX}}} - \mathsf{V}_{_{\mathsf{C}}})] * (\mathsf{V}_{\mathsf{CC}_{\mathsf{MAX}}} - \mathsf{V}_{\mathsf{C}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{\mathsf{CC}_{\mathsf{C}}}) + \mathsf{V}_{\mathsf{C}}) * (\mathsf{V}_{\mathsf{C}})]/\mathsf{R}_{_{\mathsf{C}}}] * (\mathsf{V}_{\mathsf{C}}) * (\mathsf{V}_{\mathsf{C}}) + \mathsf{V}_{\mathsf{C}}) * (\mathsf{V}_{\mathsf{C}}) * (\mathsf{V}_{\mathsf{C}}) * (\mathsf{V}_{\mathsf{C}}) + \mathsf{V}_{\mathsf{C}}) * (\mathsf{V}_{\mathsf{C}}) * (\mathsf{V}) * (\mathsf{V}_{\mathsf{C}}) * (\mathsf{$

 $\begin{aligned} \mathsf{Pd}_{\mathsf{L}} &= [(\mathsf{V}_{_{\mathsf{OL}_{\mathsf{MAX}}}} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - 2\mathsf{V}))/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OL}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OL}_{\mathsf{MAX}}}}))/\mathsf{R}_{_{\mathsf{L}}}] * (\mathsf{V}_{_{\mathsf{CC}_{\mathsf{MAX}}}} - \mathsf{V}_{_{\mathsf{OL}_{\mathsf{MAX}}}}) = [(2\mathsf{V} - 1.7\mathsf{V})/50\Omega] * 1.7\mathsf{V} = \mathbf{10.2mW} \end{aligned}$

Total Power Dissipation per output pair = Pd_H + Pd_L = **30mW**

RELIABILITY INFORMATION

TABLE 8. $\boldsymbol{\theta}_{_{JA}} \text{vs.}$ Air Flow Table for 16 Lead VFQFN

θ_{JA} at 0 Air Flow (Meters per Second)				
	0	1	2.5	
Multi-Layer PCB, JEDEC Standard Test Boards	74.9°C/W	65.5°C/W	58.8°C/W	

TRANSISTOR COUNT

The transistor count for ICS843022I-02 is: 1719

PACKAGE OUTLINE - K SUFFIX FOR 16 LEAD VFQFN



JEDEC VARIATION ALL DIMENSIONS IN MILLIMETERS				
SYMBOL	MINIMUM	MAXIMUM		
Ν	1	6		
Α	0.80	1.0		
A1	0	0.05		
A3	0.25 Reference			
b	0.18	0.30		
е	0.50 BASIC			
N _D	4			
N _E	4			
D	3.0			
D2	1.0	1.80		
E	3.0			
E2	1.0	1.80		
L	0.30	0.50		

TABLE 9. PACKAGE DIMENSIONS

Reference Document: JEDEC Publication 95, MO-220

TABLE 10. ORDERING INFORMATION

Part/Order Number	Marking	Package	Shipping Packaging	Temperature	
843022AKI-02LF	312L	16 Lead "Lead-Free" VFQFN	Tube	-40°C to 85°C	
843022AKI-02LFT	312L	16 Lead "Lead-Free" VFQFN	Tape & Reel	-40°C to 85°C	

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

	REVISION HISTORY SHEET					
Rev	Rev Table Page Description of Change					
А	T10	8 15	Updated <i>Thermal Release Path</i> section. Ordering Information Table - corrected marking from 312L to 3I2L.	2/20/08		
А	Т5	1 4 7	Deleted HiPerClockS references. Crystal Characteristics Table - added note. Deleted application note, LVCMOS to XTAL Interface.	10/22/12		
	T7	12	Deleted quantity from tape and reel.			

Innovate with IDT and accelerate your future networks. Contact:

www.IDT.com

For Sales

800-345-7015 408-284-8200 Fax: 408-284-2775 For Tech Support netcom@idt.com 480-763-2056

Corporate Headquarters

Integrated Device Technology, Inc. 6024 Silver Creek Valley Road San Jose, CA 95138 United States 800 345 7015 +408 284 8200 (outside U.S.)

NIPPON IDT KK Sanbancho Tokyu Bld. 7F, 8-1 Sanbancho

Japan

8-1 Sanbancho Chiyoda-ku, Tokyo 102-0075 +81 3 3221 9822 +81 3 3221 9824 (fax)

Asia

Integrated Device Technology IDT (S) Pte. Ltd. 1 Kallang Sector, #07-01/06 Kolam Ayer Industrial Park Singapore 349276 +65 6 744 3356 +65 6 744 1764 (fax)

Europe

IDT Europe, Limited 321 Kingston Road Leatherhead, Surrey KT22 7TU England +44 (0) 1372 363 339 +44 (0) 1372 378851 (fax)



© 2008 Integrated Device Technology, Inc. All rights reserved. Product specifications subject to change without notice. IDT, the IDT logo, ICS and HiPerClockS are trademarks of Integrated Device Technology, Inc. Accelerated Thinking is a service mark of Integrated Device Technology, Inc. All other brands, product names and marks are or may be trademarks or registered trademarks used to identify products or services of their respective owners. Printed in USA