

4-BRIDGE SERIAL INTERFACE MOTOR DRIVER

Check for Samples: [DRV8823](#)

FEATURES

- **PWM motor Driver with Four H-Bridges**
 - Drives Two Stepper Motors, One Stepper and Two DC Motors, or Four DC Motors
 - Up to 1.5-A Current Per Winding
 - Low On-Resistance
 - Programmable Maximum Winding Current
 - Three-Bit Winding Current Control Allows up to Eight Current Levels
 - Selectable Slow or Mixed Decay Modes
- **8-V to 32-V Operating Supply Voltage Range**
- **Internal Charge Pump for Gate Drive**
- **Built-in 3.3-V Reference**

- **Serial Digital Control Interface**
- **Fully Protected Against Undervoltage, Overtemperature, and Overcurrent**
- **Thermally Enhanced Surface Mount Package**

APPLICATIONS

- **Printers**
- **Scanners**
- **Office Automation Machines**
- **Gaming Machines**
- **Factory Automation**
- **Robotics**

DESCRIPTION/ORDERING INFORMATION

The DRV8823 provides an integrated motor driver solution for printers and other office automation equipment applications.

The motor driver circuit includes four H-bridge drivers. Each of the motor driver blocks employ N-channel power MOSFETs configured as an H-bridge to drive the motor windings.

A simple serial interface allows control of all functions of the motor driver with only a few digital signals. A low-power sleep function is also provided.

The motor drivers provide PWM current control capability. The current is programmable, based on an externally supplied reference voltage and an external current sense resistor. In addition, eight current levels (set via the serial interface) allow microstepping with bipolar stepper motors.

Internal shutdown functions are provided for over current protection, short circuit protection, under voltage lockout and overtemperature.

The DRV8823 is packaged in a 48 pin HTSSOP package (Eco-friendly : RoHS & no Sb/Br).

ORDERING INFORMATION⁽¹⁾

T _A	PACKAGE ⁽²⁾		ORDERABLE PART NUMBER	TOP-SIDE MARKING
–40°C to 85°C	PowerPad™ (HTSSOP) - DCA	Reel of 2000	DRV8823DCAR	DRV8823
		Tube of 40	DRV8823DCA	

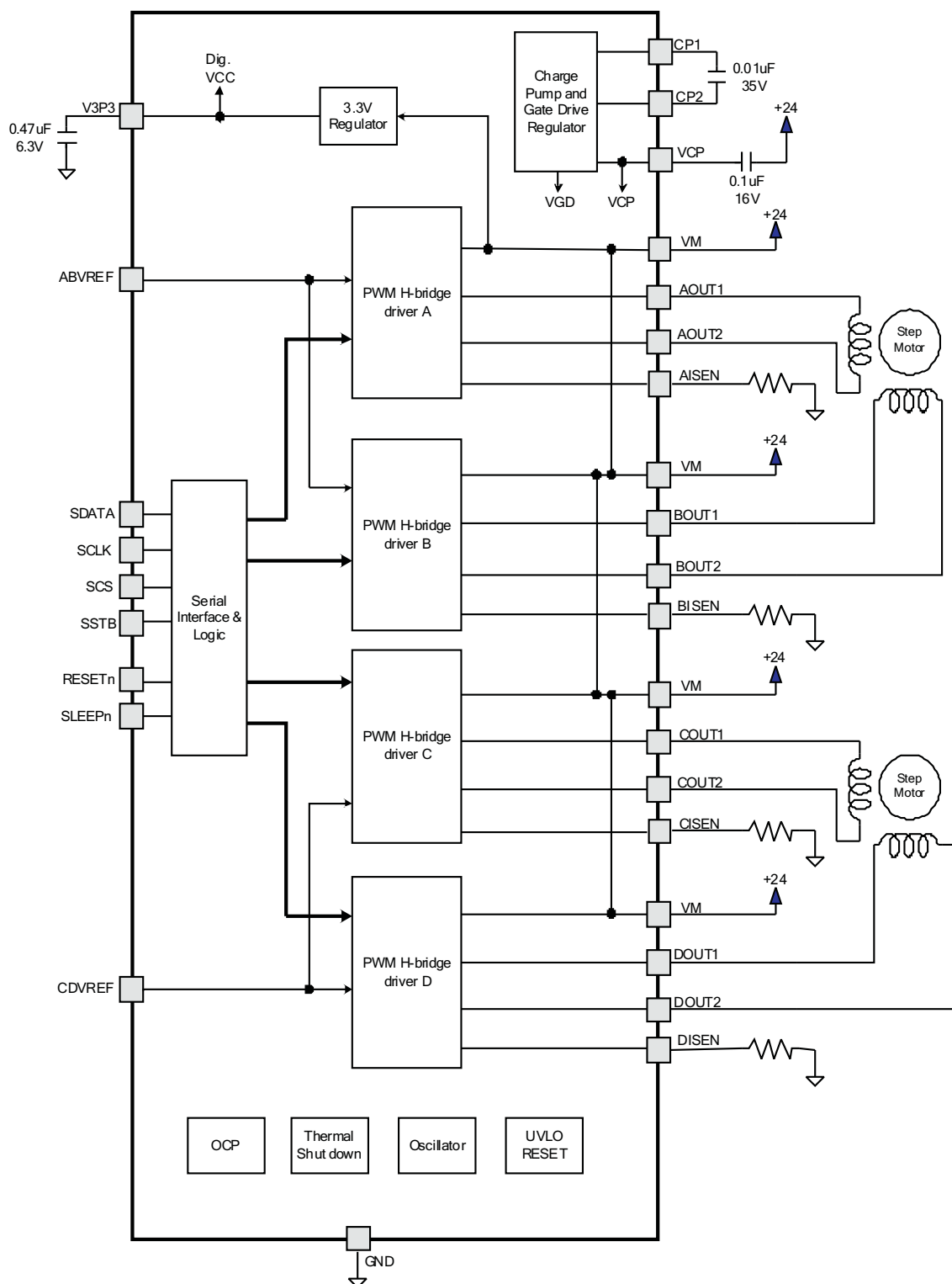
(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

(2) Package drawings, thermal data, and symbolization are available at www.ti.com/packaging.



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FUNCTIONAL BLOCK DIAGRAM

TERMINAL FUNCTIONS

NAME	NO.	I/O ⁽¹⁾	DESCRIPTION	EXTERNAL COMPONENTS OR CONNECTIONS
POWER AND GROUND				
VM (4 pins)	1, 2, 23, 24	-	Motor supply voltage (multiple pins)	Connect all VM pins together to motor supply voltage. Bypass to GND with several 0.1- μ F, 35-V ceramic capacitors.
V3P3	16	-	3.3 V regulator output	Bypass to GND with 0.47- μ F, 6.3-V ceramic capacitor.
GND	10 - 15, 34 - 39	-	Power ground (multiple pins)	Connect all PGND pins to GND and solder to copper heatsink areas.
CP1	7	IO	Charge pump flying capacitor	Connect a 0.01- μ F capacitor between CP1 and CP2
CP2	8	IO		
VCP	9	IO	Charge pump storage capacitor	Connect a 0.1- μ F, 16 V ceramic capacitor to V _M
MOTOR DRIVERS				
ABVREF	17	I	Bridge A & B current set reference voltage	Sets current trip threshold.
AOUT1	5	O	Bridge A output 1	Connect to first coil of bipolar stepper motor 1, or DC motor winding.
AOUT2	3	O	Bridge A output 2	
ISENA	4	-	Bridge A current sense	Connect to current sense resistor for bridge A.
BOUT1	48	O	Bridge B output 1	Connect to second coil of bipolar stepper motor 1, or DC motor winding.
BOUT2	46	O	Bridge B output 2	
ISENB	47	-	Bridge B current sense	Connect to current sense resistor for bridge B.
CDVREF	18	I	Bridge C & D current set reference voltage	Sets current trip threshold.
COUT1	27	O	Bridge C output 1	Connect to first coil of bipolar stepper motor 2, or DC motor winding.
COUT2	25	O	Bridge C output 2	
ISENC	26	-	Bridge C current sense	Connect to current sense resistor for bridge C.
DOUT1	22	O	Bridge D output 1	Connect to second coil of bipolar stepper motor 2, or DC motor winding.
DOUT2	20	O	Bridge D output 2	
ISEND	22	-	Bridge D current sense	Connect to current sense resistor for bridge D.
SERIAL INTERFACE				
SDATA	31	I	Serial data input	Data is clocked in on rising edge of SCLK.
SCLK	33	I	Serial input clock	Logic high enables serial data to be clocked in.
SCS	45	I	Serial chip select	Logic high latches serial data.
SSTB	30	I	Serial data strobe	Active low resets serial interface and disables outputs.
RESETn	43	I	Reset input	Active low input disables outputs and charge pump.
SLEEPn	42	I	Sleep input	
TEST PINS				
TEST	19, 28, 29, 32	I	Test inputs	Do not connect these pins - used for factory test only.

(1) Directions: I = input, O = output, OZ = 3-state output, OD = open-drain output, IO = input/output, PU = internal pullup

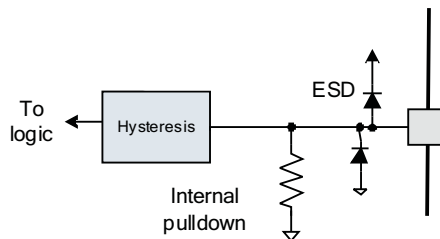
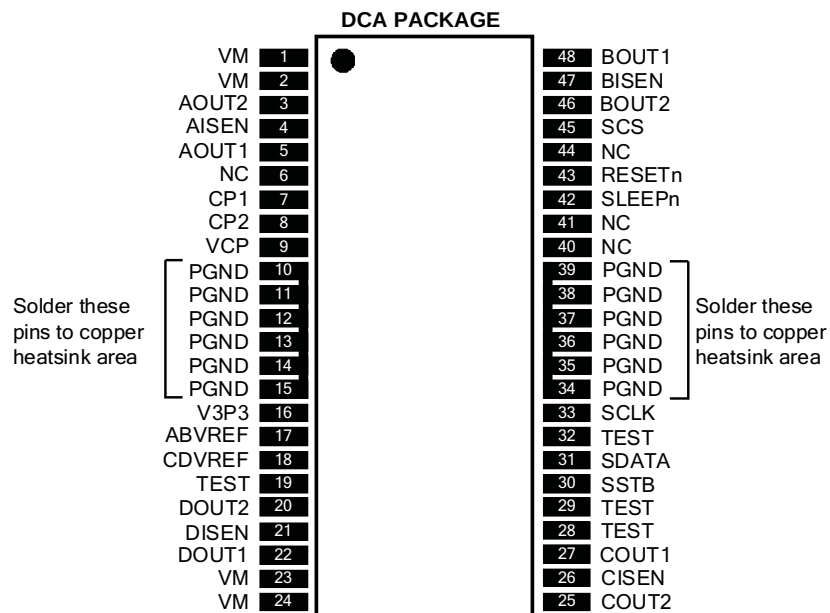


Figure 1. Logic Inputs



ABSOLUTE MAXIMUM RATINGS^{(1) (2)}

over operating free-air temperature range (unless otherwise noted)

V_M	Power supply voltage range	–0.3 to 34	V
V_I	Logic input voltage range ⁽³⁾	–0.5 to 5.75	V
$I_{O(peak)}$	Peak motor drive output current, $t < 1 \mu s$	Internally limited	
I_O	Motor drive output current ⁽⁴⁾	1.5	A
P_D	Continuous total power dissipation	See Dissipation Ratings Table	
T_J	Operating virtual junction temperature range	–40 to 150	°C
T_A	Operating ambient temperature range	–40 to 85	°C
T_{stg}	Storage temperature range	–60 to 150	°C

- (1) Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to network ground terminal.
- (3) Input pins may be driven in this voltage range regardless of presence or absence of V_M .
- (4) Power dissipation and thermal limits must be observed.

DISSIPATION RATINGS

BOARD	PACKAGE	$R_{\theta JA}$	DERATING FACTOR ABOVE $T_A = 25^\circ C$	$T_A < 25^\circ C$	$T_A = 70^\circ C$	$T_A = 85^\circ C$
Low-K ⁽¹⁾	DCA	75.7°C/W	13.2 mW/°C	1.65 W	1.06 W	0.86 W
Low-K ⁽²⁾		32°C/W	31.3 mW/°C	3.91 W	2.50 W	2.03 W
High-K ⁽³⁾		30.3°C/W	33 mW/°C	4.13 W	2.48 W	2.15 W
High-K ⁽⁴⁾		22.3°C/W	44.8 mW/°C	5.61 W	3.59 W	2.91 W

- (1) The JEDEC Low-K board used to derive this data was a 76-mm x 114-mm, 2-layer, 1.6-mm thick PCB with no backside copper.
- (2) The JEDEC Low-K board used to derive this data was a 76-mm x 114-mm, 2-layer, 1.6-mm thick PCB with 25-cm² 2-oz copper on back side.
- (3) The JEDEC High-K board used to derive this data was a 76-mm x 114-mm, 4-layer, 1.6-mm thick PCB with no backside copper and solid 1-oz internal ground plane.
- (4) The JEDEC High-K board used to derive this data was a 76-mm x 114-mm, 4-layer, 1.6-mm thick PCB with 25-cm² 1-oz copper on back side and solid 1-oz internal ground plane.

RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
V_M	Motor power supply voltage range	8		32	V
I_{MOT}	Continuous motor drive output current ⁽¹⁾		1	1.5	A
V_{REF}	VREF input voltage	1		4	V

(1) Power dissipation and thermal limits must be observed.

ELECTRICAL CHARACTERISTICS

over operating free-air temperature range (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER SUPPLIES						
I _{VM}	V _M operating supply current	V _M = 24 V, no loads		5	8	mA
V _{UVLO}	V _M undervoltage lockout voltage	V _M rising		6.5	8	V
V _{CP}	Charge pump voltage	Relative to V _M		12		V
V _{V3P3}	V _{V3P3} output voltage		3.20	3.30	3.40	V
LOGIC-LEVEL INPUTS (INTERNAL PULLDOWNS)						
V _{IL}	Input low voltage				0.7	V
V _{IH}	Input high voltage		2			V
V _{HYS}	Input hysteresis		0.3	0.45	0.6	V
I _{IN}	Input current (internal pulldown current)	V _{IN} = 3.3 V			100	μA
OVERTEMPERATURE PROTECTION						
T _{TSD}	Thermal shutdown temperature	Die temperature	150			°C
MOTOR DRIVERS						
R _{ds(on)}	Motor #1 FET on resistance (each individual FET)	V _M = 24 V, I _O = 0.8 A, T _J = 25°C		0.25		Ω
		V _M = 24 V, I _O = 0.8 A, T _J = 85°C		0.31	0.37	
R _{ds(on)}	Motor #2 FET on resistance (each individual FET)	V _M = 24 V, I _O = 0.8 A, T _J = 25°C		0.30		Ω
		V _M = 24 V, I _O = 0.8 A, T _J = 85°C		0.38	0.45	
I _{OFF}	Off-state leakage current				±12	μA
f _{PWM}	Motor PWM frequency ⁽¹⁾		45	50	55	kHz
t _{BLANK}	ITRIP blanking time ⁽²⁾			3.75		μs
t _F	Output fall time		50		300	ns
t _R	Output rise time		50		300	ns
I _{OCP}	Overcurrent protect level		1.5	3	4.5	A
t _{OCP}	Overcurrent protect trip time		2.5			μs
t _{MD}	Mixed decay percentage	Measured from beginning of PWM cycle		75		%
CURRENT CONTROL						
I _{REF}	xVREF input current	xVREF = 3.3 V	−3		3	μA
ΔI _{CHOP}	Chopping current accuracy	xVREF = 2.5 V, derived from V3P3; 71% - 100% current	−5		5	%
		xVREF = 2.5 V, derived from V3P3; 20% - 56% current	−10		10	

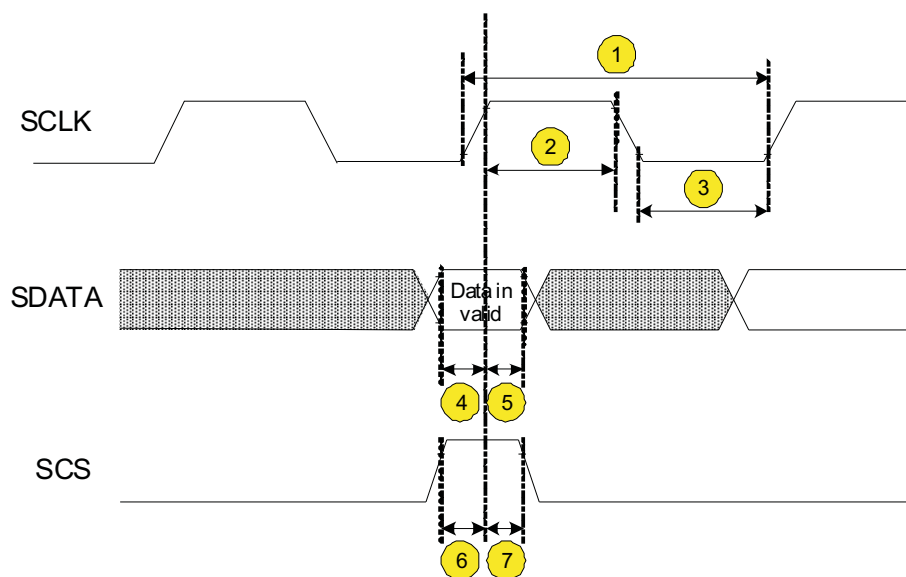
(1) Factory option 100 kHz.

(2) Factory options for 2.5 μs , 5 μs or 6.25 μs .

TIMING REQUIREMENTS

over operating free-air temperature range (unless otherwise noted)

			MIN	MAX	UNIT
1	t_{CYC}	Clock cycle time	62		ns
2	t_{CLKH}	Clock high time	25		ns
3	t_{CLKL}	Clock low time	25		ns
4	$t_{SU(SDATA)}$	Setup time, SDATA to SCLK	5		ns
5	$t_{H(DATA)}$	Hold time, SDATA to SCLK	1		ns
6	$t_{SU(SCS)}$	Setup time, SCS to SCLK	5		ns
7	$t_{H(SCS)}$	Hold time, SCS to SCLK	1		ns



FUNCTIONAL DESCRIPTION

PWM Motor Drivers

The DRV8823 contains four H-bridge motor drivers with current-control PWM circuitry. A block diagram showing drivers A and B of the motor control circuitry (as typically used to drive a bipolar stepper motor) is shown below. Drivers C and D are the same as A and B (though the rds(on) of the output FETs is different).

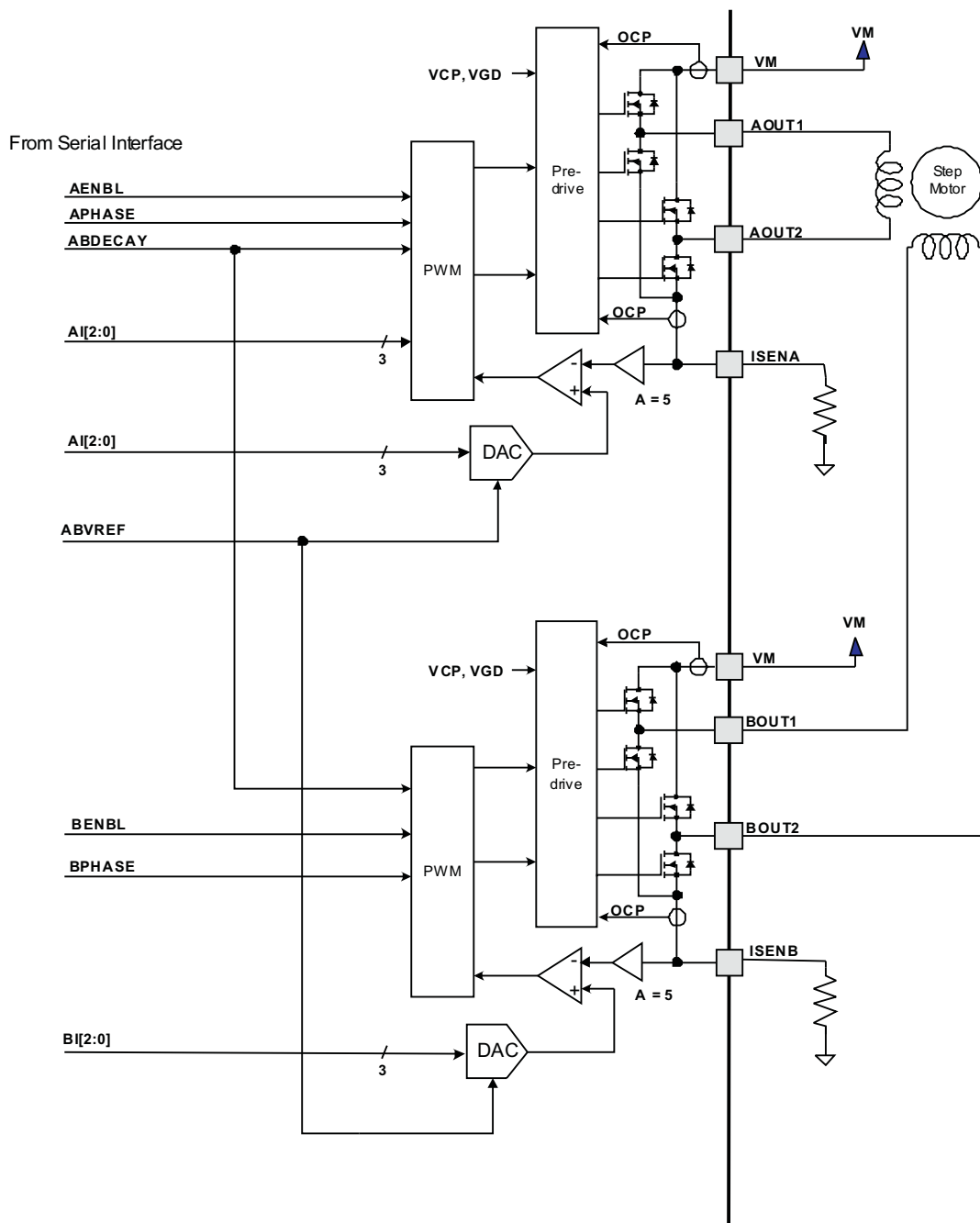


Figure 2. Block Diagram

Note that there are multiple VM motor power supply pins. All VM pins must be connected together to the motor supply voltage.

Bridge Control

The xENBL bits in the serial interface registers enable current flow in each H-bridge when set to "1".

The xPHASE bits in the serial interface registers control the direction of current flow through each H-bridge. The following table shows the logic:

xPHASE	xOUT1	xOUT2
1	H	L
0	L	H

Current Regulation

The motor driver employs fixed-frequency PWM current regulation (also called "current chopping"). When a winding is activated, the current through it rises until it reaches a threshold, then the current is switched off until the next PWM period.

The PWM frequency is fixed at 50 kHz. It may also be set to 100 kHz via factory option.

The PWM chopping current is set by a comparator which compares the voltage across a current sense resistor connected to the xISEN pins, multiplied by a factor of 5, with a reference voltage. The reference voltage is input from the VREF pin.

The full-scale (100%) chopping current is calculated as follows:

$$I_{CHOP} = \frac{V_{REFX}}{5 \cdot R_{ISENSE}} \quad (1)$$

Example:

If a 0.5-Ω sense resistor is used and the VREFx pin is 2.5 V, the full-scale (100%) chopping current is 2.5 V/(5 * 0.5 Ω) = 1 A.

Three serial interface register bits per H-bridge (xI2, xI1 and xI0) are used to scale the current in each bridge as a percentage of the full-scale current set by the VREF input pin and sense resistance. The function of the bits is shown below:

xI2	xI1	xI0	Relative Current (% full-scale chopping current)
0	0	0	20
0	0	1	38
0	1	0	56
0	1	1	71
1	0	0	83
1	0	1	92
1	1	0	98
1	1	1	100

Blanking Time

After the current is enabled in an H-bridge, the voltage on the xISEN pin is ignored for a fixed period of time before enabling the current sense circuitry. This blanking time is fixed at 3.75 μs. Note that the blanking time also sets the minimum on time of the PWM.

Decay Mode

During PWM current chopping, the H-bridge is enabled to drive through the motor winding until the PWM current chopping threshold is reached. This is shown in [Figure 3](#) as case 1. The current flow direction shown indicates positive current flow in the step table below.

Once the chopping current threshold is reached, the H-bridge can operate in two different states, fast decay or slow decay.

In fast decay mode, once the PWM chopping current level has been reached, the H-bridge reverses state to allow winding current to flow in a reverse direction. As the winding current approaches zero, the bridge is disabled to prevent any reverse current flow. Fast-decay mode is shown in [Figure 3](#) as case 2.

In slow-decay mode, winding current is re-circulated by enabling both of the low-side FETs in the bridge. This is shown in [Figure 3](#) as case 3.

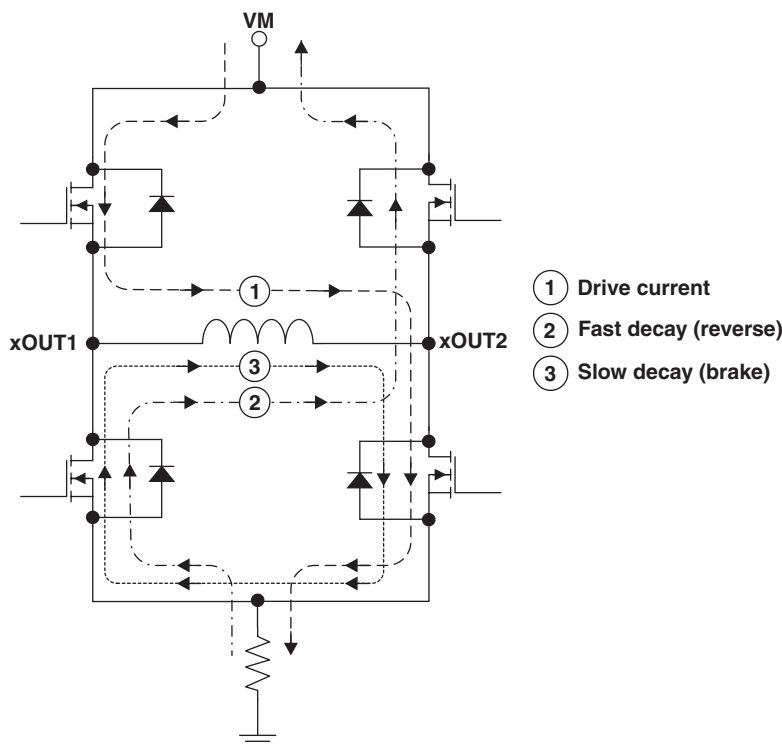


Figure 3. Decay Mode

The DRV8823 supports slow decay and also a mixed decay mode. Mixed decay mode begins as fast decay, but at a fixed period of time (75% of the PWM cycle) switches to slow decay mode for the remainder of the fixed PWM period.

Slow or mixed decay mode is selected by the state of the xDECAY bits in the serial interface registers. If the xDECAY bit is 0, slow decay is selected. If the xDECAY bit is 1, mixed decay is selected.

Protection Circuits

The DRV8823 is fully protected against undervoltage, overcurrent and overtemperature events.

Overcurrent Protection (OCP)

All of the drivers in DRV8823 are protected with an OCP (Over-Current Protection) circuit.

The OCP circuit includes an analog current limit circuit, which acts by removing the gate drive from each output FET if the current through it exceeds a preset level. This circuit will limit the current to a level that is safe to prevent damage to the FET.

A digital circuit monitors the analog current limit circuits. If any analog current limit condition exists for longer than a preset period, all drivers in the device will be disabled.

The device is re-enabled upon the removal and re-application of power at the VM pins.

Thermal Shutdown (TSD)

If the die temperature exceeds safe limits, all drivers in the device will be shut down.

The device will remain disabled until the die temperature has fallen to a safe level. After the temperature has fallen, the device may be re-enabled upon the removal and re-application of power at the VM pin.

Undervoltage Lockout (UVLO)

If at any time the voltage on the VM pins falls below the undervoltage lockout threshold voltage, all circuitry in the device will be disabled. Operation will resume when VM rises above the UVLO threshold. The indexer logic will be reset to its initial condition in the event of an undervoltage lockout.

Shoot-Through Current Prevention

The gate drive to each FET in the H-bridge is controlled to prevent any cross-conduction (shoot through current) during transitions.

Serial Data Transmission

Data transfers consist of sixteen bits of serial data, shifted into the SDATA pin LSB first.

On serial writes to DRV8823, additional clock edges following the final data bit will continue to shift data bits into the data register; therefore, the last 16 bits presented will be latched and used.

One of two registers is selected by setting bits in an address field in the four upper bits in the serial data transferred (ADDR in the tables below). One 16-bit register is used to control motor #1 (bridges A & B), and a second 16-bit register is used to control motor 2 (bridges C & D).

Data can only be transferred into the serial interface if the SCS input pin is active high.

Data is initially clocked in to a temporary holding register. This data is latched into the motor driver on the rising edge of the SSTB pin. If the SSTB pin is tied high at all times, the data will be latched in after all 16 bits have been transferred.

Data Format

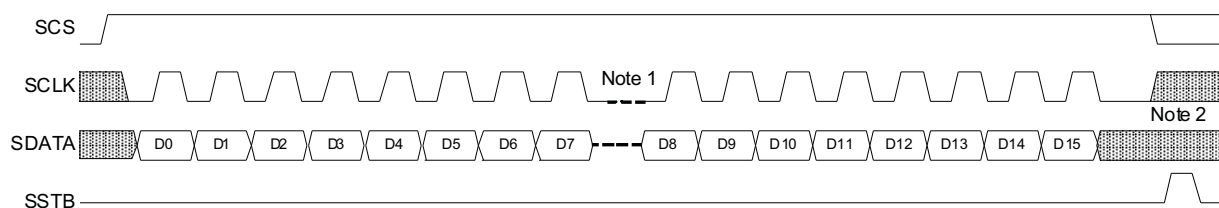
Table 1. Motor 1 Command (Bridges A & B)

Bit	D15 - D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Name	ADDR (= 0000)	BDECAY	B12	B11	B10	BPHASE	BENBL	ADECAY	A12	A11	A10	APHASE	AENBL
Reset Value	x	0	0	0	0	0	0	0	0	0	0	0	0

Table 2. Motor 2 Command (Bridges C & D)

Bit	D15 - D12	D11	D10	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0
Name	ADDR (= 0001)	DDECAY	D12	D11	D10	DPHASE	DENBL	CDECAY	C12	C11	C10	CPHASE	CENBL
Reset Value	x	0	0	0	0	0	0	0	0	0	0	0	0

Serial Data Timing


Figure 4. Serial Data Timing Diagram

Note 1: Any amount of time is allowed between clocks, or groups of clocks, as long as SCS stays active. This allows 8- or 16-bit transfers.

Note 2: If more than 16 clock edges are presented while transferring data (while SCS is still high), data will continue to be shifted into the data register.

THERMAL INFORMATION

Thermal Protection

The DRV8823 has thermal shutdown (TSD) as described above. If the die temperature exceeds approximately 150°C, the device will be disabled until the temperature drops to a safe level.

Any tendency of the device to enter thermal shutdown is an indication of either excessive power dissipation, insufficient heatsinking, or too high an ambient temperature.

Power Dissipation

Power dissipation in the DRV8823 is dominated by the power dissipated in the output FET resistance, or $R_{DS(ON)}$. Average power dissipation when running a stepper motor can be roughly estimated by [Equation 2](#).

$$P_{TOT} = 4 \cdot R_{DS(ON)} \cdot (I_{OUT(RMS)})^2 \quad (2)$$

where P_{TOT} is the total power dissipation, $R_{DS(ON)}$ is the resistance of each FET, and $I_{OUT(RMS)}$ is the RMS output current being applied to each winding. $I_{OUT(RMS)}$ is equal to the approximately 0.7x the full-scale output current setting. The factor of 4 comes from the fact that there are two motor windings, and at any instant two FETs are conducting winding current for each winding (one high-side and one low-side). Remember that the DRV8823 has two stepper motor drivers, so the power dissipation of each must be added together to determine the total device power dissipation.

The maximum amount of power that can be dissipated in the DRV8823 is dependent on ambient temperature and heatsinking. The thermal dissipation ratings table in the datasheet can be used to estimate the temperature rise for typical PCB constructions.

Note that $R_{DS(ON)}$ increases with temperature, so as the device heats, the power dissipation increases. This must be taken into consideration when sizing the heatsink.

Heatsinking

The PowerPAD™ package uses an exposed pad to remove heat from the device. For proper operation, this pad must be thermally connected to copper on the PCB to dissipate heat. On a multi-layer PCB with a ground plane, this can be accomplished by adding a number of vias to connect the thermal pad to the ground plane. On PCBs without internal planes, copper area can be added on either side of the PCB to dissipate heat. If the copper area is on the opposite side of the PCB from the device, thermal vias are used to transfer the heat between top and bottom layers.

For details about how to design the PCB, refer to TI application report [SLMA002](#), "PowerPAD™ Thermally Enhanced Package" and TI application brief [SLMA004](#), "PowerPAD™ Made Easy", available at www.ti.com.

In general, the more copper area that can be provided, the more power can be dissipated. [Figure 5](#) shows thermal resistance vs. copper plane area for both a single-sided PCB with 2-oz copper heatsink area, and a 4-layer PCB with 1-oz copper and a solid ground plane. Both PCBs are 76 mm x 114 mm, and 1.6 mm thick. It can be seen that the heatsink effectiveness increases rapidly to about 20 cm², then levels off somewhat for larger areas.

Six pins on the center of each side of the package are also connected to the device ground. A copper area can be used on the PCB that connects to the PowerPAD™ as well as to all the ground pins on each side of the device. This is especially useful for single-layer PCB designs.

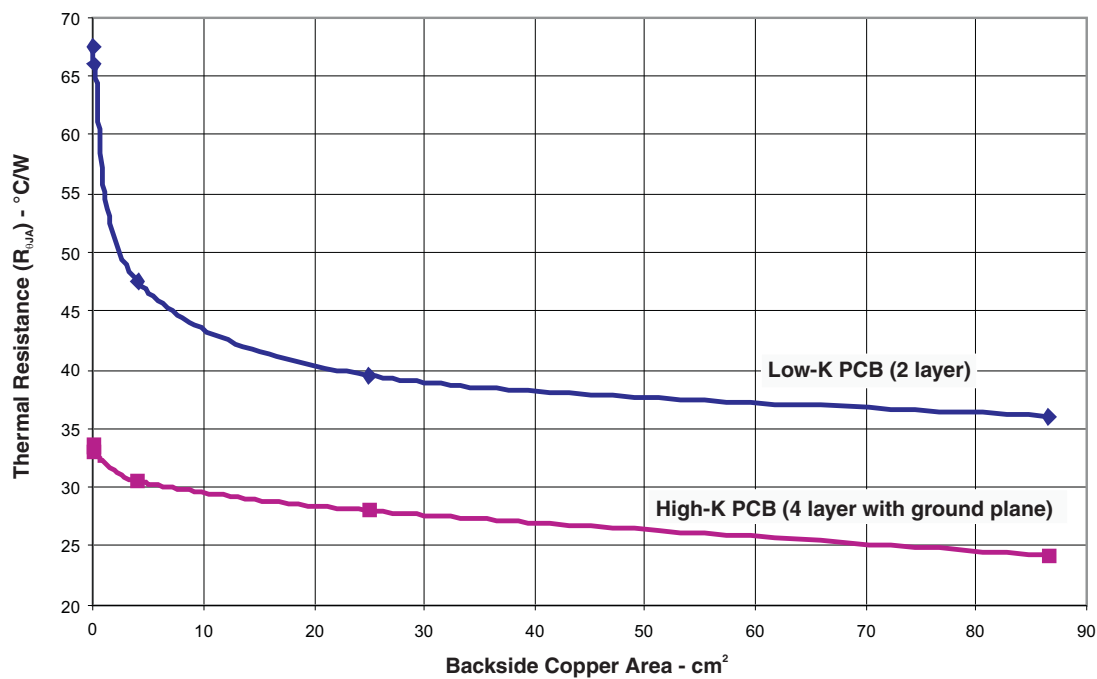


Figure 5. Thermal Resistance vs. Copper Plane Area

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
DRV8823DCA	ACTIVE	HTSSOP	DCA	48	40	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	DRV8823	Samples
DRV8823DCAR	ACTIVE	HTSSOP	DCA	48	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	DRV8823	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBsolete: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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OTHER QUALIFIED VERSIONS OF DRV8823 :

- Automotive: [DRV8823-Q1](#)

NOTE: Qualified Version Definitions:

- Automotive - Q100 devices qualified for high-reliability automotive applications targeting zero defects

TAPE AND REEL INFORMATION
REEL DIMENSIONS

TAPE DIMENSIONS

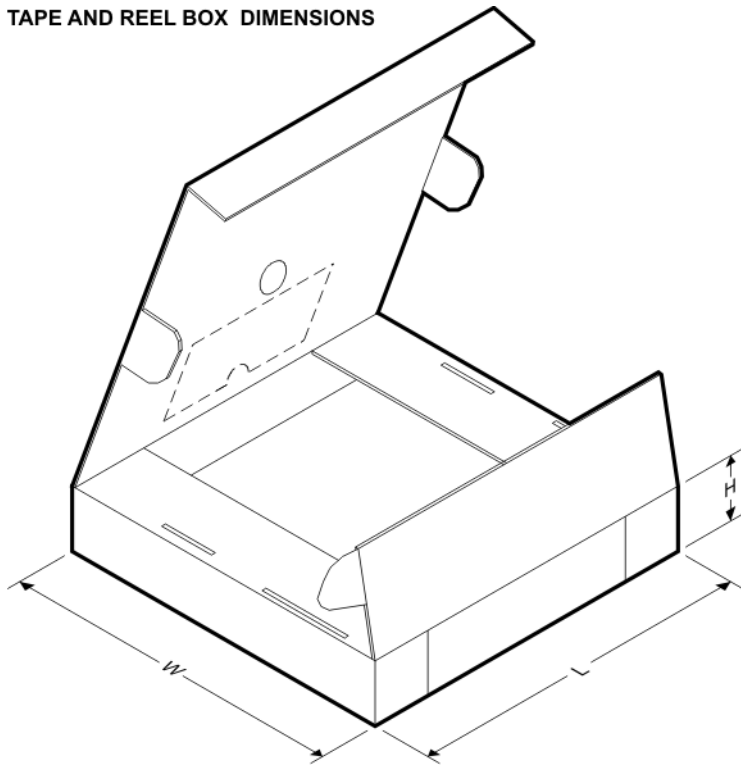

A0	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

TAPE AND REEL INFORMATION

*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
DRV8823DCAR	HTSSOP	DCA	48	2000	330.0	24.4	8.6	15.8	1.8	12.0	24.0	Q1

TAPE AND REEL BOX DIMENSIONS

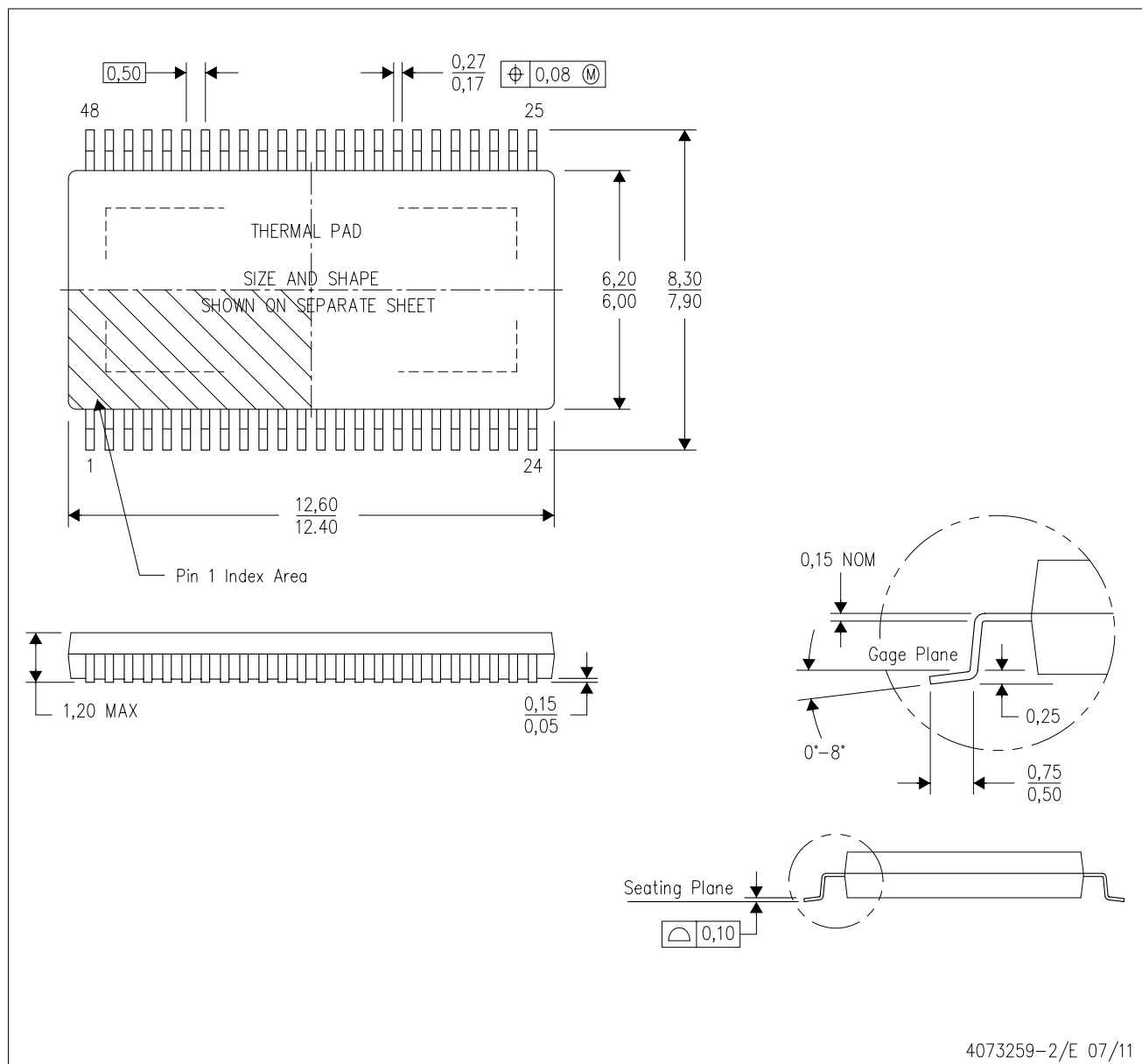


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
DRV8823DCAR	HTSSOP	DCA	48	2000	367.0	367.0	45.0

DCA (R-PDSO-G48)

PowerPAD™ PLASTIC SMALL-OUTLINE



- NOTES:
- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - B. This drawing is subject to change without notice.
 - C. Body dimensions do not include mold flash or protrusion not to exceed 0,15.
 - D. This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 for information regarding recommended board layout. This document is available at www.ti.com <<http://www.ti.com>>.
 - E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
 - F. Falls within JEDEC MO-153

PowerPAD is a trademark of Texas Instruments.

DCA (R-PDSO-G48)

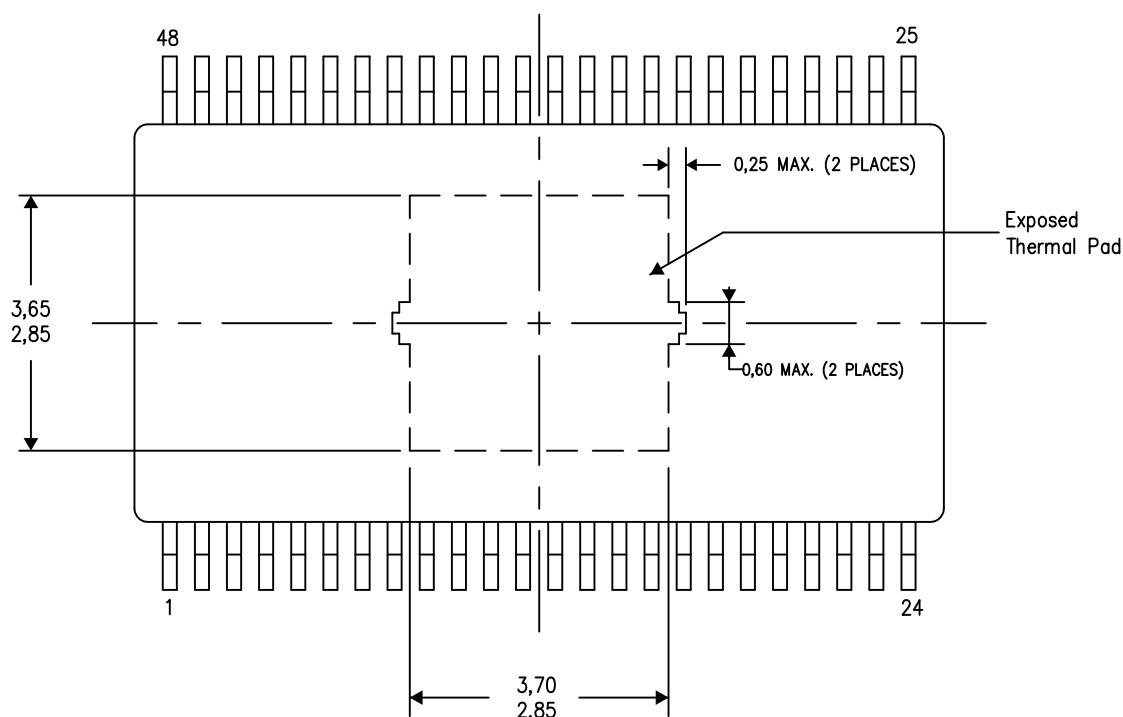
PowerPAD™ PLASTIC SMALL OUTLINE

THERMAL INFORMATION

This PowerPAD™ package incorporates an exposed thermal pad that is designed to be attached to a printed circuit board (PCB). The thermal pad must be soldered directly to the PCB. After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For additional information on the PowerPAD package and how to take advantage of its heat dissipating abilities, refer to Technical Brief, PowerPAD Thermally Enhanced Package, Texas Instruments Literature No. SLMA002 and Application Brief, PowerPAD Made Easy, Texas Instruments Literature No. SLMA004. Both documents are available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



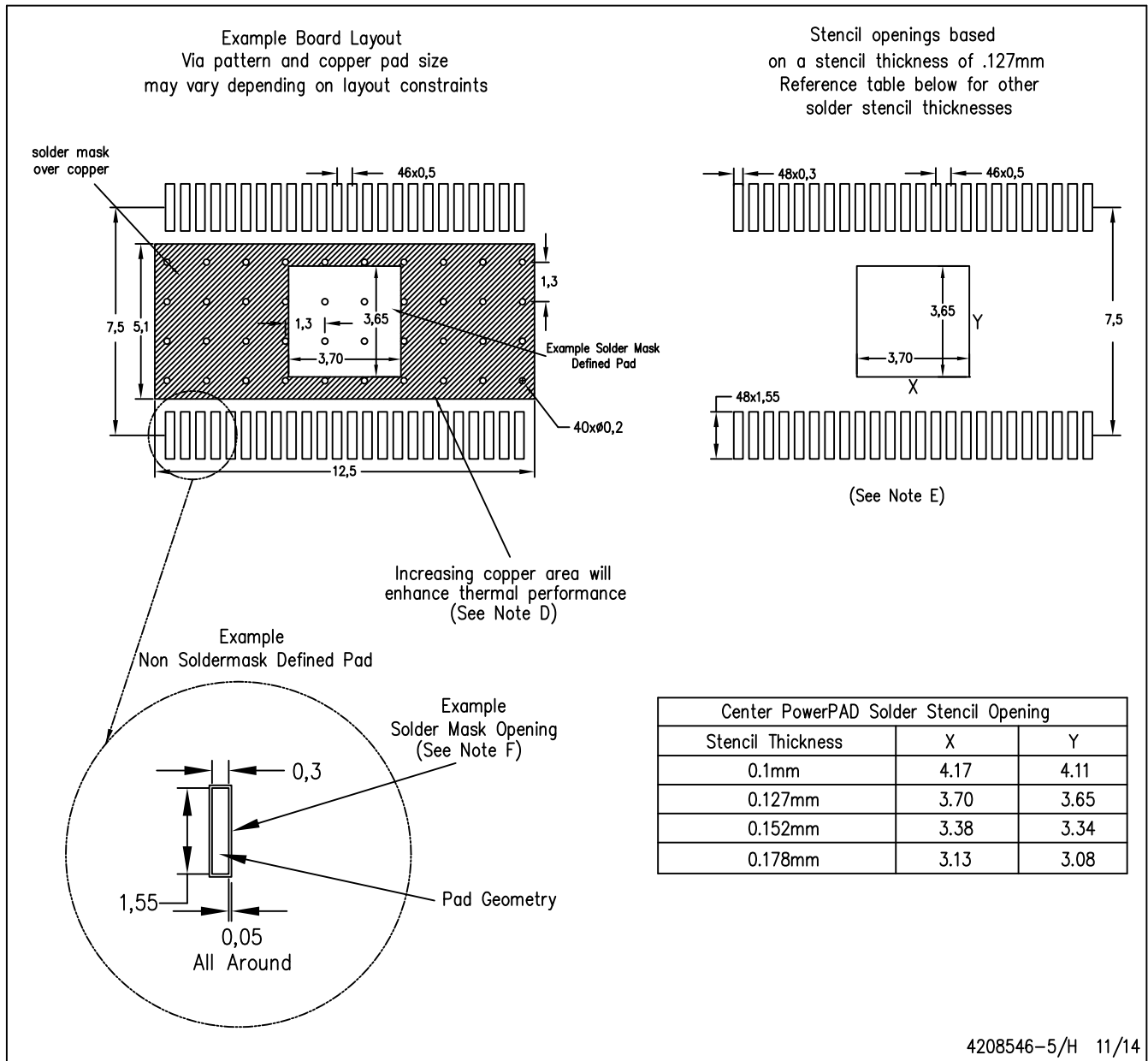
4206320-6/S 11/14

NOTE: A. All linear dimensions are in millimeters

PowerPAD is a trademark of Texas Instruments.

DCA (R-PDSO-G48)

PowerPAD™ PLASTIC SMALL OUTLINE PACKAGE



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Technical Brief, PowerPad Thermally Enhanced Package, Texas Instruments Literature No. SLMA002, SLMA004, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>. Publication IPC-7351 is recommended for alternate designs.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.
 - Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.

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