# Description

The device is intended to take 1 or 2 reference clocks, select between them, using a pin or register selection and generate up to 8 outputs that may be the same as the reference frequency or integer-divider versions of it.

The 8P79818 supports two output banks, each with its own divider and power supply. All outputs in one bank would generate the same output frequency, but each output can be individually controlled for output type, output enable or even powered-off.

The device supports a serial port for configuration of the parameters while in operation. The serial port can be selected to use the  $l^2C$  or SPI protocol. After power-up, all outputs will come up in LVDS mode and may be programmed to other configurations over the serial port. Outputs may be enabled or disabled under control of the OE input pin.

The device can operate over the -40°C to +85°C temperature range.

### Features

- Two differential inputs support LVPECL, LVDS, HCSL or LVCMOS reference clocks
  - Accepts input frequencies ranging from 1PPS (1Hz) to 700MHz
- Select which of the two input clocks is to be used as the reference clock for which divider via pin or register selection
  - Switchover will not generate any runt clock pulses on the output
- Generates eight differential outputs or eight LVCMOS outputs, Bank A only
  - Differential outputs selectable as LVPECL, LVDS, CML or HCSL
  - Differential outputs support frequencies from 1PPS to 700MHz
  - LVCMOS outputs support frequencies from 1PPS to 200MHz
  - LVCMOS outputs in the same pair may be inverted or in-phase relative to one another
- Outputs arranged in 2 banks of 4 outputs each
  - Each bank supports a separate power supply of 3.3V, 2.5V or 1.8V
  - 1.5V output voltage is also supported for LVCMOS, Bank A only
  - One divider per output bank, supporting divide ratios of 2...511 or divider bypass
- Output enable control pin
  - Output enable or disable will not cause any runt pulses
- Register programmable via I<sup>2</sup>C / SPI serial port
  - Individual output enables, output type selection and output power-down control bits supported
  - Input mux selection control bit
- Core voltage supply of 3.3V, 2.5V or 1.8V
- 40°C to +85°C ambient operating temperature
- Lead-free (RoHS 6) packaging

## **Block Diagram**

Figure 1: Block Diagram



8P79818 transistor count: 33,394

### **Pin Assignment**

Figure 2: Pin Assignments for 5mm x 5mm 32-Lead VFQFN Package (Top View)



## **Pin Description and Characteristic Tables**

#### **Table 1: Pin Description**

Number	Name	Type <sup>[a]</sup>	Description
1	SDO	Output	SPI mode data output signal. Unused in I <sup>2</sup> C mode.
2	QA0	Output	Positive differential clock output. Included in Bank A. Refer to Output Drivers section for details.
3	nQA0	Output	Negative differential clock output. Included in Bank A. Refer to Output Drivers section for details.
4	QA1	Output	Positive differential clock output. Included in Bank A. Refer to Output Drivers section for details.
5	nQA1	Output	Negative differential clock output. Included in Bank A. Refer to Output Drivers section for details.
6	V <sub>CCOA</sub>	Power	Output supply for output Bank A.
7	QA2	Output	Positive differential clock output. Included in Bank A. Refer to Output Drivers section for details.
8	nQA2	Output	Negative differential clock output. Included in Bank A. Refer to Output Drivers section for details.
9	QA3	Output	Positive differential clock output. Included in Bank A. Refer to Output Drivers section for details.
10	nQA3	Output	Negative differential clock output. Included in Bank A. Refer to Output Drivers section for details.
11	nl2C/SPI	Input (PD)	Select protocol for serial port: 0 = I <sup>2</sup> C mode 1 = SPI mode
12	V <sub>CC</sub>	Power	Core logic supply.
13	SA0/nCS	Input (PU)	SPI chip select input (active low) in SPI mode. Base address bit 0 in I <sup>2</sup> C mode.

#### Table 1: Pin Description (Cont.)

			Master output enable control
14	OE	Input (PU)	0 = All outputs high-impedance
			1 = All outputs enabled or disabled under control of register bits
15	nQB3	Output	Negative differential clock output. Included in Bank B. Refer to Output Drivers section for details.
16	QB3	Output	Positive differential clock output. Included in Bank B. Refer to Output Drivers section for details.
17	nQB2	Output	Negative differential clock output. Included in Bank B. Refer to Output Drivers section for details.
18	QB2	Output	Positive differential clock output. Included in Bank B. Refer to Output Drivers section for details.
19	V <sub>CCOB</sub>	Power	Output supply for output Bank B.
20	nQB1	Output	Negative differential clock output. Included in Bank B. Refer to Output Drivers section for details.
21	QB1	Output	Positive differential clock output. Included in Bank B. Refer to Output Drivers section for details.
22	nQB0	Output	Negative differential clock output. Included in Bank B. Refer to Output Drivers section for details.
23	QB0	Output	Positive differential clock output. Included in Bank B. Refer to Output Drivers section for details.
24	CLK_SEL	Input (PU)	Input clock selection control pin. This pin may be disabled by register control, but if enabled (default) its function is: 0 = CLK0 is selected 1 = CLK1 is selected
25	CLK1	Input (PD)	Non-inverting differential clock input.
26	nCLK1	Input (PU/ PD)	Inverting differential clock input. $V_{CC}/2$ when left floating (set by the internal pull-up and pull-down resistors).
27	V <sub>CC</sub>	Power	Core logic supply.
28	SCLK	Input (PU)	Serial port input clock for either SPI or I <sup>2</sup> C mode.
29	SDATA/ SDI	Input/Output (PU) Input (PU)	In I <sup>2</sup> C mode, this is the bi-directional data signal for the serial port In SPI mode, this is the data input signal.
30	V <sub>CC</sub>	Power	Core logic supply.
31	nCLK0	Input (PU/ PD)	Inverting differential clock input. $V_{CC}/2$ when left floating (set by the internal pull-up and pull-down resistors).
32	CLK0	Input (PD)	Non-inverting differential clock input.
EP	V <sub>EE</sub>	Ground	Must be connected to ground (GND).
L	1	1	

a. Pull-up (PU) and pull-down (PD) resistors are indicated in parentheses. *Pullup* and *Pulldown* refer to internal input resistors. See Table 10, *DC Input/ Output Characteristics*, for typical values.

## Principles of Operation

#### **Input Selection**

The 8P79818 supports two input references: CLK0 and CLK1 that may be driven with differential or single-ended clock signals. Either or both may be used as the source frequency for either output divider under control of the CLK\_SEL input pin or under register control.

The CLK\_SEL pin is the default selection mechanism and selects whether both dividers are driven by the CLK0, nCLK0 input (CLK\_SEL = Low) or by the CLK1, nCLK1 input (CLK\_SEL = High).

If the user enables register control via the SEL\_REG control bit, then there are 4 selection options available as shown in Table 2.

CLK_S	EL [1:0]	Description			
0	0	Divider A & B both driven from CLK0			
0	1	Divider A driven from CLK1 & Divider B driven from CLK0			
1	0	Divider A driven from CLK0 & Divider B driven from CLK1			
1	1	Divider A & B both driven from CLK1			

Table 2: Input Selection Register Control (SEL\_REG = 1)

#### **Output Dividers**

Each bank of outputs has its own divider. All outputs in the same bank will be driven by that divider and so will all have the same frequency. Divider A supplies the QA output bank and Divider B supplies the QB output bank. Each divider is capable of being driven by the same or a different input frequency. Each divider can pass that input frequency directly to the outputs or to divide it by any integer from 2 up to 511.

### **Output Drivers**

The QA[0:3] and QB[0:3] clock outputs are provided with register-controlled output drivers. By selecting the output drive type in the appropriate register, any of these outputs can support LVCMOS, LVPECL, CML, HCSL or LVDS logic levels.

CML operation supports both a 400mV peak-peak swing and an 800mV peak-peak swing selection.

The operating voltage ranges of each output bank is determined by its independent output power pin ( $V_{CCOA}$  or  $V_{CCOB}$ ). Output voltage levels of 1.8V, 2.5V or 3.3V are supported for differential operation and LVCMOS operation. In addition, LVCMOS output operation supports 1.5V  $V_{CCO}$ .

A global OE input pin is provided. If the OE pin is negated (Low), then all outputs will be in a high-impedance state. If the OE pin is asserted (High), then each output will behave as indicated by its individual register enable bit. Using the global OE pin to enable or disable outputs will not result in any 'runt' clock pulses on the outputs.

Each output bank may be enabled or disabled using the SYNC\_DISx register bit. Using these bits to enable or disable outputs will not result in any 'runt' clock pulses on the outputs.

Individual outputs within a bank may be enabled or disabled using the DIS\_Qxm register bits. These bits however may result in 'runt' pulses on the outputs if the output is otherwise enabled, so it is recommended that the entire bank be disabled via the appropriate SYNC\_DISx register bit while an individual output is being enabled using the DIS\_Qxm bit to avoid a possible 'runt' pulse on the output. If 'runt' pulses are not a concern, then the DIS\_Qxm bits may be used directly.

#### **LVCMOS Operation**

When a given output is configured to provide LVCMOS levels, then both the Q and nQ outputs will toggle at the selected output frequency. All the previously described configuration and control apply equally to both outputs. Frequency, voltage levels and enable / disable status apply to both the Q and nQ pins.

When configured as LVCMOS, the Q & nQ outputs can be selected to be phase-aligned with each other or inverted relative to one another. Phase-aligned outputs will have increased simultaneous switching currents which can negatively affect phase noise performance and power consumption. It is recommended that use of this selection be kept to a minimum.

#### **Power-Saving Modes**

To allow the device to consume the least power possible for a given application, the following functions are included under register control:

- Any unused output can be individually powered-off.
- If either bank is completely unused, all logic, including the dividers for that bank may be completely powered-off.
- Clock gating on logic that is not being used.

#### **Device Start-up Behavior**

The device will power-up with all outputs enabled in LVDS mode and all dividers bypassed.

## **Serial Control Port Description**

#### **Serial Control Port Configuration Description**

The device has a serial control port capable of responding as a slave in an  $I^2C$  or SPI compatible configuration, to allow access to any of the internal registers for device programming or examination of internal status. All registers are configured to have default values. See the specifics for each register for details. Selection of  $I^2C$  versus SPI protocol will be done via the nI2C/SPI input pin.

#### **SPI Mode Operation**

SPI mode can be enabled via pin selection from power-up. The following information assumes SPI mode has been selected.

In a read operation (R/W bit is '1'), data on SDO will be clocked out on the falling edge of SCLK.

In a write operation (R/W bit is '0'), data on SDI will be clocked in on the rising edge of SCLK.

#### Figure 3: SPI Read Sequencing Diagram

Read (LSB first)



During SPI Write operations, the user may continue to hold nCS low and provide further bytes of data for up to a total of 16 bytes in a single block write. Data is written directly into the appropriate register as it is received.



#### Figure 5: SPI Read/Write Timing Diagram



#### Table 3: Timing Characteristics in SPI Mode<sup>[a]</sup>

Symbol	Parameter	Min	Тур	Мах	Unit
t <sub>pw</sub>	SCLK Period	20			ns
t <sub>pw1</sub>	SCLK Pulse Width Low	8			ns
t <sub>pw2</sub>	SCLK Pulse Width High	8			ns
t <sub>su1</sub>	Valid nCS to SCLK Rising Setup Time	10			ns
t <sub>h1</sub>	Valid nCS After Valid SCLK Hold Time (CLKE = 0/1)	10			ns
t <sub>su2</sub>	Valid SDI to SCLK Rising Setup Time	5			ns
t <sub>h2</sub>	Valid SDI after valid SCLK Hold Time	5			ns
t <sub>d1</sub>	SCLK falling (rising in CLKE = 1 case) to Valid Data Delay Time			5	ns
t <sub>d2</sub>	nCS rising edge to SDO High Impedance Delay Time			10	ns
t <sub>csh</sub>	Time between Consecutive Read-Read or Read-Write Accesses (nCS rising edge to nCS falling edge)	20			ns

a. Specifications guaranteed by design and characterization.

# I<sup>2</sup>C Mode Operation

The  $I^2C$  interface is designed to fully support v1.0 of the  $I^2C$  specification for *normal* and *fast* mode operation. The device acts as a slave device on the  $I^2C$  bus at 100kHz or 400kHz using an address of 110110x (binary), where the value of 'x' is set by the SA0/nCS input pin. The interface accepts byte-oriented block write and block read operations. One address byte specifies the register address of the byte position of the first register to write or read. Data bytes (registers) are accessed in sequential order from the lowest to the highest byte (most significant bit first). Read and write block transfers can be stopped after any complete byte transfer. During a write operation, data will be written to the registers directly as each byte is received.

For full electrical I<sup>2</sup>C compliance, it is recommended to use external pull-up resistors for SDATA and SCLK. The internal pull-up resistors have a size of  $51k\Omega$  typical.

#### Figure 6: Slave Read and Write Cycle Sequencing

Curr	ent Read																	
S	Dev Addr + R	А	Data 0	А	Data 1	A	000	A	Data n	A	Р							
Sequ	uential Read																	
s	Dev Addr + W	A	Offset Addr M	SB A	Sr	Dev /	Addr + R	A	Data 0	А	Da	ita 1	А	000	А	Data n	А	Р
Sequ	ential Write																	
S	Dev Addr + W	A	Offset Addr M	SB A	Da	ta O	A	Data 1	А	000	А	Data ı	n	A	Р			
	from master from slave to			Sr = <u>A</u> = A =	start = repeat = acknow = none ac = stop	ledge												

# **Register Descriptions**

#### Table 4: Register Blocks

Register Ranges Offset (Hex)	Register Block Description
0 - 1	Device control
2 – 5	Bank A control
6 – 9	Bank B control
A – B	Reserved
C – F	Divide ratios

#### Table 5: Device Control Register Bit Field Locations

Address (Hex)	E	07	D6		D5	D4	D3	D2	D1	D0	
0	CLK	NODE	С	CLK_SEL[1:0]		SEL_REG	Rsvd	Rsvd	Rsvd	DIV_SYNC	
1	BKA	A_Vx	BKB_	3_Vx SYNC_DISA		SYNC_DISB	PWR_DNA	PWR_DNB	DIS_DIVA	DIS_DIVB	
Bit Field Nam	ne	Field	Туре	Defa	ault Value			Description			
CLKMODE	CLKMODE R/W		N		Ob	Clock switchover mode selection: 0 = Forced clock switch (may result in glitches as clocks switch) 1 = Glitch-less clock switch (may remain on original clock source if that source is no longer toggling)					
CLK_SEL[1:(	CLK_SEL[1:0] R/W		N		00ь	Select which input clock is to be used as the reference clock. These bit only in effect when SEL_REG = 1: 00 = CLK0, nCLK0 input drives both Divider A & Divider B 01 = CLK1, nCLK1 input drives Divider A & CLK0, nCLK0 drives Divider 10 = CLK0, nCLK0 input drives Divider A & CLK1, nCLK1 drives Divider 11 = CLK1, nCLK1 input drives both Divider A & Divider B					
SEL_REG		R/W		Ob         Determines if input clock selection is to be performed by pi           0b         0 = CLK_SEL input pin controls reference selection mux           1 = CLK_SEL register bits controls reference selection mux					tion mux	register:	
Rsvd		R/	N		_	Reserved. Always write '0' to this bit location. Read values are not defi					
DIV_SYNC	DIV_SYNC R/W		N		Ob	Divider synchronization control: 0 = Dividers running normally 1 = Dividers in reset (output clocks halted) 1->0 transition on this bit will synchronize the Bank A & Bank B output dividers					
BKA_Vx	BKA_Vx R/W		N		0b	Bank A voltage setting for optimal performance: $0 = V_{CCOA}$ is 3.3V $1 = V_{CCOA}$ is 2.5V,1.8V, and 1.5V					
BKB_Vx		R/	R/W		0b	Bank B voltage s 0 = V <sub>CCOB</sub> is 3.3 1 = V <sub>CCOB</sub> is 2.5	V	al performance:	:		
SYNC_DISA	/NC_DISA R/W		N	0b C		Glitch-free output enable bit for Bank A outputs: 0 = Outputs in Bank A are enabled glitch-lessly as indicated by their individual DIS_QAm bits 1 = All outputs for Bank A are high-impedance					

 SYNC\_DISB
 R/W
 0b
 Glitch-free output enable bit for Bank B outputs:
 0 = Outputs in Bank B are enabled glitch-lessly as indicated by their individual DIS\_QBm bits

 1 = All outputs for Bank B are high-impedance

Bit Field Name	Field Type	Default Value	Description
PWR_DNA	R/W	Ob	Power-down control for Bank A outputs: 0 = All outputs for Bank A are powered (SYNC_DISA should be 1 when powering-up the bank to prevent glitches on the output) 1 = All outputs in Bank A are powered-off
PWR_DNB	R/W	Ob	Power-down control for Bank B outputs: 0 = All outputs for Bank B are powered (SYNC_DISB should be 1 when powering-up the bank to prevent glitches on the output) 1 = All outputs in Bank B are powered-off
DIS_DIVA	R/W	0b	Power-down output divider for Bank A (DIVA must be set to 000h to bypass): 0 = Output divider for Bank A is powered 1 = Output divider for Bank A is powered-down
DIS_DIVB R/W		Ob	Power-down output divider for Bank B (DIVB must be set to 000h to bypass): 0 = Output divider for Bank B is powered 1 = Output divider for Bank B is powered-down

#### Table 6: Bank A Control Register Bit Field Locations

Address (Hex)	D7	D7 D6		D5	D4	D3	D2	D1	D0				
2		Rsvd			TERM_A	QA_POL3	QA_POL2	QA_POL1	QA_POL0				
3		Rsvd				MODE_QA3[2:0	]	1	MODE_QA2[2:0	]			
4		Rs	vd			MODE_QA1[2:0	]	1	MODE_QA0[2:0	]			
5	DIS_Q	QA3	DIS_	QA2	DIS_QA1	DIS_QA0		Rs	svd				
Bit Field Nar	ne	Field	Туре	Defa	ault Value			Description					
TERM_A	TERM_A R/W		W		0b	Indicates termina $0 = 33\Omega/50\Omega$ $1 = 50\Omega$	ation used on B	ank A outputs v	vhen HCSL moo	le is selected:			
QA_POLm	QA_POLm R/W			0h	Output polarity selection for output pair nQAm, QAm in LVCMOS mode: 0 = nQAm pin is inverted relative to QAm pin when in LVCMOS mode 1 = nQAm and QAm pins are in-phase when in LVCMOS mode								
MODE_QAm[2:0] R/W			010b	Output driver mo 000 = high-impe 001 = LVPECL 010 = LVDS (de 011 = LVCMOS 100 = HCSL 101 = CML 400r 110 = CML 800r 111 = Reserved	dance fault) nV swing	n for output pair	QAm, nQAm:						
DIS_QAm	S_QAm R/W			Ob	Disable output pair QAm, nQAm: 0 = Output pair QAm, nQAm is enabled (disable output bank using SYNC_DISA to prevent runt pulses when enabling) 1 = Output pair QAm, nQAm is powered-down								
Rsvd		R/	W		_	Reserved. Alway	Reserved. Always write 0 to this bit location. Read values are not defined.						

#### Table 7: Bank B Control Register Bit Field Locations

Address (Hex)	D	7	D	6	D5	D4	D3	D2	D1	D0
6			Rsvd			TERM_B	Rsvd	Rsvd	Rsvd	Rsvd
7		Rsvd				MODE_QB3[2:0	)]	1	MODE_QB2[2:0	]
8		Rs	svd			MODE_QB1[2:0	)]	1	MODE_QB0[2:0	)]
9	DIS_	QB3	DIS_	QB2	DIS_QB1	DIS_QB0		Rs	svd	
Bit Field Nar	ne	Field	Туре	Defa	ault Value	Description				
TERM_B	TERM_B R/W		W		Ob	Indicates termination used on Bank B outputs when HCSL mode is selected $0 = 33\Omega/50\Omega$ 1 = 50 $\Omega$				
MODE_QBm[2:0] R/W			010b	Output driver mo 000 = high-impe 001 = LVPECL 010 = LVDS (de 011 = Rsvd 100 = HCSL 101 = CML 400r 110 = CML 800r 111 = Reserved	dance fault) nV swing nV swing	n for output pair	QBm, nQBm:			
DIS_QBm		R/	W	0b		Disable output pair QBm, nQBm: 0 = Output pair QBm, nQBm is enabled (disable output bank using SYNC_DISB to prevent runt pulses when enabling) 1 = Output pair QBm, nQBm is powered-down				
Rsvd	Rsvd R/W		W		_	Reserved. Always write 0 to this bit location. Read values are not defin				

#### Table 8: Divide Ratio Register Field Locations

Address (Hex)	D7	D6	D5	D4	D3	D2	D1	D0		
С				DIVA	[7:0]		1			
D				DIVB	[7:0]					
E		Rsvd DIVB[8] DIVA[8]								
F		Rsvd								
Bit Field Nam	e Field	Type Defa		Description						
DIVA[8:0]	R/	W	000h	<ul> <li>Divider ratio for Bank A outputs:</li> <li>00h - 01h = Bypass divider and pass reference clock directly to the outputs</li> <li>02h - 1FFh = ratio to be used by the A divider is value written her example writing a 4 in this field will results in a divide ratio of 4 be</li> </ul>				here. For		
DIVB[8:0]	R/	W	000h	<ul> <li>Divider ratio for Bank B outputs:</li> <li>00h - 01h = Bypass divider and pass reference clock directly to the Bar outputs</li> <li>02h - 1FFh = ratio to be used by the B divider is value written here. For example writing a 4 in this field will results in a divide ratio of 4 being us</li> </ul>						
Rsvd	R/	W	_	Reserved. Always write 0 to this bit location. Read values are not defined.						

## **Absolute Maximum Ratings**

The absolute maximum ratings are stress ratings only. Stresses greater than those listed below can cause permanent damage to the device. Functional operation of the 8P79818 at absolute maximum ratings is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

#### **Table 9: Absolute Maximum Ratings**

Item	Rating
Supply voltage, V <sub>CCX</sub> <sup>[a]</sup> to GND	3.6V
Inputs SCLK, SDATA/SDI, SA0/nCS, CLK_SEL, CLK0, nCLK0, CLK1, nCLK1, OE, nI2C/SPI	-0.5V to 3.6V
Outputs, I <sub>O</sub> QA[0:3], nQA[0:3], QB[0:3], nQB[0:3]	
Continuous current	40mA
Surge current	60mA
Outputs, V <sub>O</sub> QA[0:3], nQA[0:3], QB[0:3], nQB[0:3]	-0.5V to 3.6V
Outputs, V <sub>O</sub> SDO, SDATA/SDI	-0.5V to 3.6V
Operating junction temperature	125°C
Storage temperature, T <sub>STG</sub>	-65°C to 150°C
Lead temperature (Soldering, 10s)	+260°C

a.  $V_{CCx}$  denotes  $V_{CC}$ ,  $V_{CCOA}$ , or  $V_{CCOB}$ .

## **DC Characteristics**

#### Table 10: DC Input/ Output Characteristics

Symbol		Paramete	r	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>in</sub>	Input capacita	ance				0.5		pF
		LVPECL				0.8		pF
		LVDS	QA[0:3], nQA[0:3]	1/[a] = 2.465 / cr 2.625 /		1.2		pF
		CML 400mV	QB[0:3], nQB[0:3]	V <sub>CCOX</sub> <sup>[a]</sup> = 3.465V or 2.625V		0.48		pF
	Power	CML 800mV	-			0.44		pF
0	dissipation	LVCMOS	Bank A	V <sub>CCOA</sub> = 3.465V or 2.625V		2.33		pF
C <sub>PD</sub>	capacitance	LVPECL				1.4		pF
	(per output)	LVDS	QA[0:3], nQA[0:3] QB[0:3], nQB[0:3]	V <sub>CCOX</sub> = 1.89V		1.5		pF
		CML 400mV				0.53		pF
		CML 800mV	*			0.3		pF
		LVCMOS	Bank A	V <sub>CCOA</sub> = 1.89V or 1.575V		2.1		pF
R <sub>PULLUP</sub>	Input pull-up	resistor				51		kΩ
R <sub>PULLDOWN</sub>	Input pull-dov	vn resistor				51		kΩ
				LVCMOS output type selected V <sub>CCOA</sub> = 3.3V <u>+</u> 5%		24		Ω
D	Output impos		QA[3:0], nQA[3:0]	LVCMOS output type selected $V_{CCOA} = 2.5V \pm 5\%$		15		Ω
R <sub>OUT</sub>		Output impedance QA[3:		LVCMOS output type selected V <sub>CCOA</sub> = 1.8V <u>+</u> 5%		26		Ω
				LVCMOS output type selected $V_{CCOA} = 1.5V \pm 5\%$		46		Ω

a.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}$ .

## Supply Voltage Characteristics

# Table 11: Power Supply Characteristics, $V_{CC}$ = 3.3V ±5%, $V_{EE}$ = 0V, $T_A$ = -40°C to +85°C<sup>[a], [b], [c]</sup>

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>CC</sub>	Core supply voltage		3.135		3.465	V
V <sub>CCOA</sub> , V <sub>CCOB</sub>	Output supply voltage		1.71		V <sub>CC</sub>	V
I <sub>CC</sub>	Core supply current			23	26	mA
I <sub>CCOA</sub> +	Output supply current	DIV-by-1		157	177	mA
I <sub>CCOB</sub>		DIV A = DIV B = 2		125	140	mA
I <sub>EE</sub>	Power supply current			183	206	mA

a. Internal dynamic switching current at maximum f<sub>OUT</sub> is included.

b. All outputs configured for LVEPCL logic levels and not terminated.

c.  $V_{CC} \ge V_{CCOA}$  and  $V_{CCOB}$ .

# Table 12: Power Supply Characteristics, $V_{CC}$ = 2.5V ±5%, $V_{EE}$ = 0V, $T_A$ = -40°C to +85°C<sup>[a], [b], [c]</sup>

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>CC</sub>	Core supply voltage		2.375		2.625	V
V <sub>CCOA</sub> , V <sub>CCOB</sub>	Output supply voltage		1.71		V <sub>CC</sub>	V
I <sub>CC</sub>	Core supply current			18	20	mA
I <sub>CCOA</sub> +	Output supply current	DIV-by-1		156	175	mA
I <sub>CCOB</sub>		DIV A = DIV B = 2		124	139	mA
I <sub>EE</sub>	Power supply current			177	199	mA

a. Internal dynamic switching current at maximum f<sub>OUT</sub> is included.

b. All outputs configured for LVEPCL logic levels and not terminated.

c.  $V_{CC} \ge V_{CCOA}$  and  $V_{CCOB}$ .

# Table 13: Power Supply Characteristics, $V_{CC}$ = 1.8V ±5%, $V_{EE}$ = 0V, $T_A$ = -40°C to +85°C<sup>[a], [b], [c]</sup>

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>CC</sub>	Core supply voltage		1.71		1.89	V
V <sub>CCOA</sub> , V <sub>CCOB</sub>	Output supply voltage		1.71		V <sub>CC</sub>	V
I <sub>CC</sub>	Core supply current			14	16	mA
I <sub>CCOA</sub> +		DIV-by-1		143	160	mA
I <sub>CCOB</sub>	Output supply current	DIV A = DIV B = 2		117	132	mA
I <sub>EE</sub>	Power supply current			161	181	mA

a. Internal dynamic switching current at maximum  $f_{\mbox{OUT}}$  is included.

b. All outputs configured for LVEPCL logic levels and not terminated.

c.  $V_{CC} \geq V_{CCOA} \text{ and } V_{CCOB}.$ 

### Table 14: Output Supply Current, $V_{CC}$ = 3.3V, 2.5V or 1.8V, $V_{EE}$ = 0V, $T_A$ = 25°C<sup>[a], [b]</sup>

Σ Σ			V <sub>CCOx</sub> <sup>[d]</sup> = 3.3V			V <sub>CCOx</sub> <sup>[d]</sup> = 2.5V			V <sub>CCOx</sub> <sup>[d]</sup> = 1.8V					$V_{CCOx}^{[d]} = 1.5V$ $V_{CCOx}^{[d]} = 1.5V + 5\%$					
Symbol	Parameter <sup>[c]</sup>	Test Conditions	LVPECL	LVDS	HCSL	LVCMOS	CML	LVPECL	LVDS	HCSL	LVCMOS	CML (400mV)	LVPECL	LVDS	HCSL	LVCMOS	CML (400mV)	LVCMOS	Units
	Bank A output	V <sub>CC</sub> = 3.3V, T <sub>A</sub> = 25°C	76	96	73	119	60	75	96	67	92	58	68	87	66	72	53	60	mA
I <sub>CCOA</sub>	supply current	V <sub>CC</sub> = 3.3V + 5%, T <sub>A</sub> = 85°C	88	114	87	149	70	88	113	85	111	67	80	104	78	86	63	71	mA
1	Bank B output	V <sub>CC</sub> = 3.3V, T <sub>A</sub> = 25°C	76	96	73	119	60	75	96	67	92	58	68	87	66	72	53	60	mA
ССОВ	supply current	V <sub>CC</sub> = 3.3V, T <sub>A</sub> = 85°C	88	114	87	149	70	88	113	85	111	67	80	104	78	86	63	71	mA

a. All outputs not terminated.

b.  $V_{CC} \ge V_{CCOA}$  and  $V_{CCOB}$ .

c. Internal dynamic switching current at maximum  ${\rm f}_{\rm OUT}$  is included.

d.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}.$ 

## **DC Electrical Characteristics**

Symbol		Parameter	Test Conditions	Minimum	Typical	Maximum	Units
			V <sub>CC</sub> = 3.3V	2.20		V <sub>CC</sub> +0.3	V
V <sub>IH</sub>	Input high voltag	e	V <sub>CC</sub> = 2.5V	1.85		V <sub>CC</sub> +0.3	V
			V <sub>CC</sub> = 1.8V	1.25		V <sub>CC</sub> +0.3	V
			V <sub>CC</sub> = 3.3V	-0.3		0.8	V
V <sub>IL</sub>	Input low voltage	9	V <sub>CC</sub> = 2.5V	-0.3		0.7	V
			V <sub>CC</sub> = 1.8V	-0.3		0.7	V
IIH	Input	SA0/nCS, SDATA/SDI, SCLK, CLK_SEL, OE	V <sub>CC</sub> = V <sub>IN</sub> = 3.465V, 2.625V or 1.89V			5	μΑ
	high current	nI2C/SPI	- 1.09V			150	μΑ
IIL	Input	SA0/nCS, SDATA/SDI, SCLK, CLK_SEL,OE	$V_{CC} = 3.465V, 2.625V \text{ or } 1.89V,$	-150			μΑ
	low current	nI2C/SPI	- V <sub>IN</sub> = 0V	-5			μΑ
			$V_{CC}$ = 3.3V ±5%, $I_{OH}$ = -5mA	2.6			V
V <sub>OH</sub>	Output high voltage	SDATA/SDI, SDO	$V_{CC}$ = 2.5V ±5%, $I_{OH}$ = -5mA	1.8			V
	ingit tottago		V <sub>CC</sub> = 1.8V ±5%, I <sub>OH</sub> = -5mA				V
V <sub>OL</sub>	Output low voltage	SDATA/SDI, SDO	$V_{CC}$ = 3.3V ±5% or 2.5V ±5% or 1.8V ±5%, I <sub>OL</sub> = 5mA			0.5	V

# Table 16: Differential Input DC Characteristics, V<sub>CC</sub> = $3.3V\pm5\%$ , $2.5V\pm5\%$ or $1.8V\pm5\%$ , V<sub>EE</sub> = 0V, T<sub>A</sub> = -40°C to +85°C

Symbol	Paran	neter	Test Conditions	Minimum	Typical	Maximum	Units
I <sub>IH</sub>	Input high current	CLKx, nCLKx <sup>[a]</sup>	V <sub>CC</sub> = V <sub>IN</sub> = 3.465V or 2.625V			150	μΑ
1	Input	CLKx <sup>[a]</sup>	V <sub>CC</sub> = 3.465V or 2.625V, V <sub>IN</sub> = 0V	-5			μΑ
Ι	low current	nCLKx <sup>[a]</sup>	V <sub>CC</sub> = 3.465V or 2.625V, V <sub>IN</sub> = 0V	-150			μΑ
V <sub>PP</sub>	Peak-to-peak volt	age <sup>[b]</sup>		0.15		1.3	V
V <sub>CMR</sub>	Common mode input voltage <sup>[b], [c]</sup>			0		V <sub>CC</sub>	V

a. CLKx denotes CLK0, CLK1. nCLKx denotes nCLK0, nCLK1.

b.  $V_{IL}$  should not be less than –0.3V.  $V_{IH}$  should not be higher than  $V_{CC}.$ 

c. Common mode voltage is defined as the cross-point.

			V <sub>CCC</sub>	$V_{CCOx}^{[a]} = 3.3V \pm 5\%$			$V_{CCOx}^{[a]} = 2.5V \pm 5\%$			$V_{CCOx}^{[a]} = 1.8V\pm5\%$		
Symbol	Paramete	er	Min	Тур	Мах	Min	Тур	Мах	Min	Тур	Мах	Units
V <sub>OH</sub>	Output high voltage <sup>[b]</sup>	Qx, nQx <sup>[c]</sup>	V <sub>CCOx</sub> - 1.30		V <sub>CCOx</sub> - 0.80	V <sub>CCOx</sub> - 1.35		V <sub>CCOx</sub> - 0.80	V <sub>CCOx</sub> - 1.50		V <sub>CCOx</sub> - 0.90	V
V <sub>OL</sub>	Output low voltage	Qx, nQx <sup>[c]</sup>	V <sub>CCOx</sub> - 2.00		V <sub>CCOx</sub> - 1.75	V <sub>CCOx</sub> - 2.00		V <sub>CCOx</sub> - 1.75	$V_{EE}$		0.25	V

Table 17: LVPECL DC Characteristics, V<sub>CC</sub> = 3.3V±5%, 2.5V±5% or 1.8V±5%, V<sub>EE</sub> = 0V, T<sub>A</sub> = -40°C to +85°C

a.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}$ .

b. Outputs terminated with 50Ω to V<sub>CCOx</sub> – 2V when V<sub>CCOx</sub> = 3.3V±5% or 2.5V±5%. Outputs terminated with 50Ω to ground when V<sub>CCOx</sub> = 1.8V±5%.

c. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3. nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 18: LVDS DC Characteristics, $V_{CC}$ = 3.3V ±5%, 2.5V ±5% or 1.8V ±5%, $V_{CCOA}$ = $V_{CCOB}$ = 3.3V ±5%, 2.5V ±5% or 1.8V ±5%, $T_A$ = -40°C to +85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OD</sub>	Differential output voltage	Qx, nQx <sup>[a]</sup>		247		480	mV
$\Delta V_{OD}$	V <sub>OD</sub> magnitude change	Qx, nQx <sup>[a]</sup>	Terminated 100 $\Omega$ across			50	mV
V <sub>OS</sub>	Offset voltage	Qx, nQx <sup>[a]</sup>	Qx and nQx	1.125		1.375	V
$\Delta V_{OS}$	V <sub>OS</sub> magnitude change	Qx, nQx <sup>[a]</sup>				50	mV

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 19: CML (400mV Swing) DC Characteristics, $V_{CC} = 3.3V \pm 5\%$ , 2.5V $\pm 5\%$ or 1.8V $\pm 5\%$ , $V_{CCOA} = V_{CCOB} = 3.3V \pm 5\%$ , 2.5V $\pm 5\%$ or 1.8V $\pm 5\%$ , $T_A = -40$ °C to +85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OH</sub>	Output high voltage	Qx, nQx <sup>[a]</sup>	<b>T</b>	V <sub>CCOx</sub> <sup>[b]</sup> – 0.10		V <sub>CCOx</sub> <sup>[b]</sup>	V
V <sub>OL</sub>	Output low voltage	Qx, nQx <sup>[a]</sup>	Terminated with 50 $\Omega$ to $V_{CCOx}^{[b]}$	V <sub>CCOx</sub> <sup>[b]</sup> – 0.50		$V_{CCOx}^{[b]} - 0.30$	V
V <sub>OUT</sub>	Output voltage swing	Qx, nQx <sup>[a]</sup>	- VCCOx'	300		500	mV

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

b.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}$ .

# Table 20: CML (800mV Swing) DC Characteristics, $V_{CC}$ = 3.3V ±5%, 2.5V ±5% or 1.8V ±5%, $V_{CCOA}$ = $V_{CCOB}$ = 3.3V ±5%, 2.5V ±5% or 1.8V ±5%, $T_A$ = -40°C to +85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OH</sub>	Output high voltage	Qx, nQx <sup>[a]</sup>		$V_{CCOx}^{[b]} - 0.10$		V <sub>CCOx</sub>	V
V <sub>OL</sub>	Output low voltage	Qx, nQx <sup>[a]</sup>	Terminated with 50 $\Omega$ to $V_{CCOx}^{[b]}$	V <sub>CCOx</sub> <sup>[b]</sup> – 0.95		V <sub>CCOx</sub> <sup>[b]</sup> – 0.70	V
V <sub>OUT</sub>	Output voltage swing	Qx, nQx <sup>[a]</sup>		575		1000	mV

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

b.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}$ .

# Table 21: LVCMOS Clock Outputs DC Characteristics, $V_{CC}$ = 3.3V±5%, 2.5V±5% or 1.8V±5%, $V_{EE}$ = 0V, $T_A$ = -40°C to +85°C

		Test	V <sub>CCO</sub>	A = 3.3	V±5%	V <sub>CCO</sub>	<sub>A</sub> = 2.5	V±5%	V <sub>CCO</sub>	<sub>A</sub> = 1.8'	V ±5%	V <sub>CCO</sub>	<sub>A</sub> = 1.5'	V ±5%	
Symbol	Parameter	Conditions	Min	Тур	Мах	Min	Тур	Мах	Min	Тур	Max	Min	Тур	Max	Units
V <sub>OH</sub>	Output high voltage QAx, nQAx <sup>[a]</sup>	I <sub>OH</sub> = -8mA	2.6			1.1			1.1			1.1			V
V <sub>OL</sub>	Output low voltage QAx, nQAx <sup>[a]</sup>	I <sub>OL</sub> = 8mA			0.5			0.5			0.5			0.5	V

a. QAm denotes QA0, QA1, QA2, QA3. nQAm denotes nQA0, nQA1, nQA2, nQA3.

### Table 22: Input Frequency Characteristics, $V_{CC}$ = 3.3V±5%, 2.5V±5% or 1.8V±5%, $T_A$ = -40°C to +85°C

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
f <sub>IN</sub>	Input frequency,	LVPECL,LVDS, HCSL, CML		1PPS		700MHz	
CLP	CLKx, nCLKx	LVCMOS		1PPS		200MHz	
idc	Input duty cycle <sup>[a]</sup>				50		%
f	Serial port clock SCLK	I <sup>2</sup> C operation		100		400	kHz
<sup>†</sup> SCLK	Senai port clock SOLK	SPI operation				50	MHz

a. Any deviation from a 50% duty cycle on the input may be reflected in the output duty cycle.

# AC Electrical Characteristics

Table 23: AC Characteristics,  $V_{CC}$  = 3.3V ±5%, 2.5V ±5% or 1.8V ±5%,  $V_{CCOA}$  =  $V_{CCOB}$  = 3.3V ±5%, 2.5V ±5% or 1.8V ±5%,  $T_A$  = -40°C to +85°C

Symbol	Parame	eter <sup>[a]</sup>	Test	Conditions	Minimum	Typical	Maximum	Units
f <sub>оит</sub>	Output frequency	LVPECL, LVDS HCSL, CML <sup>[b]</sup>			1PPS		700MHz	
		LVCMOS			1PPS		200MHz	
		LVPECL	20% to 80%		125		700	ps
			$V_{CCOx}^{[c]} = 3.3V$	20% to 80%	405		550	
		LVDS	$V_{CCOx}^{[c]} = 2.5V$	20% to 80%	- 125		550	ps
			V <sub>CCOx</sub> <sup>[c]</sup> = 1.8V	20% to 80%	175		550	ps
+ /+	Output rise and fall	CML, 400mV	20% to 80%		100		675	ps
t <sub>R</sub> / t <sub>F</sub>	times	CML, 800mV	20% to 80%		125		825	ps
			V <sub>CCOA</sub> = 3.3V	20% to 80%				
		LVCMOS	V <sub>CCOA</sub> = 2.5V	20% to 80%	200		800	ps
			V <sub>CCOA</sub> = 1.8V	20% to 80%				
			V <sub>CCOA</sub> = 1.5V	20% to 80%		650	1300	ps
		LVPECL				15	50	ps
		LVDS				20	60	ps
t <sub>sk</sub> (b)	Bank skew <sup>[d], [e], [f]</sup>	CML				10	35	ps
		HCSL				10	35	ps
		LVCMOS				50	100	ps
		LVPECL,LVDS, HCSL, CML	Even divide ratio	5	45	50	55	%
		LVPECL,LVDS, HCSL, CML	Odd divide ratios	/ bypass	43	50	57	%
ada	Output		<u> </u>	Even divide ratios	45	50	55	%
odc duty cycle <sup>[g]</sup>	duty cycle <sup>[g]</sup>	LVCMOS	V <sub>CCOA</sub> = 3.3V, 2.5V, or 1.8V	Odd divide ratios / bypass	40	50	60	%
				Even divide ratios	40	50	60	%
		LVCMOS	V <sub>CCOA</sub> = 1.5V	Odd divide ratios / bypass	38	50	62	%
MUX <sub>ISOL</sub>	Mux isolation		156.25MHz, V <sub>SW</sub>	<sub>ING</sub> = 800mV		61		dB
	Noise floor		Offset >10MHz fr	om156.25MHz carrier		-154		dBc/Hz

a. Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

- b. CML denotes CML 400mV and CML 800mV, unless otherwise stated.
- c.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}$ .

- d. This parameter is guaranteed by characterization. Not tested in production.
- e. This parameter is defined in accordance with JEDEC Standard 65.
- f. Defined as skew within a bank of outputs at the same supply voltage and with equal load conditions.
- g. Measured using 50% duty cycle on input reference.

# Table 24: HCSL AC Characteristics, $V_{CC} = 3.3V \pm 5\%$ , 2.5V $\pm 5\%$ or 1.8V $\pm 5\%$ , $V_{CCOA} = V_{CCOB} = 3.3V \pm 5\%$ , 2.5V $\pm 5\%$ or 1.8V $\pm 5\%$ , $T_A = -40^{\circ}$ C to $+85^{\circ}$ C

Symbol	Parameter <sup>[a]</sup>	Test Conditions <sup>[b]</sup>	Minimum	Typical	Maximum	Units
t	Rise/ fall edge rate <sup>[c]</sup>	V <sub>CCOx</sub> = 3.3V or 2.5V	0.6		4	V/ns
t <sub>SLEW</sub> Ri		V <sub>CCOx</sub> = 1.8V	0.45		4	V/ns
V <sub>MAX</sub>	Absolute max. output voltage <sup>[d], [e]</sup>	V <sub>CCOx</sub> = 3.3V, 2.5V, 1.8V			1150	mV
V <sub>MIN</sub>	Absolute min. output voltage <sup>[d], [f]</sup>	V <sub>CCOx</sub> = 3.3V, 2.5V, 1.8V	-150			mV
V <sub>CROSS</sub>	Absolute crossing voltage <sup>[g], [h]</sup>	V <sub>CCOx</sub> = 3.3V, 2.5V, 1.8V			550	mV
$\Delta V_{CROSS}$	Total variation of $V_{CROSS}$ over all edges <sup>[g], [i]</sup>	V <sub>CCOx</sub> = 3.3V, 2.5V, 1.8V			140	mV

a. Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

- b.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}$ .
- c. Measured from -150mV to +150mV on the differential waveform (derived from Qx minus nQx). The signal must be monotonic through the measurement region for rise and fall time. The 300mV measurement window is centered on the differential zero crossing.
- d. Measurement taken from single ended waveform.
- e. Defined as the maximum instantaneous voltage including overshoot.
- f. Defined as the minimum instantaneous voltage including undershoot.
- g. Measured at crossing point where the instantaneous voltage value of the rising edge of Qm equals the falling edge of nQm.
- h. Refers to the total variation from the lowest crossing point to the highest, regardless of which edge is crossing. Refers to all crossing points for this measurement.
- i. Defined as the total variation of all crossing voltages of rising Qm and falling nQm, This is the maximum allowed variance in V<sub>CROSS</sub> for any particular system.

# Table 25: Additive Jitter, $V_{CC}$ = 3.3V, 2.5V or 1.8V, $V_{CCOA}$ = $V_{CCOB}$ = 3.3V, 2.5V, 1.8V or 1.5V (1.5V only supported for LVCMOS outputs), $T_A$ = 25°C

Symbol	Paramete	er	Т	est Conditions <sup>[a]</sup>	Minimum	Typical	Maximum	Units
			f <sub>OUT</sub> =	V <sub>CCOx</sub> <sup>[b]</sup> = 3.3V or 2.5V		77	92	fs
		LVPECL	156.25MHz	V <sub>CCOx</sub> <sup>[b]</sup> = 1.8V		90	117	fs
		LVFLOL	f <sub>OUT</sub> =	V <sub>CCOx</sub> <sup>[b]</sup> = 3.3V or 2.5V		50	60	fs
			625MHz	V <sub>CCOx</sub> <sup>[b]</sup> = 1.8V		60	84	fs
			f <sub>OUT</sub> =	V <sub>CCOx</sub> <sup>[b]</sup> = 3.3V or 2.5V		85	104	fs
		r LVDS	156.25MHz	V <sub>CCOx</sub> <sup>[b]</sup> = 1.8V		126	185	fs
	RMS additive jitter		f <sub>OUT</sub> = 625MHz	V <sub>CCOx</sub> <sup>[b]</sup> = 3.3V or 2.5V		48	61	fs
t <sub>jit</sub> (f)	(random); Integration range:			V <sub>CCOx</sub> <sup>[b]</sup> = 1.8V		57	84	fs
	12kHz – 20MHz		f <sub>OUT</sub> = 156.25MHz	V <sub>CCOx</sub> <sup>[b]</sup> = 3.3V or 2.5V		92	132	fs
				V <sub>CCOx</sub> <sup>[b]</sup> = 1.8V		92	133	fs
		TICOL	f <sub>OUT</sub> =	V <sub>CCOx</sub> <sup>[b]</sup> = 3.3V or 2.5V		61	73	fs
			625MHz	V <sub>CCOx</sub> <sup>[b]</sup> = 1.8V		67	93	fs
			V <sub>CCOA</sub> = 3.3V or 2.5V		98	166	fs	
		LVCMOS <sup>f</sup> out = 156.25	f <sub>OUT</sub> = 156.25MHz	V <sub>CCOA</sub> = 1.8V		128	204	fs
				V <sub>CCOA</sub> = 1.5V		198	314	fs

a. All outputs configured for the specific output type, as shown in the table.

b.  $V_{CCOx}$  denotes  $V_{CCOA}$  and  $V_{CCOB}.$ 

## **Applications Information**

### **Recommendations for Unused Input and Output Pins**

#### Inputs:

#### CLK/nCLK Inputs

For applications not requiring the use of the differential input, both CLK and nCLK can be left floating. Though not required, but for additional protection, a  $1k\Omega$  resistor can be tied from CLK to ground.

#### LVCMOS Control Pins

All control pins have internal pullups and pulldowns; additional resistance is not required but can be added for additional protection. A 1k $\Omega$  resistor can be used.

#### Outputs:

#### LVCMOS Outputs

All unused LVCMOS outputs can be left floating. It is recommended that there is no trace attached.

#### LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

#### LVDS Outputs

All unused LVDS output pairs can be either left floating or terminated with  $100\Omega$  across. If they are left floating, there should be no trace attached.

#### Differential Outputs

All unused Differential outputs can be left floating. It is recommended that there is no trace attached.

### **Power Dissipation and Thermal Considerations**

The 8P79818 is a multi-functional, high speed device that targets a wide variety of clock frequencies and applications. Since this device is highly programmable with a broad range of features and functionality, the power consumption will vary as each of these features and functions is enabled.

The 8P79818 device was designed and characterized to operate within the ambient industrial temperature range of -40°C to +85°C. The ambient temperature represents the temperature around the device, not the junction temperature. When using the device in extreme cases, such as maximum operating frequency and high ambient temperature, external air flow may be required in order to ensure a safe and reliable junction temperature. Extreme care must be taken to avoid exceeding 125°C junction temperature.

The power calculation examples below were generated using a maximum ambient temperature and supply voltage. For many applications, the power consumption will be much lower. Please contact IDT technical support for any concerns on calculating the power dissipation for your own specific configuration.

#### **Power Domains**

The 8P79818 device has a number of separate power domains that can be independently enabled and disabled via register accesses (all power supply pins must still be connected to a valid supply voltage). Figure 7 below indicates the individual domains and the associated power pins.

#### Figure 7: 8P79818 Power Domains



### **Power Consumption Calculation**

Determining total power consumption involves several steps:

1.Determine the power consumption using maximum current values for core voltage from Table 11, Table 12 and Table 13, Page 18 for the appropriate case of how many dividers are enabled.

2. Determine the nominal power consumption of each enabled output path.

a. This consists of a base amount of power that is independent of operating frequency, as shown in Table 27 through Table 43 (depending on the chosen output protocol).

b. Then there is a variable amount of power that is related to the output frequency. This can be determined by multiplying the output frequency by the FQ\_Factor shown in Table 27 through Table 43.

3.All of the above totals are then summed.

### **Example Calculations**

#### Table 26: Example 1. Common Customer Configuration (3.3V Core Voltage)

Bank	Configuration	Frequency (MHz)	V <sub>CCO</sub>
Bank A	LVDS	125	3.3V
Bank B	LVDS	125	2.5V

- Core supply current, I<sub>CC</sub> = 24.7mA (max.)
- Output supply current, Bank A = 0.06 \* 125 + 71.615 = 79.115mA
- Output supply current, Bank B = 0.06 \* 125 + 71.615 = 79.115mA
- Total device current = 24.7mA + 79.115mA + 79.115mA = 182.93mA
- Total device power = 3.465V \* 183.93mA = 633.934mW

With an ambient temperature of 85°C and no airflow, the junction temperature is:

 $T_J = 85^{\circ}C + 35.23^{\circ}C/W * 0.634W = 107.3^{\circ}C$ 

### **Thermal Considerations**

Once the total power consumption has been determined, it is necessary to calculate the maximum operating junction temperature for the device under the environmental conditions it will operate in. Thermal conduction paths, air flow rate and ambient air temperature are factors that can affect this. The thermal conduction path refers to whether heat is to be conducted away via a heat-sink, via airflow or via conduction into the PCB through the device pads (including the ePAD). Thermal conduction data is provided for typical scenarios in Table 44, Page 30. Please contact IDT for assistance in calculating results under other scenarios.

#### **Current Consumption Data and Equations**

#### Table 27: 3.3V LVDS Output Calculation Table

LVDS	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.06	71.615

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 28: 2.5V LVDS Output Calculation Table

LVDS	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.06	71.544

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3. nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 29: 1.8V LVDS Output Calculation Table

LVDS	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.1	50.284

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 30: 3.3V LVPECL Output Calculation Table

LVPECL	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.05	53.475

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 31: 2.5V LVPECL Output Calculation Table

LVPECL	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.04	20.874

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 32: 1.8V LVPECL Output Calculation Table

LVPECL	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.04	19.962

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 33: 3.3V HCSL Output Calculation Table

HCSL	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.05	54.911

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 34: 3.3V CML Output (400mV) Calculation Table

CML	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.03	51.889

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 35: 2.5V CML Output (400mV) Calculation Table

CML	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.02	49.220

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 36: 1.8V CML Output (400mV) Calculation Table

CML	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.02	47.326

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3. nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 37: 3.3V CML Output (800mV) Calculation Table

CML	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.02	51.474

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 38: 2.5V CML Output (800mV) Calculation Table

CML	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.02	48.906

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

# Table 39: 1.8V CML Output (800mV) Calculation Table

CML	FQ_Factor (mA/MHz), per output	Base_Current (mA)
Qx, nQx <sup>[a]</sup>	0.02	47.334

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3. nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 40: 3.3V LVCMOS Output Calculation Table

LVCMOS	Base_Current (mA)	
Qx, nQx <sup>[a]</sup>	62.289	

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 41: 2.5V LVCMOS Output Calculation Table

LVCMOS	Base_Current (mA)	
Qx, nQx <sup>[a]</sup>	51.097	

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 42: 1.8V LVCMOS Output Calculation Table

LVCMOS	Base_Current (mA)	
Qx, nQx <sup>[a]</sup>	47.745	

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

#### Table 43: 1.5V LVCMOS Output Calculation Table

LVCMOS	Base_Current (mA)	
Qx, nQx <sup>[a]</sup>	41.485	

a. Qx denotes QA0, QA1, QA2, QA3, QB0, QB1, QB2, QB3.
 nQx denotes nQA0, nQA1, nQA2, nQA3, nQB0, nQB1, nQB2, nQB3.

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Applying the values to the following equation will yield output current by frequency:  $(mA) = FQ_Factor * Frequency (MHz) + Base_Current where:$ 

Qx Current is the specific output current according to output type and frequency

FQ\_Factor is used for calculating current increase due to output frequency

Base\_Current is the base current for each output path independent of output frequency

The second step is to multiply the power dissipated by the thermal impedance to determine the maximum power gradient, using the following equation:

 $\mathsf{T}_\mathsf{J} = \mathsf{T}_\mathsf{A} + (\theta_\mathsf{J}\mathsf{A} * \mathsf{Pd}_\mathsf{total})$ 

where:

 $T_J$  is the junction temperature (°C)

 $T_A$  is the ambient temperature (°C)

θ<sub>JA</sub> is the thermal resistance value from Table 44, Page 30, dependent on ambient airflow (°C/W)

Pd<sub>total</sub> is the total power dissipation of the 8P79818 under usage conditions, including power dissipated due to loading (W)

Note that for LVPECL outputs the power dissipation through the load is assumed to be 27.95mW. When selecting LVCMOS outputs, power dissipation through the load will vary based on a variety of factors including termination type and trace length. For these examples, power dissipation through loading will be calculated using  $C_{PD}$  (found in Table 10, Page 16) and output frequency:

 $Pd_{OUT} = C_{PD} * f_{OUT} * V_{CCO}^2$ 

where:

Pdout is the power dissipation of the output (W)

C<sub>PD</sub> is the power dissipation capacitance (pF)

f<sub>OUT</sub> is the output frequency of the selected output (MHz)

 $V_{CCO}$  is the voltage supplied to the appropriate output (V)

#### Table 44: $\theta_{\text{JA}}$ vs. Air Flow Table for a 32-lead 5mm x 5mm VFQFN

$\theta_{JA}$ vs. Air Flow			
Meters per Second	0	1	2
Multi-Layer PCB, JEDEC Standard Test Boards	35.23°C/W	31.6°C/W	30.0°C/W

## Package Drawings

#### Figure 8: Package Drawings



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### Figure 9: Package Drawings (Continued)



# Marking Diagram

IDT 8P79818 NLGI #YYWW\$	<ol> <li>Line 1 is the prefix of the part number.</li> <li>Line 2 and Line 3 is the part number.</li> <li>Line 4 "YYWW\$"         <ul> <li>a. "YY" are the last digits of the year and "WW" is the work week that the part was assembled.</li> <li>b. "\$" denotes mark code.</li> </ul> </li> </ol>
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# **Ordering Information**

#### Table 45: Ordering Information

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8P79818NLGI	IDT8P79818NLGI	32-lead VFQFN, Lead Free	Tray	-40°C to +85°C
8P79818NLGI8	IDT8P79818NLGI	32-lead VFQFN, Lead Free	Tape & Reel	-40°C to +85°C
8P79818NLGI/W	IDT8P79818NLGI	32-lead VFQFN, Lead Free	Tape & Reel	-40°C to +85°C

#### Table 46: Pin 1 Orientation in Tape and Reel Packaging

Part Number Suffix	Pin 1 Orientation	Illustration
NLG18	Quadrant 1 (EIA-481-C)	Correct Pin 1 ORIENTATION (Round Sprocket Holes) COOODOCOCOCOCOCOCOCOCOCOCOCOCOCOCOCOCOC
NLGI/W	Quadrant 2 (EIA-481-D)	Correct Pin 1 ORIENTATION CARRIER TAPE TOPSIDE (Round Sprocket Holes)

### **Revision History**

Revision Date	Description of Change		
December 19, 2016	Initial datasheet.		



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