

DATA SHEET

TEA1103; TEA1103T; TEA1103TS

**Fast charge ICs for NiCd and NiMH
batteries**

Preliminary specification
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Fast charge ICs for NiCd and NiMH batteries

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FEATURES

- Safe and fast charging of Nickel Cadmium (NiCd) and Nickel Metal Hydride (NiMH) batteries
- Pin compatible with the TEA1102x, fast charge ICs for Lilon, SLA, NiCd and NiMH batteries
- Three charge states for NiCd or NiMH; fast, top-off and trickle or voltage regulation (optional)
- Adjustable fast charge current [0.5CA to 5CA nominal (CA = Capacity Amperes)]
- DC top-off and pulsating trickle charge current (NiCd and NiMH)
- Temperature dependent $\Delta T/\Delta t$ battery full detection
- Automatic switch-over to accurate peak voltage detection ($-1/4\%$) if no NTC is applied
- Possibility to use both $\Delta T/\Delta t$ and peak voltage detection as main fast charge termination
- Support of inhibit during all charging states
- Manual refresh with regulated adjustable discharge current (NiCd and NiMH)
- Voltage regulation in the event of no battery
- Support of battery voltage based charge indication and buzzer signalling at battery insertion, end of refresh and at full detection
- Single, dual and separate LED outputs for indication of charge status state
- Minimum and maximum temperature protection
- Time-out protection
- Short-circuit battery voltage protection
- Can be applied with few low-cost external components.

GENERAL DESCRIPTION

The TEA1103x are fast charge ICs which are able to fast charge NiCd and NiMH batteries.

The main fast charge termination for NiCd and NiMH batteries are $\Delta T/\Delta t$ and peak voltage detection, both of which are well proven techniques. The TEA1103x automatically switches over from $\Delta T/\Delta t$ to peak voltage detection if the thermistor fails or is not present. The $\Delta T/\Delta t$ detection sensitivity is temperature dependent, thus avoiding false charge termination. Three charge states can be distinguished; fast, top-off and trickle.

Several LEDs, as well as a buzzer, can be connected to the TEA1103x for indicating battery insertion, charge states, battery full condition and protection mode.

The TEA1103x are contained in a 20-pin package and are manufactured in a BiCMOS process, essentially for integrating the complex mix of requirements in a single chip solution. Only a few external low cost components are required in order to build a state of the art charger.

The TEA1103x are pin compatible with the TEA1102x, fast charge ICs for Lilon, SLA, NiCd and NiMH batteries.

ORDERING INFORMATION

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
TEA1103	DIP20	plastic dual in-line package; 20 leads (300 mil)	SOT146-1
TEA1103T	SO20	plastic small outline package; 20 leads; body width 7.5 mm	SOT163-1
TEA1103TS	SSOP20	plastic shrink small outline package; 20 leads; body width 5.3 mm	SOT339-1

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QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_P	supply voltage		5.5	–	11.5	V
I_P	supply current	outputs off	–	4	–	mA
$\Delta V_{NTC}/V_{NTC}$	temperature rate dependent ($\Delta T/\Delta t$) detection level	$V_{NTC} = 2\text{ V};$ $T_j = 0\text{ to }50\text{ }^\circ\text{C}$	–	–0.25	–	%
$\Delta V_{bat}/V_{bat}$	voltage peak detection level with respect to top value	$V_{bat} = 2\text{ V};$ $T_j = 0\text{ to }50\text{ }^\circ\text{C}$	–	–0.25	–	%
I_{Vbat}	input current battery monitor	$V_{bat} = 0.3\text{ to }1.9\text{ V}$	–	1	–	nA
$V_{bat(l)}$	voltage at pin 19 for detecting low battery voltage		–	0.30	–	V
I_{IB}	battery charge current	fast charge	10	–	100	μA
		top-off mode	–	3	–	μA
$I_{IB(max)}$	maximum battery charge current	voltage regulation full NiCd and NiMH battery	–	10	–	μA
$I_{IB(Lmax)}$	maximum load current	no battery	–	40	–	μA
f_{osc}	oscillator frequency		10	–	200	kHz
V_{reg}	regulating voltage	NiCd and NiMH (pin V_{stb} open-circuit)	–	1.325 or V_{stb}	–	V
		open battery	–	1.9	–	V

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BLOCK DIAGRAM

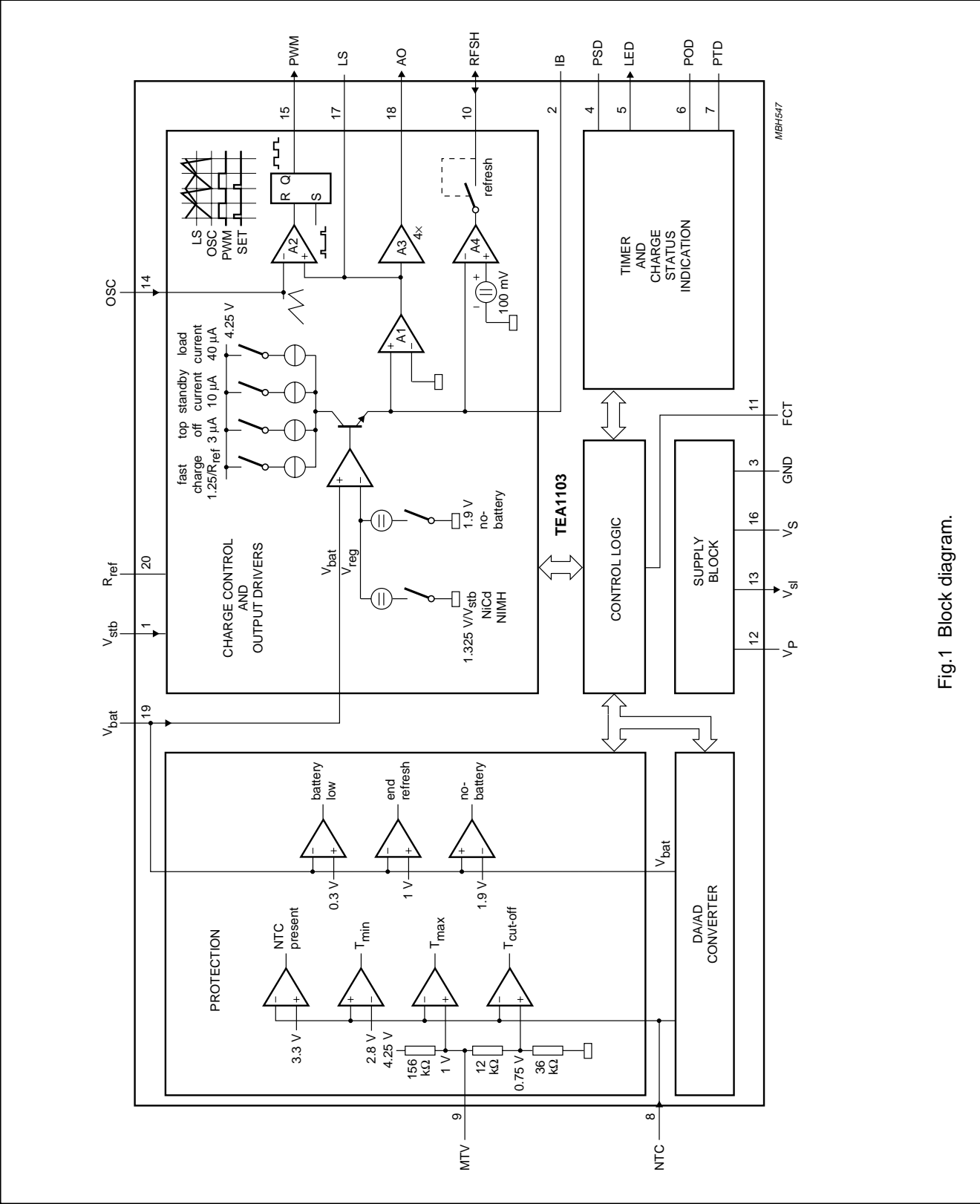


Fig.1 Block diagram.

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PINNING

SYMBOL	PIN	DESCRIPTION
V _{stb}	1	standby regulation voltage input (NiCd and NiMH)
IB	2	charge current setting
GND	3	ground
PSD	4	program pin sample divider
LED	5	LED output
POD	6	program pin oscillator divider
PTD	7	program pin time-out divider
NTC	8	temperature sensing input
MTV	9	maximum temperature voltage
RFSH	10	refresh input/output
FCT	11	fast charge termination and battery chemistry identification
V _P	12	positive supply voltage
V _{sl}	13	switched reference voltage output
OSC	14	oscillator input
PWM	15	pulse width modulator output
V _S	16	stabilized reference voltage
LS	17	loop stability pin
AO	18	analog output
V _{bat}	19	single-cell battery voltage input
R _{ref}	20	reference resistor pin

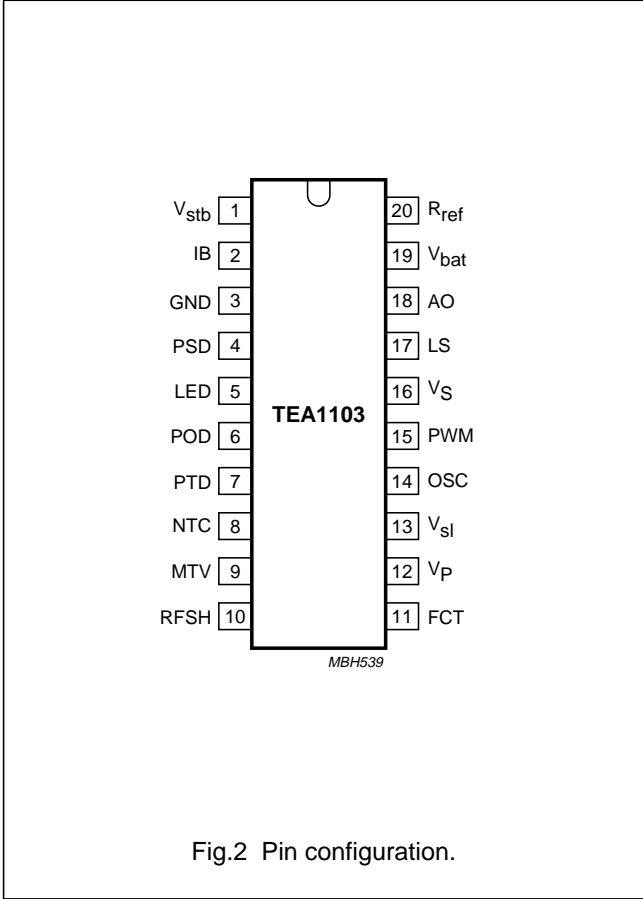


Fig.2 Pin configuration.

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INTRODUCTION

All battery types are initially fast charged with an adjustable high current. Fast charge termination depends upon the battery type. With NiCd and NiMH batteries the main fast charge termination will be the $\Delta T/\Delta t$ (temperature detection) and/or peak voltage detection.

The fast charge period is followed by a top-off period for NiCd and NiMH batteries. During the top-off period the NiCd and NiMH batteries are charged to maximum capacity by reduced adjustable charge current.

The top-off period ends after time-out or one hour respectively.

After the top-off period, the TEA1103x switches over to the standby mode. For NiCd and NiMH batteries either the voltage regulation or trickle charge mode can be selected. The voltage regulation mode is selected when the battery includes a fixed load. Trickle charge prevents a discharge of the battery over a long period of time.

Charging principles

CHARGING NiCd/NiMH BATTERIES

Fast charging of the battery begins when the power supply voltage is applied and at battery insertion.

During fast charge of NiCd and NiMH batteries, the battery temperature and voltage are monitored. Outside the initialized temperature and voltage window, the system switches over to the top-off charge current.

The TEA1103x supports detection of fully charged NiCd and NiMH batteries by either of the following criteria:

- $\Delta T/\Delta t$
- Voltage peak detection.

If the system is programmed with $\Delta T/\Delta t$ and V_{peak} or, $\Delta T/\Delta t$ or V_{peak} as the main fast charge termination, it automatically switches to voltage peak detection if the battery pack is not provided with a temperature sensing input (NTC). In this way both packages, with and without temperature sensor, can be used randomly independent of the applied full detection method. Besides $\Delta T/\Delta t$ and/or voltage peak detection, fast charging is also protected by temperature cut-off and time-out.

To avoid false fast charge termination by peak voltage detection or $\Delta T/\Delta t$, full detection is disabled during a short hold-off period at the start of a fast charge session. After fast charge termination, the battery is extra charged by a top-off period. During this period of approximately one hour, the charge current is lowered thus allowing the

battery to be charged to nearly 100% before the system switches over to standby.

After the battery has been charged to nearly 100% by the top-off period, discharge of the battery (caused by a load or by the self-discharge) can be avoided by voltage regulation or by trickle charge.

If batteries are charged in combination with a load, the TEA1103x can be programmed to apply voltage regulation during the standby mode. In this way, discharge of the battery caused by self-discharge or by an eventual load is avoided. The regulating voltage is adjustable to the voltage characteristic of the battery. For battery safety the charge current is limited and the temperature is monitored during voltage regulation. If a trickle charge is applied, the self-discharge of the battery will be compensated by a pulsating charge current.

To avoid the so called 'memory effect' in NiCd batteries, a refresh can be manually activated. The discharge current is regulated by the IC in combination with an external power transistor. After discharging the battery to 1 V per cell, the system automatically switches over to fast charge.

FUNCTIONAL DESCRIPTION

Control logic

The main function of the control logic is to support the communication between several blocks. It also controls the charge method, initialization and battery full detection. The block diagram of the TEA1103x is illustrated in Fig.1.

Conditioning charge method and initializations

At system switch-on, or at battery insertion, the control logic sets the initialization mode in the timer block. After the initialization time the timer program pins can be used to indicate the charging state using several LEDs. The charge method is defined at the same time by the following methods:

- If the FCT pin is floating, the system will charge the battery according to the charge characteristic of NiCd and NiMH batteries.
- The standby charge method (NiCd and NiMH), trickle charge or voltage regulation, is defined by the input pin V_{stb} . By biasing this voltage with a set voltage, the output voltage will be regulated to the V_{stb} set voltage. If this pin is connected to V_{S} , or no NTC is connected the system applies trickle charge.

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If pin RFSH is connected to ground by depressing the switch, the TEA1103x discharges the battery via an external transistor connected to pin RFSH. The discharge current is regulated with respect to the external (charge) sense resistor (R_{sense}). End-of-discharge is reached when the battery is discharged to 1 V per cell. Refreshing the battery can only be activated during charging of NiCd and NiMH batteries.

The inhibit mode has the main priority. This mode is activated when the V_{stb} input pin is connected to ground. Inhibit can be activated at any charge/discharge state, whereby the output control signals will be zero, all LEDs will be disabled and the charger timings will be set on hold. Table 1 gives an operational summary.

Table 1 Functionality of program pins

FUNCTION	FCT	NTC	RFSH	V_{stb}
Inhibit	X ⁽¹⁾	X ⁽¹⁾	X ⁽¹⁾	low
Refresh	not low ⁽²⁾	X ⁽¹⁾	low	not low
$\Delta T/\Delta t$ detection	floating	note 3	not low	not low
$\Delta T/\Delta t$ and voltage peak detection	high	note 3	not low	not low
Voltage peak detection	not low	note 4	not low	not low
Trickle charge at standby	not low	X ⁽¹⁾	not low	high
	not low	note 4	not low	not low
Voltage regulation at standby	not low	note 3	not low	floating ⁽⁵⁾

Notes

- Where X = don't care.
- Not low means floating or high.
- The NTC voltage has been to be less than 3.3 V, which indicates the presence of an NTC.
- The NTC voltage is outside the window for NTC detection.
- V_{stb} has to be floating or set to a battery regulating voltage in accordance with the specification.

Supply block

The supply block delivers the following outputs:

- A power-on reset pulse to reset all digital circuitry at battery insertion or supply switch-on. After a general reset the system will start fast charging the battery.
- A 4.25 V stabilized voltage source (V_S) is externally available. This source can be used to set the thermistor biasing, to initialize the programs, to supply the external circuitry for battery voltage based charge indication and to supply other external circuitry.
- A 4.25 V bias voltage (V_{SI}) is available for use for more indication LEDs. This output pin will be zero during the initialization period at start-up, thus avoiding any interference of the extra LEDs when initializing.

Charge control

The charge current is sensed via a low-ohmic resistor (R_{sense}), see Fig.4. A positive voltage is created across resistor R_b by means of a current source I_{ref} which is set by

R_{ref} in the event of fast charge and by an internal bias current source in the event of top-off and trickle charge (I_{IB}), see Fig.1. The positive node of R_b will be regulated to zero via error amplifier A1, which means that the voltage across R_b and R_{sense} will be the same. The fast charge current is defined by the following equation:

$$I_{\text{fast}} \times R_{\text{sense}} = R_b \times I_{\text{ref}} \quad (1)$$

The output of amplifier A1 is available at the loop stability pin LS, consequently the time constant of the current loop can be set. When V_{peak} (NiCd and NiMH) is applied, the current sensing for the battery voltage will be reduced, implying that the charge current will be regulated to zero during:

$$t_{\text{sense}} = 2^{10} \times \text{POD} \times t_{\text{osc}} \quad (2)$$

Actually battery voltage sensing takes place in the last oscillator cycle of this period.

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To avoid modulation on the output voltage, the top-off charge current is DC regulated, defined by the following equation:

$$I_{\text{top-off}} \times R_{\text{sense}} = R_b \times 3 \times 10^{-6} \quad (3)$$

where:

$$t_{\text{top-off}} = 2^{27} \times \text{TOD} \times t_{\text{osc}} \quad (4)$$

The top-off charge current will be approximately 0.15CA, which maximizes the charge in the battery under safe and slow charging conditions. The top-off charge period will be approximately one hour, so the battery will be extra charged with approximately 0.15 Q. In this way the battery is fully charged before the system switches over to standby.

When pin 1 (V_{stb}) is connected to V_S , or no NTC is connected the system compensates the (self) discharge of the battery by trickle charge. The trickle charge current will be pulsating, defined by the following equation:

$$I_{\text{trickle}} \times R_{\text{sense}} = R_b \times \frac{15}{16} \times 10^{-6} \quad (5)$$

During the non current periods at trickle charge the charge current is regulated to zero, so that the current for a load connected in series across the battery with the sense resistor will be supplied by the power supply and not by the battery.

If at pin 1 (V_{stb}) a reference voltage is set in accordance with the specification, and no NTC is connected the charge mode will switch over from current to voltage regulation after top-off. The reference regulating voltage can be adjusted to the battery characteristic by external resistors connected to pin V_{stb} .

This reference voltage has to be selected in such a way that it equals the rest voltage of the battery. By using voltage regulation, the battery will not be discharged at a load occurrence. If the V_{stb} input pin is floating, the TEA1103x will apply voltage regulation at 1.325 V during the standby mode (NiCd and NiMH). The current during voltage regulation is limited to 0.5CA. If the battery charge current is maximized to 0.5CA for more than 2 hours charging will be stopped. Moreover, if the temperature exceeds T_{max} , charging will be stopped completely. As voltage regulation is referred to one cell, the voltage on the V_{bat} pin must be the battery voltage divided by the number of cells (NiCd and NiMH).

When charging, the standby mode can only be entered after a certain period of time depending on time-out. To support full test of the TEA1103x at application, the standby mode is also entered when $V_{\text{bat}} < V_{\text{bat(l)}}$ at top-off.

Timer

The timing of the circuit is controlled by the oscillator frequency.

The timer block defines the maximum charging time by 'time-out'. At a fixed oscillator frequency, the time-out time can be adapted by the Programmable Time-out Divider (PTD) using the following equation.

$$t_{\text{time-out}} = 2^{26} \times \text{POD} \times \text{PTD} \times t_{\text{osc}} \quad (6)$$

The time-out timer is put on hold by low voltage, temperature protection and during the inhibit mode. The Programmable Oscillator Divider (POD) enables the oscillator frequency to be increased without affecting the sampling time and time-out. Raising the oscillator frequency will reduce the size of the inductive components that are used.

At fast charging, after battery insertion, after refresh or supply interruption, the full detector will be disabled for a period of time to allow a proper start with flat or inverse polarized batteries. This hold-off period is disabled at fast charging by raising pin V_{stb} to above ± 5 V (once).

So for test options it is possible to slip the hold-off period. The hold-off time is defined by the following equation:

$$t_{\text{hold-off}} = 2^{-5} \times t_{\text{time-out}} \quad (7)$$

Table 2 gives an overview of the settings of timing and discharge/charge currents.

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Table 2 Timing and current formulae

SYMBOL	DESCRIPTION	FORMULAE
t_{osc}	timing	see Fig.3
$T_{sampling} (\Delta T/\Delta t)$	NTC voltage sampling frequency	$2^{17} \times POD \times PSD \times t_{osc}$
$T_{sampling} (V_{peak})$	battery voltage sampling frequency	$2^{16} \times POD \times t_{osc}$
$t_{top-off}$		$2^{27} \times POD \times t_{osc}$
$t_{time-out}$		$2^{26} \times POD \times PTD \times t_{osc}$
$t_{hold-off}$		$2^{-5} \times t_{time-out}$
t_{LED}	inhibit or protection	$2^{14} \times POD \times t_{osc}$
t_{sense}		$2^{10} \times POD \times t_{osc}$
t_{switch}		$2^{21} \times POD \times PTD \times t_{osc}$
I_{fast}	charge/discharge currents	$\frac{R_b}{R_{sense}} \times \frac{V_{ref}}{R_{ref}}$
$I_{top-off}$		$\frac{R_b}{R_{sense}} \times 3 \times 10^{-6}$
$I_{trickle}$		$\frac{R_b}{R_{sense}} \times \frac{15}{16} \times 10^{-6}$
$I_{load-max}$		$\frac{R_b}{R_{sense}} \times 40 \times 10^{-6}$
I_{RFSH}		$\frac{100 \text{ mV}}{R_{sense}}$

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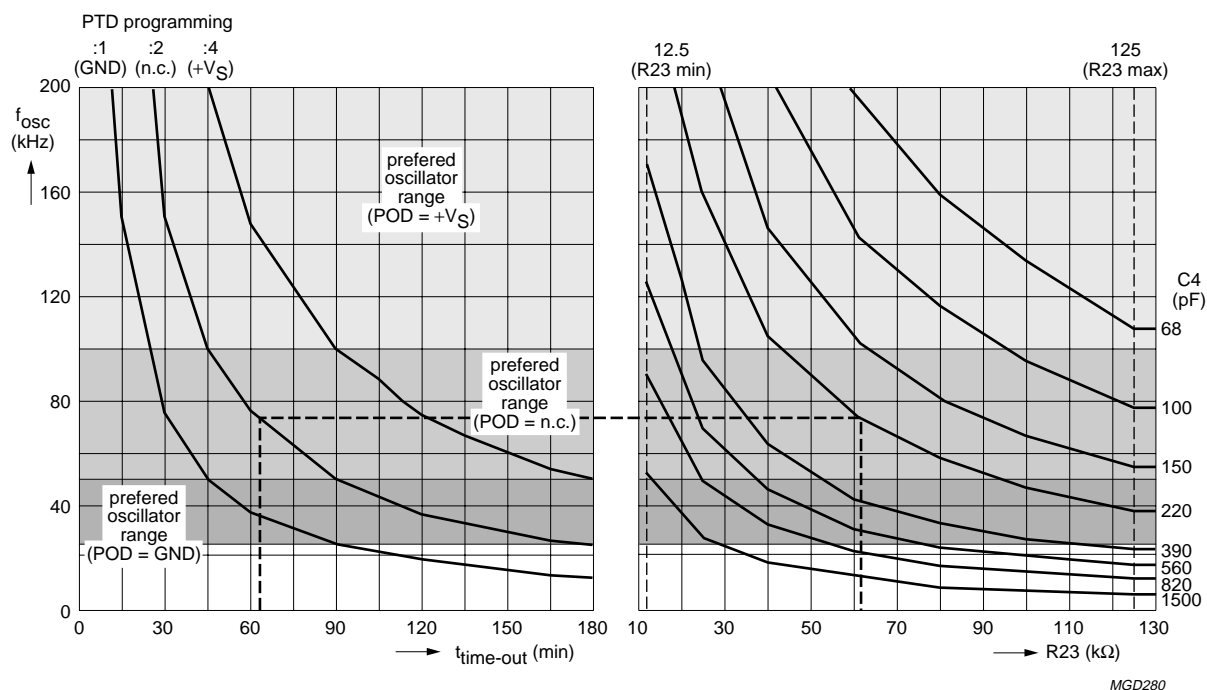


Fig.3 t_{time-out} as a function of R23 and PTD with C4 as parameter.

LED indication

With few external components, indication LEDs can be connected to the program pins and the LED pin of the TEA1103x. These program pins change their function from an input to an output pin after a short initialization time at system switch-on or battery insertion. Output pin V_{SI} enables the external LEDs to be driven and avoids interaction with the programming of the dividers during the initialization period.

The applied LEDs indicate:

- Protection
- Refresh
- Fast charge
- 100%
- No-battery.

The LED output pin can also indicate the charging state by one single LED. The indication LED can be connected directly to the LED output. This single LED indicates:

- Fast charge (LED on)
- 100% or refresh (LED off)
- Protection or inhibit (LED floating).

The refresh can be indicated by an extra LED connected to pin 4 (PSD). A buzzer can also be driven from the TEA1103x to indicate battery insertion end of refresh or full battery.

AD/DA converter

When battery full is determined by peak voltage detection, the V_{bat} voltage is sampled at a rate given by the following equation:

$$t_{\text{sampling}}(V_{\text{peak}}) = 2^{16} \times \text{POD} \times t_{\text{osc}} \quad (8)$$

The analog value of a V_{bat} sample is then digitized and stored in a register. On the following sample, the digitized value is converted back to the analog value of V_{bat} and compared with the 'new' V_{bat} sample.

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At an increase of the battery voltage the 14-bit Analog-to-Digital Converter (ADC) is refreshed with this new value. Therefore, the digitized value always represents the maximum battery voltage. A decreased V_{bat} voltage is not stored, but is compared to the stored value.

Full is detected when the voltage decrease of V_{bat} is $\frac{1}{4}\%$ of the stored peak battery value. To avoid interference due to the resistance of the battery contacts during battery voltage sensing, the charge current is regulated to zero during $t = 2^{10} \times POD \times t_{osc}$, via the regulation pins AO and PWM. At the last period, the V_{bat} voltage is sensed and stored in a sample-and-hold circuit. This approach ensures very accurate detection of the battery full condition (minus $\frac{1}{4}\%$).

When battery full is determined by $\Delta T/\Delta t$, the voltage on the NTC pin is used as the input voltage to the AD/DA converter. The sampling time at $\Delta T/\Delta t$ sensing is given by the following equation:

$$t_{\text{sampling}} \left(\frac{\Delta T}{\Delta t} \right) = 2^{17} \times POD \times PSD \times t_{osc} \quad (9)$$

After this initialized sample time the new temperature voltage is compared to the preceding AD/DA voltage and the AD/DA is refreshed with this new value. A certain increase of the temperature is detected as full battery, depending on the initialization settings. The decision of full detection by $\Delta T/\Delta t$ or V_{peak} is digitally filtered thus avoiding false battery full detection.

Output drivers

The charge current regulation signal is available at two output pins, AO and PWM.

ANALOG OUTPUT

The analog control voltage output at pin 18 (AO) can be used to drive an opto-coupler in mains separated applications when an external resistor is connected between AO and the opto-coupler. The maximum current through the opto-coupler diode is 2 mA. The voltage gain of amplifier A2 is typical 11 dB (times 3.5). The DC voltage transfer is given by the following equation:

$$V_{AO} = 3.5 \times (V_{LS} - 1.35).$$

The AO output can be used for:

- Linear (DC) applications
- Not mains isolated SMPS with a separate controller
- Mains isolated SMPS, controlled by an opto-coupler.

PULSE WIDTH MODULATOR (PWM)

The LS voltage is compared internally with the oscillator voltage to deliver a pulse width modulated output at PWM (pin 15) to drive an output switching device in a SMPS converter application via a driver stage. The PWM output is latched to prevent multi-pulsing. The maximum duty factor is internally fixed to 79% (typ.). The PWM output can be used for synchronization and duty factor control of a primary SMPS via a pulse transformer.

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LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134); note 1

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Voltages						
V_P	positive supply voltage		-0.5	—	+11.5	V
V_{oLED}	output voltage at pin 5		-0.5	—	+15	V
V_n	voltage at pins PWM, LS and NTC		-0.5	—	+ V_S	V
V_{IB}	voltage at pin 2		-0.5	—	+1.0	V
Currents						
I_{VS}	current at pin 16		-3	—	+0.01	mA
I_{Vsl}	current at pin 13		-1	—	+0.3	mA
I_{oLED}	output current at pin 5		—	—	12	mA
I_{AO}	output current at pin 18		-10	—	+0.05	mA
I_{oPWM}	output current at pin 15		-15	—	+14	mA
I_{Rref}	current at pin 20		-1	—	+0.01	mA
I_P	positive supply current	$T_j < 100\text{ °C}$	—	—	30	mA
$I_{P(stb)}$	supply standby current	$V_P = 4\text{ V}$	—	35	45	μA
Dissipation						
P_{tot}	total power dissipation	$T_{amb} = 85\text{ °C}$	—	—	1.2	W
	SOT146-1		—	—	0.6	W
	SOT163-1		—	—	0.45	W
	SOT339-1		—	—		
Temperatures						
T_{amb}	operating ambient temperature		-20	—	+85	°C
T_j	junction temperature		—	—	150	°C
T_{stg}	storage temperature		-55	—	+150	°C

Note

1. All voltages are measured with respect to ground; positive currents flow into the IC; all pins not mentioned in the voltage list are not allowed to be voltage driven. The voltage ratings are valid provided that other ratings are not violated; current ratings are valid provided that the power rating is not violated.

QUALITY SPECIFICATION

In accordance with the general quality specification for integrated circuits: "SNW-FQ-611E".

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CHARACTERISTICS

$V_P = 10\text{ V}$; $T_{\text{amb}} = 25\text{ °C}$; $R_{\text{ref}} = 62\text{ k}\Omega$; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Supplies; pins V_P, V_S, R_{ref} and V_{sl}						
V_P	supply voltage		5.5	–	11.5	V
I_P	supply current	outputs off; $V_P = 11.5\text{ V}$	–	4	6	mA
I_{stb}	standby current	$V_P = 4\text{ V}$	–	35	45	μA
V_{clamp}	clamping voltage (pin 12)	$I_{\text{clamp}} = 30\text{ mA}$	11.5	–	12.8	V
V_{start}	start voltage		6.1	6.4	6.7	V
V_{LSP}	low supply protection level		5.1	5.3	5.5	V
V_S	source voltage (stabilized)	$I_S = 2\text{ mA}$	4.14	4.25	4.36	V
V_{SL}	LED source voltage	$I_{\text{LED}} = 50\text{ }\mu\text{A}$	4.05	4.25	4.45	V
V_{ref}	reference voltage	$I_{\text{ref}} = 20\text{ }\mu\text{A}$; $V_P = 10\text{ V}$	1.21	1.25	1.29	V
$\text{TC}_{V_{\text{ref}}}$	temperature coefficient of the reference voltage	$T_{\text{amb}} = 0\text{ to }45\text{ °C}$; $I_{\text{ref}} = 20\text{ }\mu\text{A}$; $V_{\text{ref}} = 1.25\text{ V}$	0	± 60	± 120	ppm/K
$\Delta V_{\text{ref}}/\Delta V_P$	power supply rejection ratio of the reference voltage	$f = 100\text{ Hz}$; $V_P = 8\text{ V}$; $\Delta V_P = 2\text{ V (p-p)}$	–46	–	–	dB
ΔV_{ref}	load rejection of source voltage	$\Delta I_S = 20\text{ mA}$; $V_P = 10\text{ V}$	–	–	5	mV
I_{Rref}	current range of reference resistor		10	–	100	μA
Charge current regulation; pins I_B and R_{ref}						
I_B/I_{ref}	fast charge ratio	$V_{IB} = 0$ $I_{\text{ref}} = 10\text{ }\mu\text{A}$ $I_{\text{ref}} = 100\text{ }\mu\text{A}$	0.93 0.93	1.03 1.0	1.13 1.07	
V_{thIB}	threshold voltage at pin I_B	$T_{\text{amb}} = 25\text{ °C}$	–2	–	+2	mV
		$T_{\text{amb}} = 0\text{ to }45\text{ °C}$	–3	–	+3	mV
I_{IB}	charge current	top-off mode; $V_{IB} = 0$	2.6	3.2	3.8	μA
$I_{B(\text{max})}$	maximum charge current	voltage regulation full NiCd/NiMH battery; $V_{IB} = 0$	9	10.5	12	μA
$I_{B(\text{Lmax})}$	maximum load current	open battery; $V_{IB} = 0$	34	42	50	μA
$I_{B(\text{LI})}$	input leakage current	currentless mode	–	–	170	nA
Refresh; pin RFSH						
V_{Rsense}	sense resistor voltage	$I_{\text{refresh}} = \frac{V_{IB}}{R_{\text{sense}}}$; refresh mode; $I_{\text{refresh}} = 18\text{ mA}$	75	100	125	mV
V_{RFSH}	refresh voltage for programming start of refresh	NiCd/NiMH	0	–	250	mV
V_{bat}	voltage at pin V_{bat} for detecting end of refresh	NiCd/NiMH	0.96	1.0	1.04	V

Fast charge ICs for NiCd and NiMH batteries

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$I_{\text{source(max)}}$	maximum source current	$V_{\text{IB}} = 75 \text{ mV}$; $V_{\text{P}} = 10 \text{ V}$ $V_{\text{RFSH}} = 2.7 \text{ V}$; $T_{\text{amb}} = 25 \text{ }^{\circ}\text{C}$	1.4	2	2.6	mA
$V_{\text{RFSH(max)}}$	maximum refresh voltage	$I_{\text{RFSH}} = 1 \text{ mA}$	2.7	–	–	V
$V_{\text{RFSH(off)}}$	voltage at pin RFSH when refresh is off		700	770	840	mV
Temperature related inputs; pins NTC and MTV						
V_{NTCh}	input voltage at pin NTC for detecting high temperature	pin MTV open-circuit	0.9	1	1.1	V
		MTV setting	0.95MTV	MTV	1.05MTV	V
$V_{\text{NTCh(hy)}}$	hysteresis of V_{NTCh}		–	80	–	mV
V_{NTCl}	input voltage at pin NTC, detecting low temperature		2.7	2.8	2.9	V
$V_{\text{NTCl(hy)}}$	hysteresis of V_{NTCl}		–	75	–	mV
$V_{\text{NTC(co)}}$	input voltage at pin NTC for detecting temperature cut-off		0.7MTV	0.75MTV	0.8MTV	V
$V_{\text{NTC(bat)}}$	maximum input voltage at pin NTC for detecting battery with NTC		3.22	3.3	3.38	V
I_{NTC}	input current at pin NTC	$V_{\text{NTC}} = 2 \text{ V}$	–5	–	+5	μA
V_{MTV}	voltage level at pin MTV	default (open-circuit)	0.95	1	1.05	V
			0.5	–	2.5	V
$\Delta V_{\text{NTC}}/V_{\text{NTC}}$	$\Delta T/\Delta t$ detection level	$V_{\text{NTC}} = 2 \text{ V}$; $T_{\text{J}} = 0 \text{ to } 50 \text{ }^{\circ}\text{C}$	–	–0.25	–	%
Voltage regulation						
V_{reg}	regulation voltage	NiCd and NiMH; pin V_{stb} open-circuit	1.34	1.325	1.40	V
		NiCd and NiMH; $V_{\text{stb}} = 1.5 \text{ V}$	$0.99V_{\text{stb}}$	V_{stb}	$1.01V_{\text{stb}}$	V
		open battery	1.86	1.9	1.94	V
$\text{TC}_{V_{\text{reg}}}$	temperature coefficient of regulation voltage	$V_{\text{reg}} = 1.325 \text{ V}$; $T_{\text{amb}} = 0 \text{ to } 45 \text{ }^{\circ}\text{C}$	0	± 60	± 120	ppm/K
g_{m}	transconductance of amplifier A3	$V_{\text{bat}} = 1.9 \text{ V}$; no battery mode	–	2.0	–	mA/V
Program pin V_{stb}						
V_{stb}	open voltage at pin V_{stb}		1.30	1.325	1.35	V
$V_{\text{stb(im)}}$	voltage at pin V_{stb} for programming inhibit mode		0	–	0.8	V
$V_{\text{stb(st)}}$	voltage at pin V_{stb} for programming voltage regulation at standby	NiCd and NiMH	1.0	–	2.2	V
$V_{\text{stb(tc)}}$	voltage at pin V_{stb} for programming trickle charge at standby	NiCd and NiMH	2.6	–	V_{S}	V

Fast charge ICs for NiCd and NiMH batteries

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Program pins; PSD, POD and PTD						
$V_{4,6,7}$	voltage level at pins PSD, POD or PTD	default (open-circuit)	1.9	2.1	2.3	V
$V_{4,6,7(1)}$	voltage level at pins PSD, POD or PTD for programming the divider = 1		0	–	1.2	V
$V_{4,6,7(2)}$	voltage level at pins PSD, POD or PTD for programming the divider = 2		1.6	–	2.5	V
$V_{4,6,7(4)}$	voltage level at pins PSD, POD or PTD for programming the divider = 4		3.1	–	V_S	V
$I_{PODsink}$	protection current for multi-LED indication	$V_{POD} = 1.5\text{ V}$	8	10	12	mA
$I_{PTDsink}$	full battery current for multi-LED indication	$V_{PTD} = 1.5\text{ V}$	8	10	12	mA
$I_{PSDsink}$	refresh current for multi-LED indication	$V_{PSD} = 1.5\text{ V}$	8	10	12	mA
I_{LI}	input leakage current	$V_{POD} = 4.25\text{ V};$ $V_{PTD} = 4.25\text{ V};$ $V_{PSD} = 4.25\text{ V}$	0	–	50	μA
Program pin FCT						
$V_{FCT(or)}$	voltage level for programming $\Delta T/\Delta t$ or V_{peak} as fast charge termination	NiCd and NiMH	0.0	–	3.3	V
$V_{FCT(and)}$	voltage level for programming $\Delta T/\Delta t$ and V_{peak} as fast charge termination	NiCd and NiMH	3.7	–	V_S	V
V_{FCT}	voltage level at pin FCT	default (open-circuit)	2.3	2.6	2.9	V
Program pin LED						
$V_{LED(m)}$	output voltage level for programming multi-LED indication		0	–	2.5	V
$V_{LED(s)}$	output voltage level for programming single LED indication		3.1	–	V_P	V
$I_{sink(max)}$	maximum sink current	$V_{LED} = 1.5\text{ V}$	8	10	12	mA
$I_{LI(LED)}$	input leakage current	$V_{LED} = 10\text{ V}$	0	–	70	μA
		$V_{LED} = 0.6\text{ V}$	0	–	5	μA
$V_{o(max)}$	maximum output voltage		–	–	15	V

Fast charge ICs for NiCd and NiMH batteries

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
Output drivers; AO, LS and PWM						
$I_{AO(source)}$	analog output source current	$V_{AO} = 3\text{ V (p-p)}$; $V_{LS} = 2.8\text{ V}$	-9	-	0	mA
$I_{AO(sink)}$	analog output sink current	$V_{AO} = 3\text{ V (p-p)}$; $V_{LS} = 1.2\text{ V}$	50	-	-	μA
g_{m1}	transconductance of amplifier A1	$V_{IB} = 50\text{ mV}$	-	250	-	$\mu\text{A/V}$
$G_{v1,2}$	voltage gain of amplifiers A1 and A2	$V_{AO} = 3\text{ V (p-p)}$	-	72	-	dB
G_{v2}	voltage gain of amplifier A2	$V_{AO} = 2\text{ V (p-p)}$	-	11	-	dB
$I_{LS(source)}$	maximum source current (pin LS)	$V_{LS} = 2.25\text{ V}$	-25	-21	-16	μA
$I_{LS(sink)}$	maximum sink current (pin LS)	$V_{LS} = 2.25\text{ V}$	16	21	25	μA
$I_{OH(PWM)}$	HIGH level output current	$V_{PWM} = 3\text{ V}$	-19	-15	-11	mA
$I_{OL(PWM)}$	LOW level output current	$V_{PWM} = 0.7\text{ V}$	10	14	18	mA
δ_{PWM}	maximum duty factor		-	79	-	%
Battery monitor; V_{bat}						
I_{Vbat}	battery monitor input current	$V_{bat} = 1.85\text{ V}$	-	1	-	nA
V_{bat}	voltage range of V_{peak} detection		0.3	-	2	V
$\Delta V_{bat}/V_{bat}$	V_{peak} detection level with respect to top level	$V_{bat} = 1.85\text{ V}$; $T_j = 0\text{ to }50\text{ }^\circ\text{C}$	-	-0.25	-	%
ΔV_{bat}	voltage resolution for V_{peak}		-	0.6	-	mV
Protections; V_{bat}						
$V_{bat(l)}$	maximum voltage at pin V_{bat} for detecting low battery voltage		0.25	0.30	0.35	V
Oscillator; pin OSC						
$V_{osc(H)}$	HIGH level oscillator switching voltage		-	2.5	-	V
$V_{osc(L)}$	LOW level oscillator switching voltage		-	1.5	-	V
$f_{osc(min)}$	minimum oscillator frequency	$R_{ref} = 125\text{ k}\Omega$; $C_{osc} = 400\text{ pF}$	20.9	23	25.1	kHz
$f_{osc(max)}$	maximum oscillator frequency	$R_{ref} = 12.5\text{ k}\Omega$; $C_{osc} = 400\text{ pF}$	158	174	190	kHz

Fast charge ICs for NiCd and NiMH batteries

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APPLICATION INFORMATION

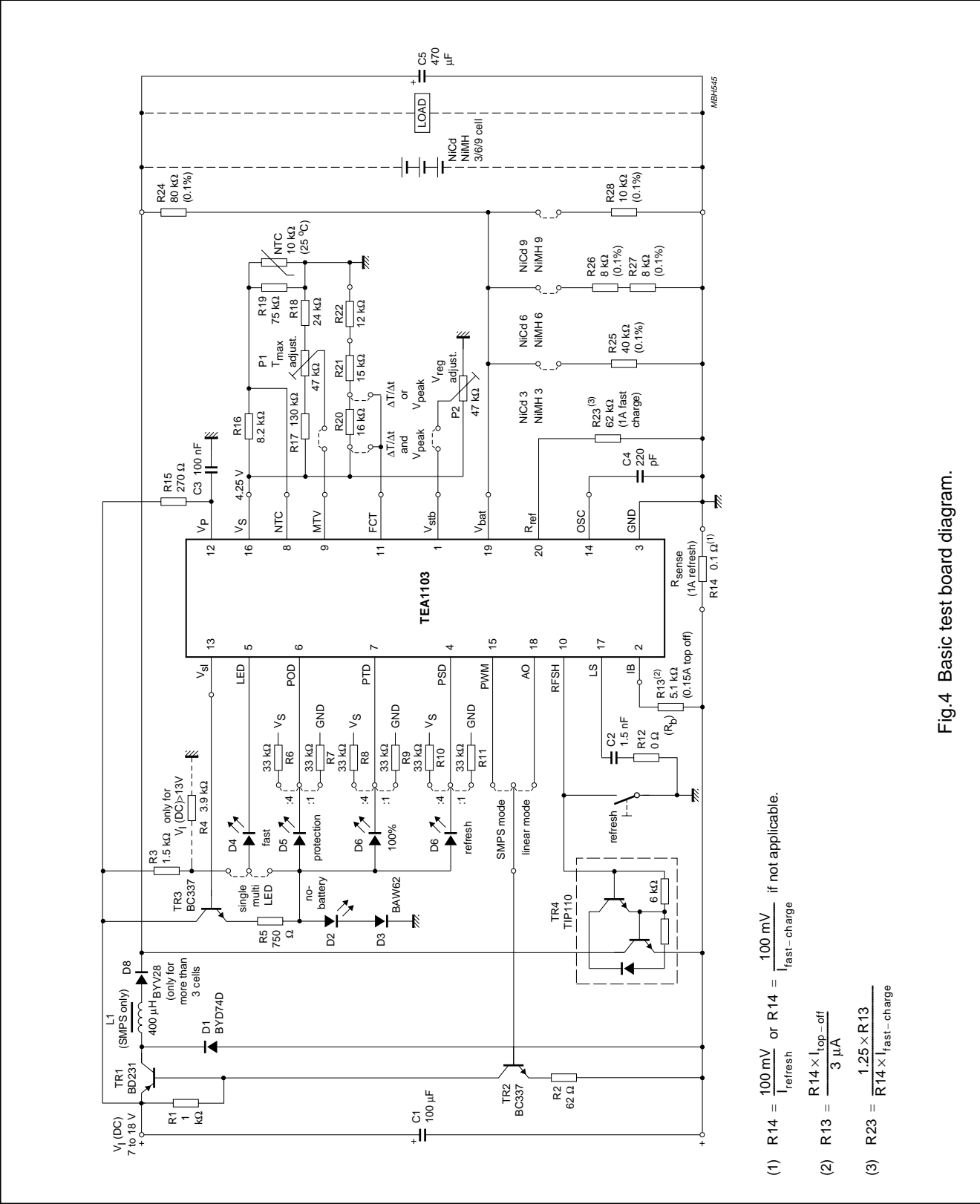


Fig.4 Basic test board diagram.

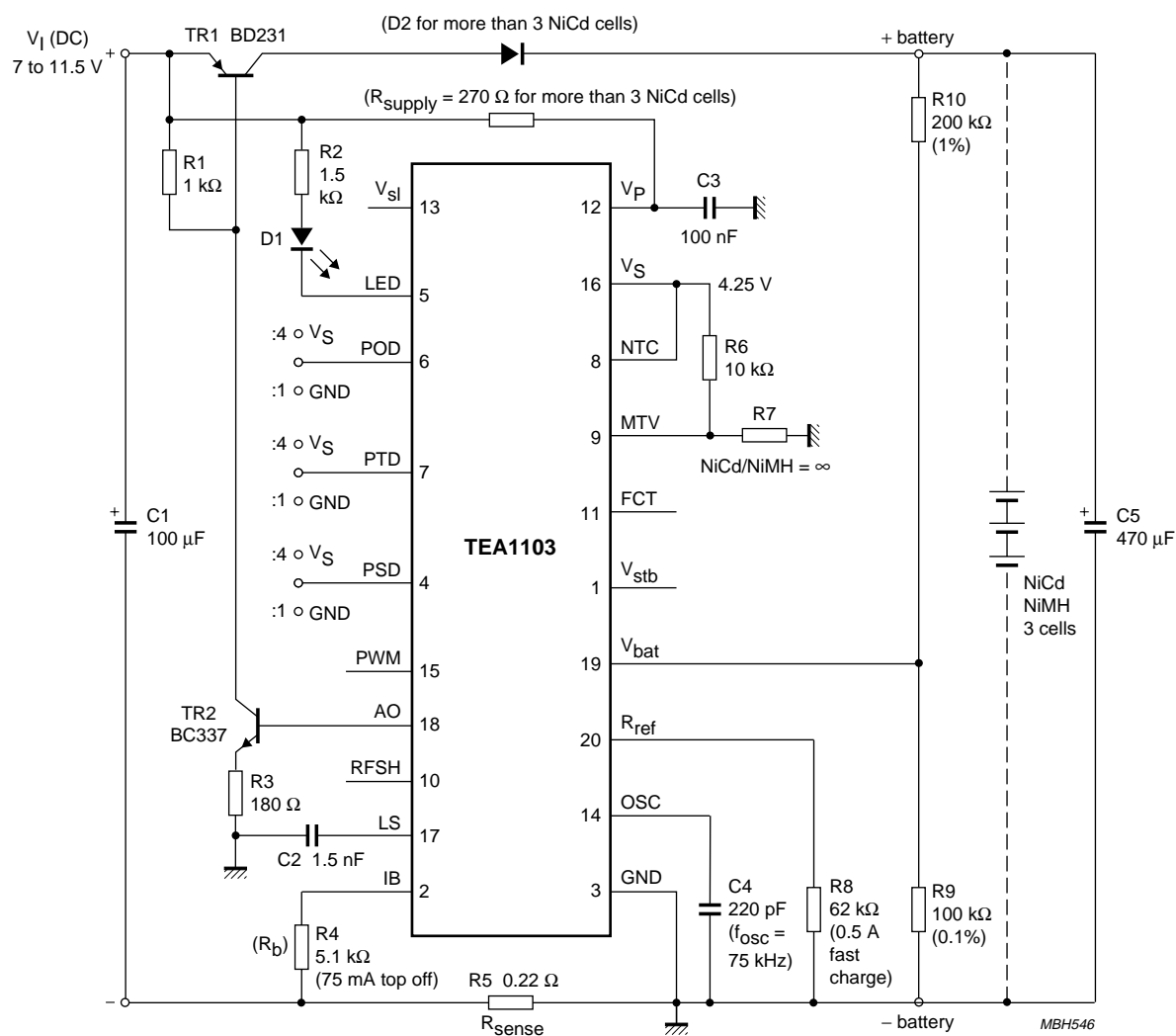
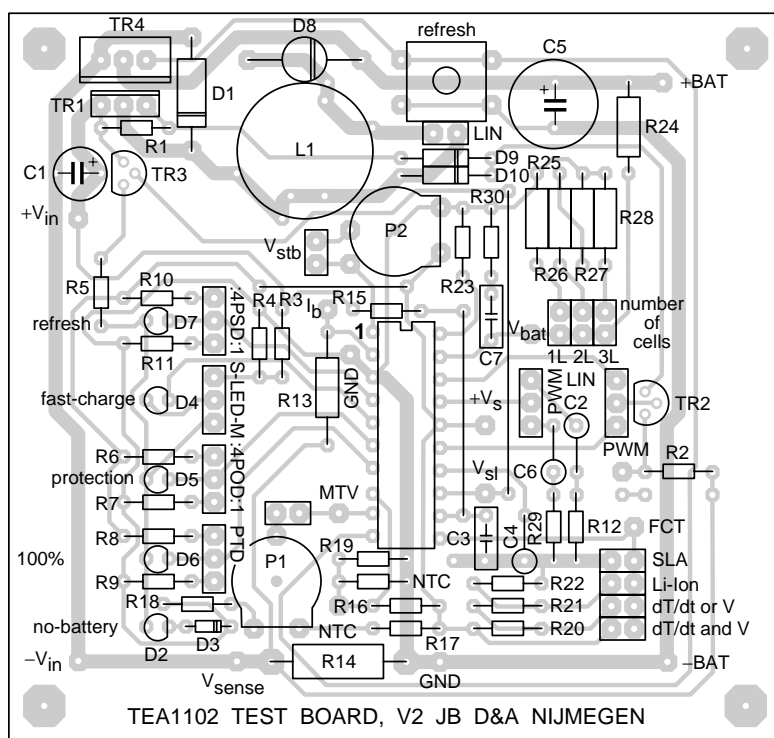
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Fig.5 Linear application diagram.

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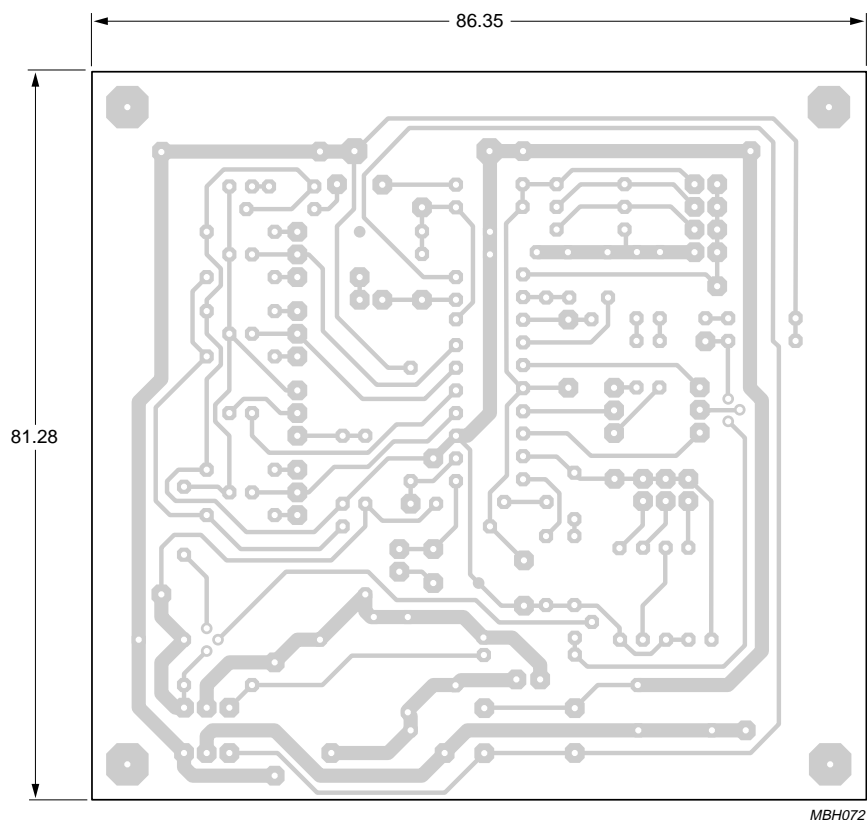
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MBH073

This test board (designed for the TEA1102x) can also be used for the TEA1103x.

Fig.6 Component side of printed-circuit board (test board).

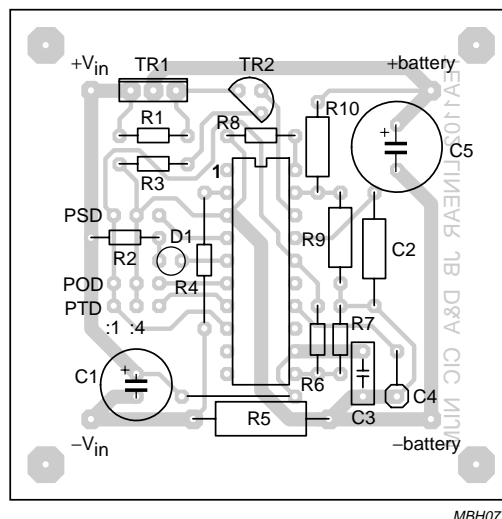
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Dimensions in mm.

Fig.7 Track side of printed-circuit board (test board).

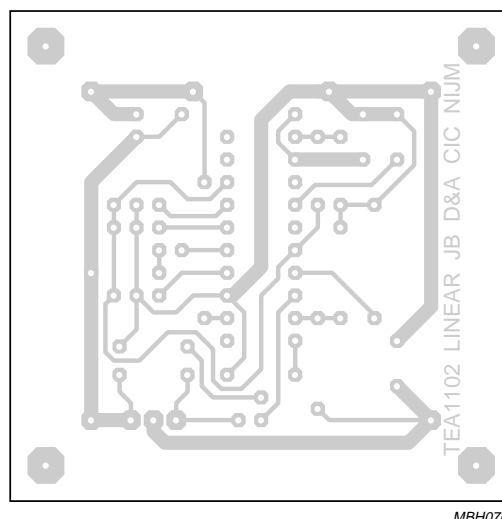
Fast charge ICs for NiCd and NiMH batteries

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TEA1103TS



This printed-circuit board (designed for the TEA1102x) can also be used for the TEA1103x.

Fig.8 Component side of printed-circuit board (linear application).



This printed-circuit board (designed for the TEA1102x) can also be used for the TEA1103x.

Fig.9 Track side of printed-circuit board (linear application).

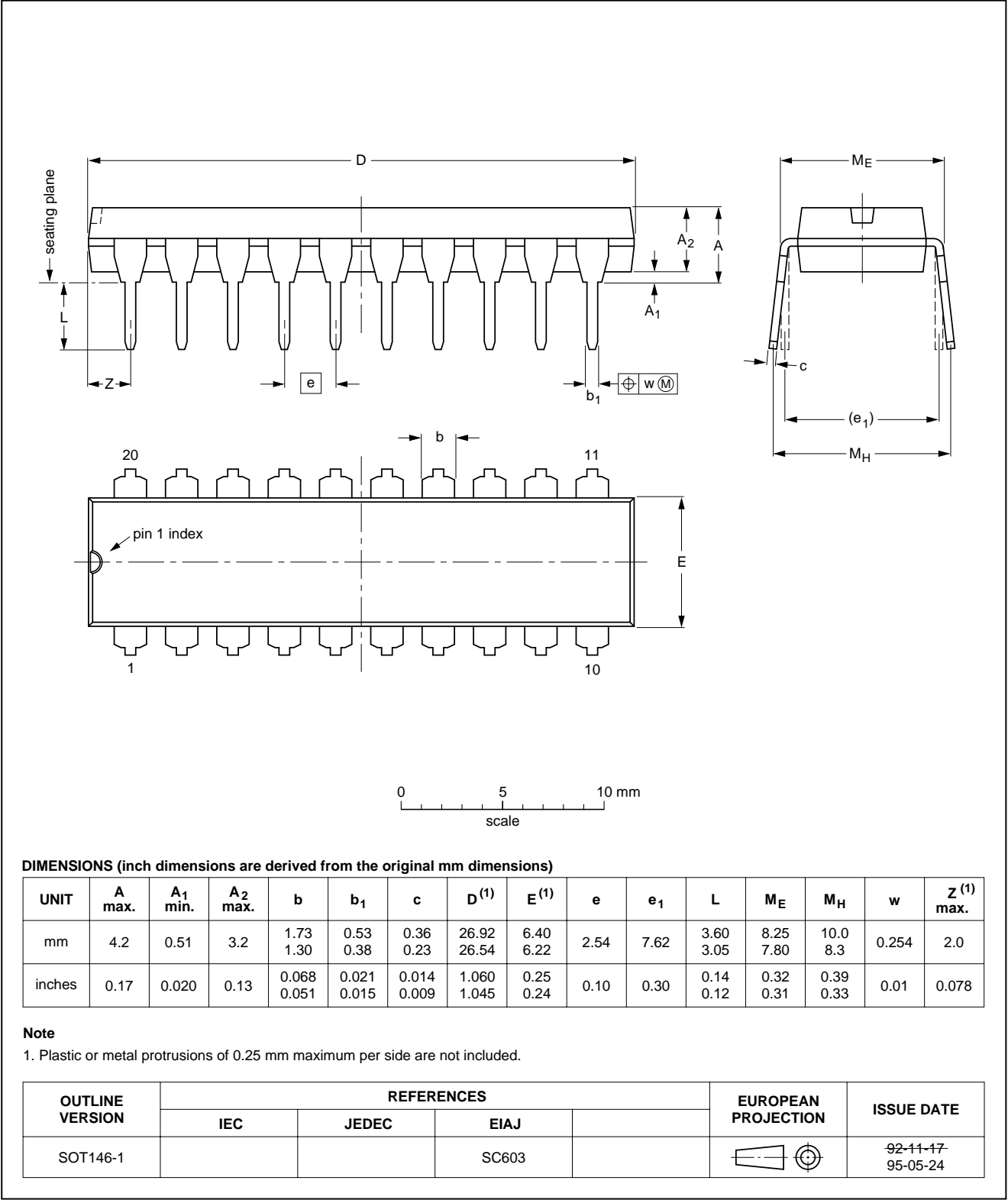
Fast charge ICs for NiCd and NiMH
batteries

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PACKAGE OUTLINES

DIP20: plastic dual in-line package; 20 leads (300 mil)

SOT146-1

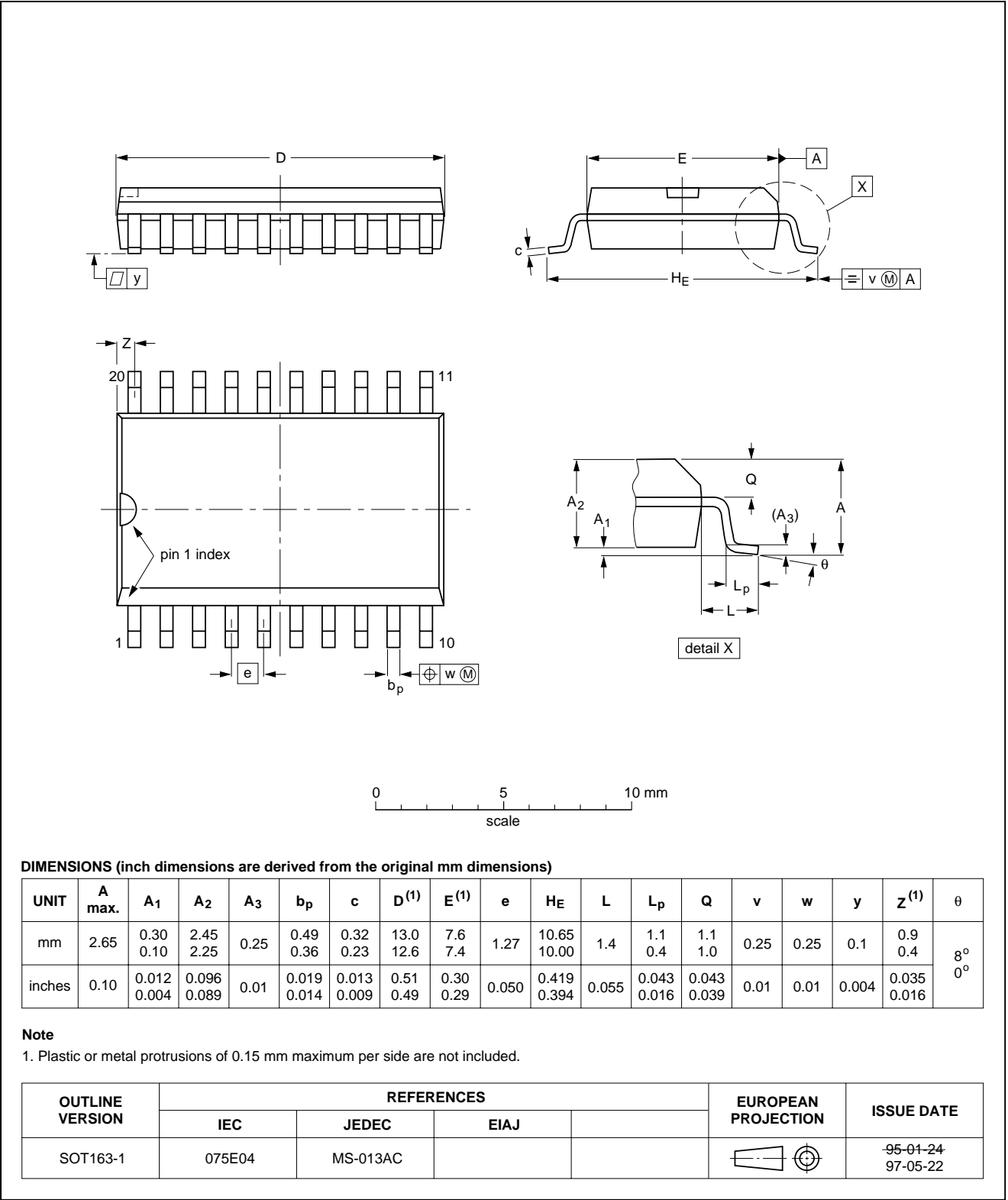


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SO20: plastic small outline package; 20 leads; body width 7.5 mm

SOT163-1

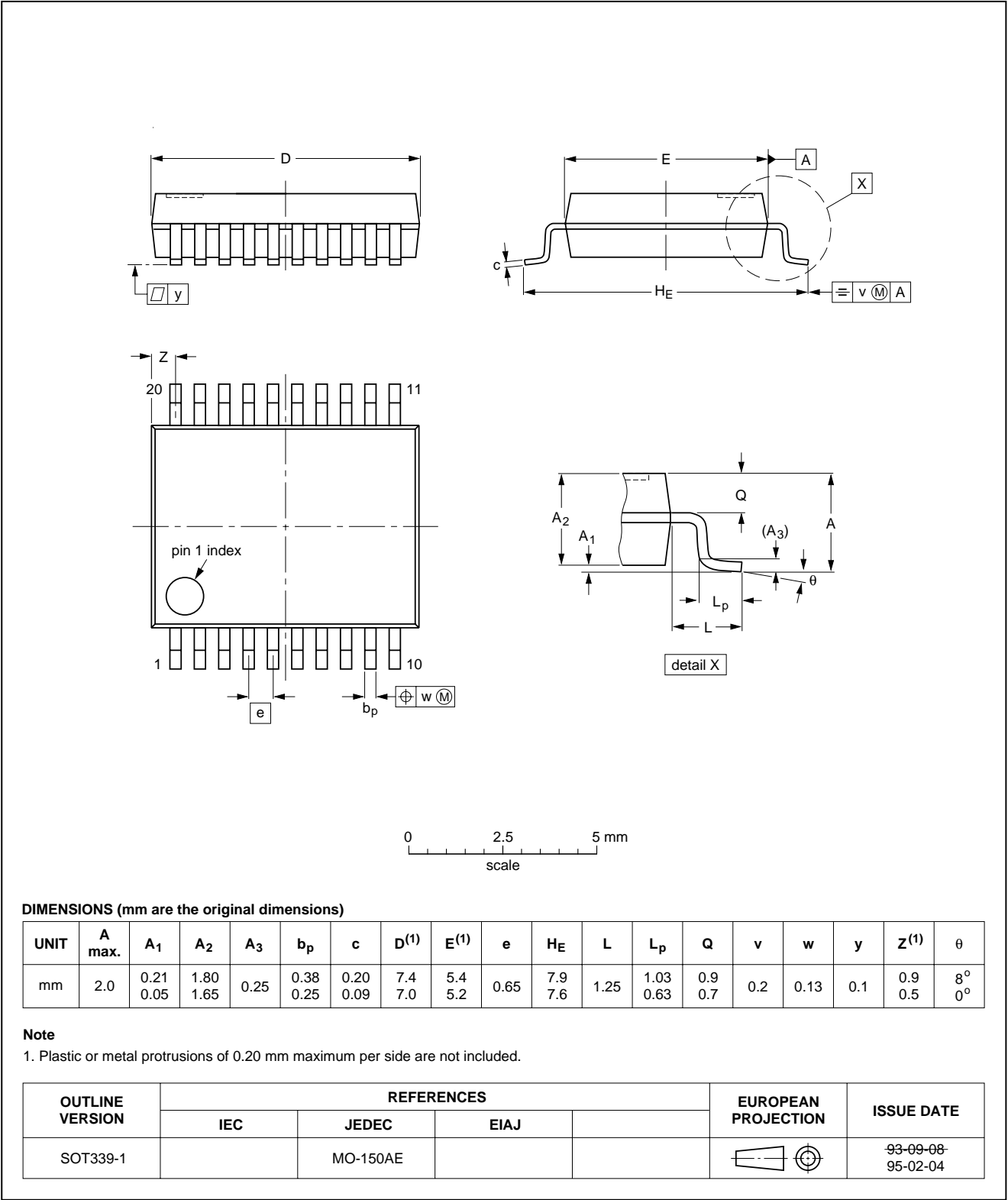


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SSOP20: plastic shrink small outline package; 20 leads; body width 5.3 mm

SOT339-1



Fast charge ICs for NiCd and NiMH batteries

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SOLDERING

Introduction

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"Data Handbook IC26; Integrated Circuit Packages"* (document order number 9398 652 90011).

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mount components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mount ICs, or for printed-circuit boards with high population densities. In these situations reflow soldering is often used.

Through-hole mount packages

SOLDERING BY DIPPING OR BY SOLDER WAVE

The maximum permissible temperature of the solder is 260 °C; solder at this temperature must not be in contact with the joints for more than 5 seconds. The total contact time of successive solder waves must not exceed 5 seconds.

The device may be mounted up to the seating plane, but the temperature of the plastic body must not exceed the specified maximum storage temperature ($T_{stg(max)}$). If the printed-circuit board has been pre-heated, forced cooling may be necessary immediately after soldering to keep the temperature within the permissible limit.

MANUAL SOLDERING

Apply the soldering iron (24 V or less) to the lead(s) of the package, either below the seating plane or not more than 2 mm above it. If the temperature of the soldering iron bit is less than 300 °C it may remain in contact for up to 10 seconds. If the bit temperature is between 300 and 400 °C, contact may be up to 5 seconds.

Surface mount packages

REFLOW SOLDERING

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several methods exist for reflowing; for example, infrared/convection heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 100 and 200 seconds depending on heating method.

Typical reflow peak temperatures range from 215 to 250 °C. The top-surface temperature of the packages should preferably be kept below 230 °C.

WAVE SOLDERING

Conventional single wave soldering is not recommended for surface mount devices (SMDs) or printed-circuit boards with a high component density, as solder bridging and non-wetting can present major problems.

To overcome these problems the double-wave soldering method was specifically developed.

If wave soldering is used the following conditions must be observed for optimal results:

- Use a double-wave soldering method comprising a turbulent wave with high upward pressure followed by a smooth laminar wave.
- For packages with leads on two sides and a pitch (e):
 - larger than or equal to 1.27 mm, the footprint longitudinal axis is **preferred** to be parallel to the transport direction of the printed-circuit board;
 - smaller than 1.27 mm, the footprint longitudinal axis **must** be parallel to the transport direction of the printed-circuit board.

The footprint must incorporate solder thieves at the downstream end.

- For packages with leads on four sides, the footprint must be placed at a 45° angle to the transport direction of the printed-circuit board. The footprint must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

MANUAL SOLDERING

Fix the component by first soldering two diagonally-opposite end leads. Use a low voltage (24 V or less) soldering iron applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C.

When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

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Suitability of IC packages for wave, reflow and dipping soldering methods

MOUNTING	PACKAGE	SOLDERING METHOD		
		WAVE	REFLOW ⁽¹⁾	DIPPING
Through-hole mount	DBS, DIP, HDIP, SDIP, SIL	suitable ⁽²⁾	–	suitable
Surface mount	HLQFP, HSQFP, HSOP, SMS	not suitable ⁽³⁾	suitable	–
	PLCC ⁽⁴⁾ , SO	suitable	suitable	–
	LQFP, QFP, TQFP	not recommended ⁽⁴⁾⁽⁵⁾	suitable	–
	SQFP	not suitable	suitable	–
	SSOP, TSSOP, VSO	not recommended ⁽⁶⁾	suitable	–

Notes

1. All surface mount (SMD) packages are moisture sensitive. Depending upon the moisture content, the maximum temperature (with respect to time) and body size of the package, there is a risk that internal or external package cracks may occur due to vaporization of the moisture in them (the so called popcorn effect). For details, refer to the Drypack information in the *"Data Handbook IC26; Integrated Circuit Packages; Section: Packing Methods"*.
2. For SDIP packages, the longitudinal axis must be parallel to the transport direction of the printed-circuit board.
3. These packages are not suitable for wave soldering as a solder joint between the printed-circuit board and heatsink (at bottom version) can not be achieved, and as solder may stick to the heatsink (on top version).
4. If wave soldering is considered, then the package must be placed at a 45° angle to the solder wave direction. The package footprint must incorporate solder thieves downstream and at the side corners.
5. Wave soldering is only suitable for LQFP, QFP and TQFP packages with a pitch (e) equal to or larger than 0.8 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.65 mm.
6. Wave soldering is only suitable for SSOP and TSSOP packages with a pitch (e) equal to or larger than 0.65 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.5 mm.

DEFINITIONS

Data sheet status	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
Limiting values	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
Application information	
Where application information is given, it is advisory and does not form part of the specification.	

LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.

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