

## Low Skew, 1-to-5, Differential-to-HSTL **Fanout Buffer**

ICS85214I

DATA SHEET

## GENERAL DESCRIPTION

The ICS85214I is a low skew, high performance 1-to-5 Differential-to-HSTL Fanout Buffer. The CLK0, nCLK0 pair can accept most standard differential input levels. The single ended CLK1 input accepts LVCMOS or LVTTL input levels. Guaranteed output and part to part skew characteristics make the ICS85214I ideal for those clock distribution applications demanding well defined performance and repeatability.

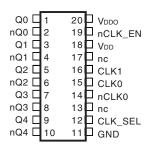
## **F**EATURES

- Five differential HSTL compatible outputs
- Selectable differential CLK0, nCLK0 or LVCMOS/LVTTL clock inputs
- · CLK0, nCLK0 pair can accept the following differential input levels: LVDS, LVPECL, HSTL, HCSL
- CLK1 can accept the following input levels: LVCMOS or LVTTL
- Output frequency up to 700MHz
- Translates any single ended input signal to HSTL levels with resistor bias on nCLK0 input
- Output skew: 40ps (maximum)
- Part-to-part skew: 300ps (maximum)
- Propagation delay: 1.8ns (maximum)
- 3.3V core, 1.8V output operating supply
- Available in Lead-Free (RoHS 6) package
- -40°C to 85°C ambient operating temperature

## **BLOCK DIAGRAM**

## nCLK\_EN -Q LE CLK0 Q0 nQ0 CLK1 Q1 nQ1 CLK\_SEL Ω2 nQ2 Q3 nQ3 Ω4

## PIN ASSIGNMENT



### ICS85214I 20-Lead TSSOP 6.5mm x 4.4mm x 0.92mm package body **G** Package Top View

TABLE 1. PIN DESCRIPTIONS

Number	Name	Ту	ре	Description
1, 2	Q0, nQ0	Output		Differential output pair. HSTL interface levels.
3, 4	Q1, nQ1	Output		Differential output pair. HSTL interface levels.
5, 6	Q2, nQ2	Output		Differential output pair. HSTL interface levels.
7, 8	Q3, nQ3	Output		Differential output pair. HSTL interface levels.
9, 10	Q4, nQ4	Output		Differential output pair. HSTL interface levels.
11	GND	Power		Power supply ground.
12	CLK_SEL	Input	Pulldown	Clock select input. When HIGH, selects CLK1 input. When LOW, selects CLK0, nCLK0 input. LVTTL / LVCMOS interface levels.
13, 17	nc	Unused		No connect.
14	nCLK0	Input	Pullup	Inverting differential clock input.
15	CLK0	Input	Pulldown	Non-inverting differential clock input.
16	CLK1	Input	Pulldown	Clock input. LVTTL / LVCMOS interface levels.
18	$V_{_{\mathrm{DD}}}$	Power		Power supply pin.
19	nCLK_EN	Input	Pulldown	Synchronizing clock enable. When LOW, clock outputs follow clock input. When HIGH, Q outputs are forced low, nQ outputs are forced high. LVTTL / LVCMOS interface levels.
20	$V_{\scriptscriptstyle DDO}$	Power		Output supply pin.

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance			4		pF
R <sub>PULLUP</sub>	Input Pullup Resistor			51		kΩ
R <sub>PULLDOWN</sub>	Input Pulldown Resistor			51		kΩ

TABLE 3A. CONTROL INPUT FUNCTION TABLE

Inputs	Out	puts
nCLK_EN	Q0:Q4	nQ0:nQ4
0	Enabled	Enabled
1	Disabled; LOW	Disabled; HIGH

After nCLK\_EN switches, the clock outputs are disabled or enabled following a rising and falling input clock edge as shown in Figure 1.

In the active mode, the state of the outputs are a function of the CLK0, nCLK0 inputs as described in Table 3B.

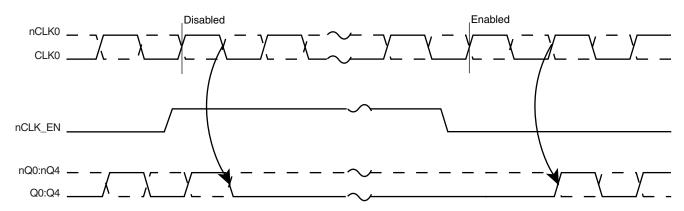


FIGURE 1. nCLK\_ENTIMING DIAGRAM

TABLE 3B. CLOCK INPUT FUNCTION TABLE

	Input	s		Ou	itputs	Input to Output Mode	Polarity
CLK_SEL	CLK0	nCLK0	CLK1	Q0:Q4	Q0:Q4 nQ0:nQ4		Polarity
0	0	1	Х	LOW	HIGH	Differential to Differential	Non Inverting
0	1	0	Х	HIGH	LOW	Differential to Differential	Non Inverting
0	0	Biased; NOTE 1	Х	LOW	HIGH	Single Ended to Differential	Non Inverting
0	1	Biased; NOTE 1	Х	HIGH	LOW	Single Ended to Differential	Non Inverting
0	Biased; NOTE 1	0	X	HIGH	LOW	Single Ended to Differential	Inverting
0	Biased; NOTE 1	1	Х	LOW	HIGH	Single Ended to Differential	Inverting
1	X	X	0	LOW	HIGH	Single Ended to Differential	Non Inverting
1	Х	Х	1	HIGH	LOW	Single Ended to Differential	Non Inverting

NOTE 1: Please refer to the Application Information section, "Wiring the Differential Input to Accept Single Ended Levels".

#### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage, V<sub>DD</sub> 4.6V

 $\begin{array}{ll} \text{Inputs, V}_{\text{\tiny I}} & \text{-0.5V to V}_{\text{\tiny DD}} + \text{0.5V} \\ \text{Outputs, I}_{\text{\tiny O}} & \text{-0.5V to V}_{\text{\tiny DD}} + \text{0.5V} \\ \text{Package Thermal Impedance, } \theta_{\text{\tiny JA}} & \text{73.2°C/W (0 lfpm)} \\ \text{Storage Temperature, T}_{\text{\tiny STG}} & \text{-65°C to 150°C} \end{array}$ 

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Table 4A. Power Supply DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $V_{DDO} = 1.8V \pm 0.2V$ ,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>DD</sub>	Power Supply Voltage		3.135	3.3	3.465	V
V <sub>DDO</sub>	Output Power Supply Voltage		1.6	1.8	2.0	V
I <sub>DD</sub>	Power Supply Current				80	mA

Table 4B. LVCMOS / LVTTL DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $V_{DDO} = 1.8V \pm 0.2V$ ,  $T_A = -40^{\circ}C$  to  $85^{\circ}C$ 

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V	Input High Voltage	nCLK_EN, CLK_SEL		2		$V_{DD} + 0.3$	V
V <sub>IH</sub>	Input riigh voltage	CLK1		2		$V_{DD} + 0.3$	V
\/	V	nCLK_EN, CLK_SEL		-0.3		0.8	V
V <sub>IL</sub>	Input Low Voltage	CLK1		-0.3		1.3	V
I <sub>IH</sub>	Input High Current	CLK1, CLK_SEL, nCLK_EN	$V_{DD} = V_{IN} = 3.465V$			150	μΑ
I <sub>IL</sub>	Input Low Current	CLK1, CLK_SEL, nCLK_EN	$V_{DD} = 3.465V, V_{IN} = 0V$	-5			μΑ

Table 4C. Differential DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $V_{DDO} = 1.8V \pm 0.2V$ , Ta = -40°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
	Input High Current	nCLK0	$V_{DD} = V_{IN} = 3.465V$			5	μΑ
I'IH	Imput High Current	CLK0	$V_{DD} = V_{IN} = 3.465V$			150	μΑ
		nCLK0	$V_{DD} = 3.465V, V_{IN} = 0V$	-150			μΑ
I <sub>IL</sub>	Input Low Current	CLK0	$V_{DD} = 3.465V, V_{IN} = 0V$	-5			μΑ
V <sub>PP</sub>	Peak-to-Peak Input Voltage			0.15		1.3	V
V <sub>CMR</sub>	Common Mode Input Voltage; NOTE 1, 2			0.5		V <sub>DD</sub> - 0.85	V

NOTE 1: For single ended applications the maximum input voltage for CLK0, nCLK0 is V<sub>nn</sub> + 0.3V.

NOTE 2: Common mode voltage is defined as  $V_{\rm HI}$ .

Table 4D. HSTL DC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $V_{DDO} = 1.8V \pm 0.2V$ , Ta = -40°C to 85°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OH</sub>	Output High Voltage; NOTE 1		1		1.4	V
V <sub>OL</sub>	Output Low Voltage; NOTE 1		0		0.4	V
V <sub>ox</sub>	Output Crossover Voltage		$38\% \times (V_{OH} - V_{OL}) + V_{OL}$		60% x (V <sub>OH</sub> - V <sub>OL</sub> ) + V <sub>OL</sub>	V
V <sub>SWING</sub>	Peak-to-Peak Output Voltage Swing		0.6		1.1	V

NOTE 1: Outputs terminated with  $50\Omega$  to ground.

Table 5. AC Characteristics,  $V_{DD} = 3.3V \pm 5\%$ ,  $V_{DDO} = 1.8V \pm 0.2V$ , Ta = -40°C to 85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
f	Output Fraguanay	CLK0, nCLK0				700	MHz
OUT	Output Frequency	CLK1				300	MHZ
t <sub>PD</sub>	Propagation Delay;	NOTE 1	<i>f</i> ≤ 700MHz	1.0		1.8	ns
tsk(o)	Output Skew; NOTE	E 2, 4				40	ps
tsk(pp)	Part-to-Part Skew; I	NOTE 3, 4				300	ps
t <sub>R</sub> / t <sub>F</sub>	Output Rise/Fall Tir	ne	20% to 80%	200		800	ps
ada	Output Duty Cycle	CLK0, nCLK0		46		54	%
odc	Output Duty Cycle CLK1		<i>f</i> ≤ 266MHz	44		56	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE: All parameters measured at  $f_{\text{OUT}}$  unless noted otherwise.

NOTE: The cycle to cycle jitter on the input will equal the jitter on the output. The part does not add jitter.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

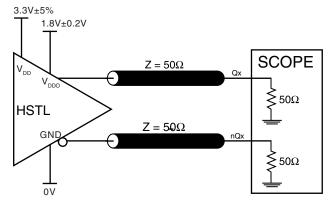
NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions.

Measured at output differential cross points.

NOTE 3: Defined as skew between outputs on different devices operating at the same supply voltages and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

NOTE 4: This parameter is defined in accordance with JEDEC Standard 65.

## PARAMETER MEASUREMENT INFORMATION



nCLK0

-V<sub>PP</sub>

-CLK0

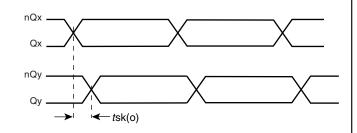
-V<sub>CMR</sub>

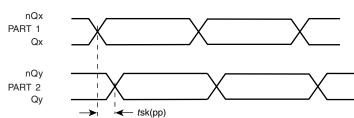
-CLK0

3.3V/1.8V OUTPUT LOAD ACTEST CIRCUIT

DIFFERENTIAL INPUT LEVEL

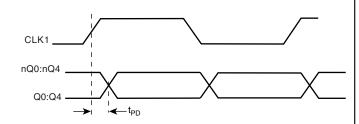
**GND** 

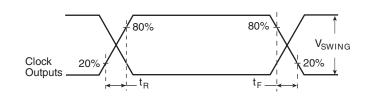


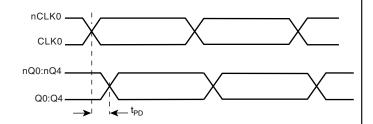


OUTPUT SKEW

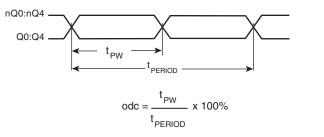
PART-TO-PART SKEW







OUTPUT RISE/FALLTIME



PROPAGATION DELAY

## **APPLICATIONS INFORMATION**

#### WIRING THE DIFFERENTIAL INPUT TO ACCEPT SINGLE ENDED LEVELS

Figure 2 shows how a differential input can be wired to accept single ended levels. The reference voltage  $V_{\text{REF}} = V_{\text{DD}}/2$  is generated by the bias resistors R1 and R2. The bypass capacitor (C1) is used to help filter noise on the DC bias. This bias circuit should be located as close to the input pin as possible. The ratio of R1 and R2 might need to be adjusted to position the  $V_{\text{REF}}$  in the center of the input voltage swing. For example, if the input clock swing is 2.5V and  $V_{\text{DD}} = 3.3V$ , R1 and R2 value should be adjusted to set  $V_{\text{REF}}$  at 1.25V. The values below are for when both the single-ended swing and  $V_{\text{DD}}$  are at the same voltage. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the input will attenuate the signal in half. This can be done in one of two ways. First, R3 and R4 in parallel should equal the transmission line

impedance. For most 50 applications, R3 and R4 can be  $100\Omega$ . The values of the resistors can be increased to reduce the loading for slower and weaker LVCMOS driver. When using single ended signaling, the noise rejection benefits of differential signaling are reduced. Even though the differential input can handle full rail LVCMOS signaling, it is recommended that the amplitude be reduced. The datasheet specifies a lower differential amplitude, however this only applies to differential signals. For single-ended applications, the swing can be larger, however  $V_{\rm IL}$  cannot be less than -0.3V and  $V_{\rm IH}$  cannot be more than  $V_{\rm DD}$  + 0.3V. Though some of the recommended components might not be used, the pads should be placed in the layout. They can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a differential signal.

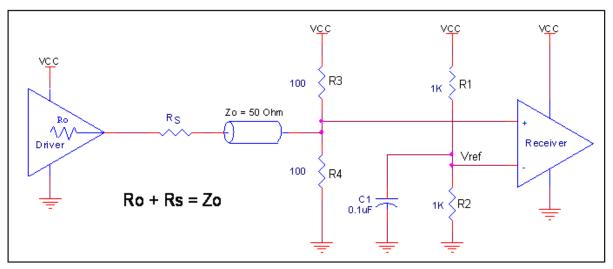


FIGURE 2. SINGLE ENDED SIGNAL DRIVING DIFFERENTIAL INPUT

#### DIFFERENTIAL CLOCK INPUT INTERFACE

The CLK /nCLK accepts LVDS, LVPECL, HSTL, HCSL and other differential signals. Both differential signals must meet the  $V_{PP}$  and  $V_{CMR}$  input requirements. Figures 3A to 3E show interface examples for the CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are

FIGURE 3A. CLK/nCLK INPUT DRIVEN BY AN IDT OPEN EMITTER LVHSTL DRIVER

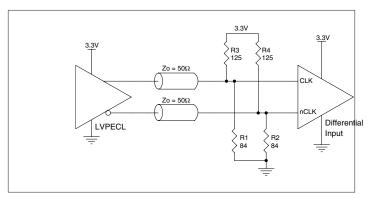


FIGURE 3C. CLK/nCLK INPUT DRIVEN BY A 3.3V LVPECL DRIVER

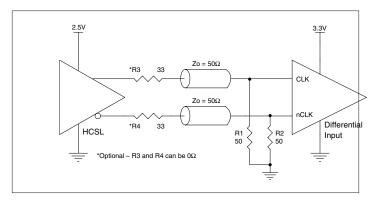


FIGURE 3E. CLK/nCLK INPUT DRIVEN BY A 3.3V HCSL DRIVER

examples only. Please consult with the vendor of the driver component to confirm the driver termination requirements. For example in Figure 3A, the input termination applies for IDT open emitter LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination

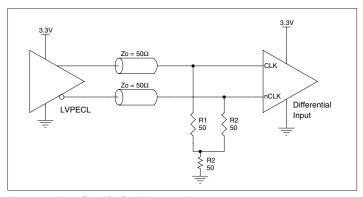


FIGURE 3B. CLK/nCLK INPUT DRIVEN BY A 3.3V LVPECL DRIVER

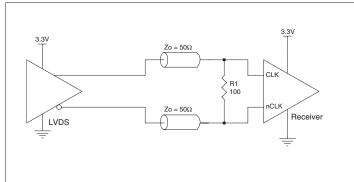


FIGURE 3D. CLK/nCLK INPUT DRIVEN BY A 3.3V LVDS DRIVER

## SCHEMATIC EXAMPLE

Figure 4 shows a schematic example of the ICS85214I. In this example, the input is driven by an HSTL driver. The decoupling capacitors should be physically located

near the power pin. For ICS85214I, the unused outputs can be left floating.

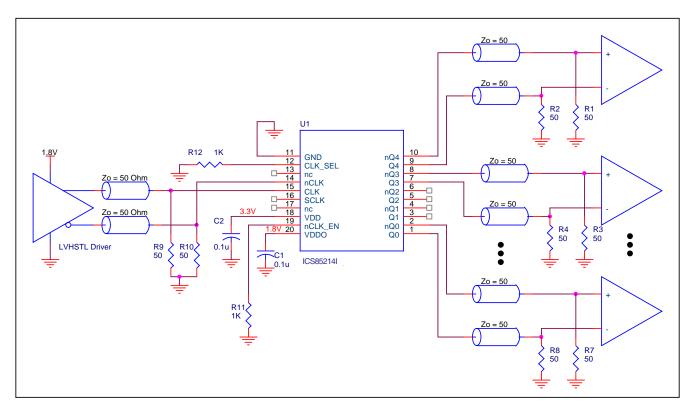


FIGURE 4. ICS85214I HSTL BUFFER SCHEMATIC EXAMPLE

## Power Considerations

This section provides information on power dissipation and junction temperature for the ICS85214I. Equations and example calculations are also provided.

#### 1. Power Dissipation.

The total power dissipation for the ICS85214I is the sum of the core power plus the power dissipated in the load(s).

The following is the power dissipation for  $V_{DD} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> = V<sub>DD MAX</sub> \* I<sub>DD MAX</sub> = 3.465V \* 80mA = 227.2mW
- Power (outputs)<sub>MAX</sub> = 32.8mW/Loaded Output pair
   If all outputs are loaded, the total power is 5 \* 32.8mW = 164mW

Total Power MAX (3.465V, with all outputs switching) = 227.2mW + 164mW = 391.2mW

#### 2. Junction Temperature.

Junction temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows:  $Tj = \theta_{JA} * Pd\_total + T_A$ 

Tj = Junction Temperature

 $\theta_{14}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

 $T_{\Delta}$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 66.6°C/W per Table 6 below. Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}\text{C} + 0.391\text{W} * 66.6^{\circ}\text{C/W} = 111^{\circ}\text{C}$ . This is below the limit of  $125^{\circ}\text{C}$ .

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow,

Table 6. Thermal Resistance  $\theta_{JA}$  For 20-pin TSSOP, Forced Convection

# θ<sub>JA</sub> by Velocity (Linear Feet per Minute) 0 200 500 Single-Layer PCB, JEDEC Standard Test Boards 114.5°C/W 98.0°C/W 88.0°C/W Multi-Layer PCB, JEDEC Standard Test Boards 73.2°C/W 66.6°C/W 63.5°C/W NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

#### 3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

HSTL output driver circuit and termination are shown in Figure 5.

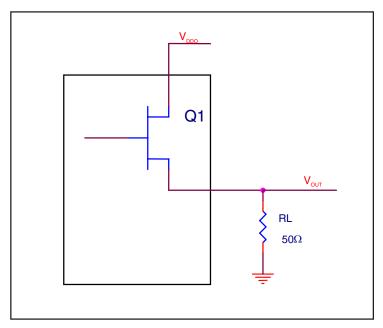


FIGURE 5. HSTL DRIVER CIRCUIT AND TERMINATION

To calculate worst case power dissipation into the load, use the following equations which assume a  $50\Omega$  load.

Pd\_H is power dissipation when the output drives high.

Pd\_L is the power dissipation when the output drives low.

$$\begin{split} & Pd\_H = (V_{OH\_MIN}/R_L) * (V_{DDO\_MAX} - V_{OH\_MIN}) \\ & Pd\_L = (V_{OL\_MAX}/R_L) * (V_{DDO\_MAX} - V_{OL\_MAX}) \end{split}$$

$$Pd_H = (1.0V/50\Omega) * (2V - 1.0V) = 20mW$$
  
 $Pd_L = (0.4V/50\Omega) * (2V - 0.4V) = 12.8mW$ 

Total Power Dissipation per output pair = Pd\_H + Pd\_L = 32.8mW

## RELIABILITY INFORMATION

Table 7.  $\theta_{\text{JA}} \text{vs. Air Flow Table for 20 Lead TSSOP}$ 

## $\theta_{AA}$ by Velocity (Linear Feet per Minute)

0200500Single-Layer PCB, JEDEC Standard Test Boards114.5°C/W98.0°C/W88.0°C/WMulti-Layer PCB, JEDEC Standard Test Boards73.2°C/W66.6°C/W63.5°C/W

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

#### TRANSISTOR COUNT

The transistor count for ICS85214I is: 674

## PACKAGE OUTLINE - G SUFFIX FOR 20 LEAD TSSOP

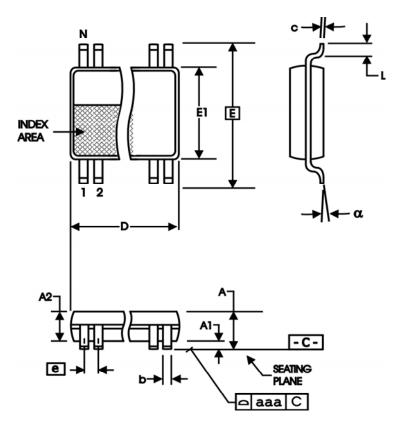


TABLE 8. PACKAGE DIMENSIONS

SYMBOL	Millin	neters
STWIDOL	Minimum	Maximum
N	2	0
А		1.20
A1	0.05	0.15
A2	0.80	1.05
b	0.19	0.30
С	0.09	0.20
D	6.40	6.60
E	6.40 E	BASIC
E1	4.30	4.50
е	0.65 E	BASIC
L	0.45	0.75
α	0°	8°
aaa		0.10

Reference Document: JEDEC Publication 95, MO-153

**TABLE 9. ORDERING INFORMATION** 

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
85214AGI	ICS85214AGI	20 lead TSSOP	tube	-40°C to 85°C
85214AGIT	ICS85214AGI	20 Lead TSSOP	2500 tape & reel	-40°C to 85°C
85214AGILF	ICS85214AGIL	20 Lead "Lead-Free" TSSOP	tube	-40°C to 85°C
85214AGILFT	ICS85214AGIL	20 Lead "Lead-Free" TSSOP	2500 tape & reel	-40°C to 85°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

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REVISION HISTORY SHEET								
Rev	Table	Page	Description of Change	Date				
Α	Т9	14	Added Lead-Free marking in Ordering Information table.	6/1/05				
			Updated datasheet's header/footer with IDT from ICS.					
В	T9	15	Removed ICS prefix from Part/Order Number column.	7/25/10				
		16	Added Contact Page.					
		4 - 5	AC/DC Characteristics Tables - corrected temperature in table headings from 0°C to -40°C.					
	T5	5	AC Characteristics Table - Added thermal note.					
В		7	Updated Wiring the differential Input to Accept Single Ended Levels application note.	5/6/11				
	T9	14	Ordering Information Table - Part/Order Number column, added suffix "T" in the order					
			number for table and reel.					
			Updated header and footer of the document.					

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