#### TOSHIBA CMOS Digital Integrated Circuit Silicon Monolithic

#### TC74HC4538AFN

#### Dual Retriggerable Monostable Multivibrator

The TC74HC4538A is a high speed CMOS MONOSTABLE MULTIVIBRATOR fabricated with silicon gate  $C^2MOS$  technology.

It achieves the high speed operation similar to equivalent LSTTL while maintaining the CMOS low power dissipation.

There are two trigger inputs, A input (positive edge input), and  $\overline{B}$  input (negative edge input). These inputs are valid for a slow rise/fall time signal ( $t_r = t_f = 1$  s) as they are schmitt trigger inputs.

After triggering, the output stays in a MONOSTABLE state for the time period determined by the external resistor and capacitor (Rx, Cx). A low level at  $\overline{\text{CD}}$  input breaks this STABLE STATE. In the MONOSTABLE state, if a new trigger is applied, it makes the MONOSTABLE period longer (retrigger mode).

Limitations for CX and RX are as follows:

External capacitor Cx ....... No limitation

External resistor Rx................. $V_{CC} = 2.0 \text{ V}$  more than  $5 \text{ k}\Omega$ 

 $V_{CC} \ge 3.0 \text{ V more than } 1 \text{ k}\Omega$ 

All inputs are equipped with protection circuits against static discharge or transient excess voltage.

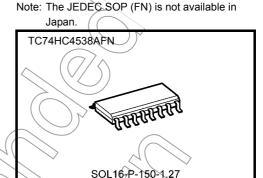
#### Features (Note)

- High speed:  $t_{pd} = 25 \text{ ns (typ.)}$  at  $V_{CC} = 5 \text{ V}$
- Low power dissipation

Stand by state:  $I_{CC} = 4 \mu A \text{ (max)}$  at  $Ta = 25^{\circ}\text{C}$ Active state:  $I_{CC} = 300 \mu A \text{ (max)}$  at  $Ta = 25^{\circ}\text{C}$ 

- High noise immunity: VNIH = VNIL = 28% VCC (min)
- Output drive capability: 10 LSTTL loads
- Symmetrical output impedance: |IOH| = IOL = 4 mA (min)
- Balanced propagation delays: t<sub>pLH</sub> ≃ t<sub>pHL</sub>
- Wide operating voltage range: VCC (opr) = 2 V to 6 V
- Pin and function compatible with 4538B

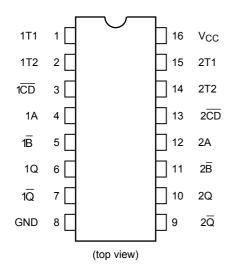
Note: In the case of using only one circuit,  $\overline{CD}$  should be tied to GND, T1 T2  $Q \cdot \overline{Q}$  should be tied to OPEN, the other inputs should be tied to VCC or GND.



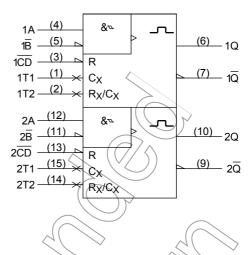
Weight

SOL16-P-150-1.27 (2.13 g (typ.)

## **Pin Assignment**



## **IEC Logic Symbol**



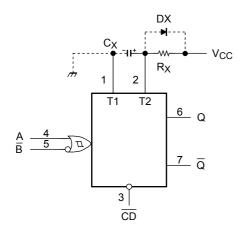
## **Truth Table**

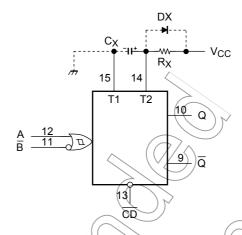
	Inputs			puts	Note
Α	B	CD	Q	ΙQ	Note
	Н	Н	Л		Output Enable
Х	L	Н	L	Н	Inhibit
Н	Х	Н	L	Н	Inhibit
L	$\rightarrow$	Н	Л		Output Enable
Х	Х	L	L	Н	Reset ( )





#### **Block Diagram (Note 1)(Note 2)**





Note 1: C<sub>X</sub>, R<sub>X</sub>, DX are external.

Capacitor, resistor, and diode, respectively.

Note 2: External clamping diode, DX

The external capacitor is charged to  $V_{CC}$  level in the wait state, i.e. when no trigger is applied. Supply voltage is turned off and  $C_X$  is discharged mainly through the internal (parasitic) diode. If  $C_X$  is sufficiently large and  $V_{CC}$  drops rapidly, there will be some possibility of damaging the IC by rush current or latch-up. If the capacitance of the supply voltage filter is large enough and  $V_{CC}$  drops slowly, the rush current is automatically limited and damage to the IC is avoided.

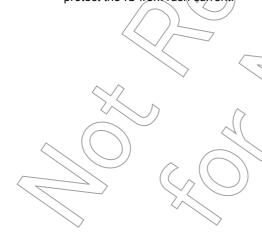
The maximum value of forward current through the parasitic diode is ±20 mA.

In the case of a large  $C_X$ , the limitation of fall time of the supply voltage is determined as follows:

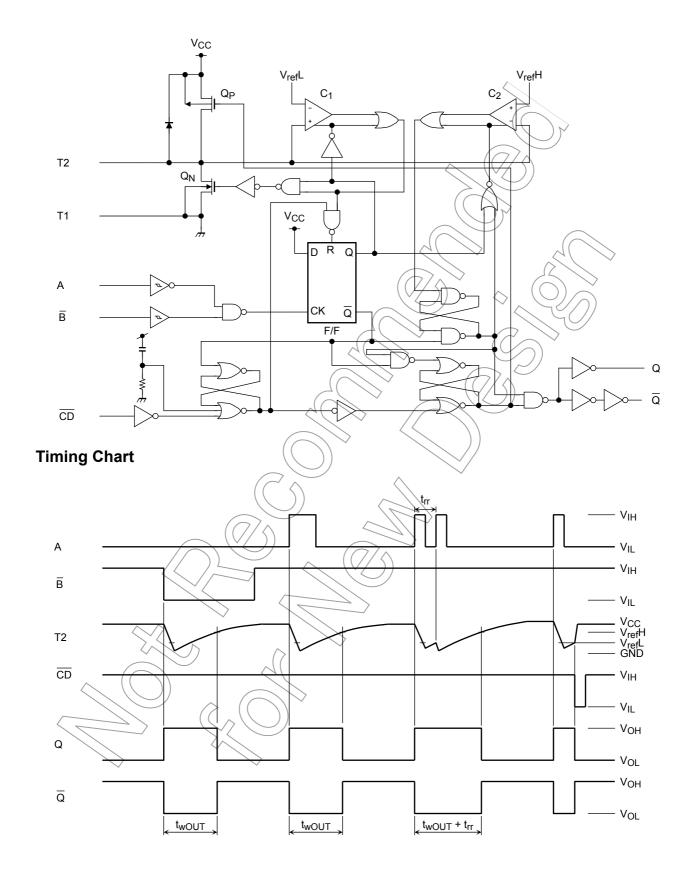
$$t_f \ge (V_{CC} - 0.7) C_X/20 \text{ mA}$$

(If is the time from the voltage supply turning off to the level of supply voltage reaching 0.4 VCC.)

In the care of a system that does not satisfy the above condition, an external clamping diode is needed to protect the IC from rush current.



## **System Diagram**



#### **Functional Description**

(1) Stand-by state

The external capacitor is fully charge to  $V_{\rm CC}$  in the stand-by state. That means, before triggering,  $Q_{\rm P}$  and  $Q_{\rm N}$  transistors which are connected to the T2 node are in the off state. Two comparators that relate to the timing of the output pulse, and two reference voltage supplies stop their operation. The total supply current is only leakage current.

(2) Trigger operation

Trigger operation is effective in either of the following two cases. One is the condition where the A input is low, and the  $\overline{B}$  input has a falling signal. The other, where the  $\overline{B}$  input is high, and the A input has a rising signal.

After trigger becomes effective, comparators  $C_1$  and  $C_2$  start operating, and  $Q_N$  is turned on. The external capacitor discharges through  $Q_N$ . The voltage level at the  $T_2$  node drops. If the  $T_2$  voltage level falls to the internal reference voltage  $V_{ref}L$ , the output of  $C_1$  becomes low. The flip-flop is then reset and  $Q_N$  turns off. At that moment  $C_1$  stops but  $C_2$  continues operating.

After QN turns off, the voltage at T2 start rising at a rate determined by the time constant of external capacitor Cx and resistor Rx.

After the triggering, output Q becomes high, following some delay time of the internal F/F and gates. It stays high even if the voltage of T2 changes from falling to rising. When T2 reaches the internal reference voltage  $V_{ref}H$ , the output of  $C_2$  becomes low, the output Q goes low and  $C_2$  stops its operation. That means, after triggering, when the voltage level of T2 reaches  $V_{ref}H$ , the IC returns to its MONOSTABLE state.

In the case of large value of CX and RX, and ignoring the discharge time of the capacitor and internal delays of the IC, the width of the output pulse, (twoUT), is as follows:

 $t_{wOUT} = 0.70 \cdot C_X \cdot R_X$ 

(3) Retrigger operation

When another new trigger is applied to input A or  $\overline{B}$  while in the MONOSTABLE state, it is effective only if the IC is charging Cx. The voltage level of T2 then falls to  $V_{ref}L$  level again.

Therefore the Q output stays high if the next trigger comes in before the time period set by CX and RX.

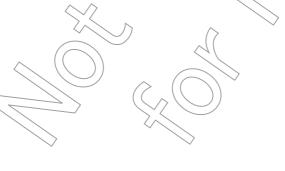
If the 2<sup>nd</sup> trigger is very close to previous trigger, such as application during the discharge cycle, the 2<sup>nd</sup> trigger will not be effective.

The minimum time for effective  $2^{nd}$  trigger,  $t_{rr}$  (min), depends on  $V_{CC}$  and  $C_X$ .

(4) Reset operation

In normal operation,  $\overrightarrow{CD}$  input is held high. If  $\overrightarrow{CD}$  is low, a trigger has no effect because the Q output is held low and the trigger control F/F is reset. Also QP turns on and CX is charged rapidly to VCC.

This means if CD input is set low, the IC goes into a wait state.



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#### **Absolute Maximum Ratings (Note)**

Characteristics	Symbol	Rating	Unit
Supply voltage range	V <sub>CC</sub>	−0.5 to 7	V
DC input voltage	V <sub>IN</sub>	-0.5 to V <sub>CC</sub> + 0.5	V
DC output voltage	V <sub>OUT</sub>	-0.5 to V <sub>CC</sub> + 0.5	V
Input diode current	I <sub>IK</sub>	±20	mA
Output diode current	lok	±20	mA
DC output current	lout	±25	mA
DC V <sub>CC</sub> /ground current	Icc	±50	)) mA
Power dissipation	PD	180	mW
Storage temperature	T <sub>stg</sub>	-65 to 150	°C

Note: Exceeding any of the absolute maximum ratings, even briefly, lead to deterioration in IC performance or even destruction.

Using continuously under heavy loads (e.g. the application of high temperature/current/voltage and the significant change in temperature, etc.) may cause this product to decrease in the reliability significantly even if the operating conditions (i.e. operating temperature/current/voltage, etc.) are within the absolute maximum ratings and the operating ranges.

Please design the appropriate reliability upon reviewing the Toshiba Semiconductor Reliability Handbook ("Handling Precautions"/"Derating Concept and Methods") and individual reliability data (i.e. reliability test report and estimated failure rate, etc).

#### **Operating Ranges (Note 1)**

Characteristics	Symbol	Rating	Unit
Supply voltage	Vec	2 to 6	٧
Input voltage	((VIN))	0 to Vcc	٧
Output voltage	Vout	0 to Vcc	٧
Operating temperature	Topr	=40 to 85	°C
Input rise and fall time ( CD only)	$t_{r}, t_{f}$	0 to 1000 (V <sub>CC</sub> = 2.0 V) 0 to 500 (V <sub>CC</sub> = 4.5 V) 0 to 400 (V <sub>CC</sub> = 6.5 V)	ns
External capacitor	CX	No limitation (Note 2)	F
External resistor	Rx.	≥ 5 k (Note 5) (V <sub>CC</sub> = 2.0 V) ≥ 1 k (Note 5) (V <sub>CC</sub> ≥ 3.0 V)	Ω

Note 1: The operating ranges must be maintained to ensure the normal operation of the device.
Unused inputs must be tied to either VCC or GND.

Note 2: The maximum allowable values of  $C_X$  and  $R_X$  are a function of leakage of capacitor  $C_X$ , the leakage of TC74HC4538A, and leakage due to board layout and surface resistance. Susceptibility to externally induced noise signals may occur for  $R_X > 1$  M $\Omega$ .



## **Electrical Characteristics**

#### **DC Characteristics**

Characteristics	Symbol	Test Condition		Ta = 25°C			Ta = −40 to 85°C		Unit	
Ondracteristics	Cymbol			V <sub>CC</sub> (V)	Min	Тур.	Max	Min	Max	Offic
				2.0	1.50	_	1	1.50	_	
High-level input voltage	$V_{IH}$		_		3.15	_		3.15	_	V
				6.0	4.20	_	1	4.20	_	
Low lovel input				2.0	_	(0)	0.50	_	0.50	
Low-level input voltage	$V_{IL}$		_	4.5			1/.35	_	1.35	V
			<b>I</b>	6.0	-(	7	1.80	_	1.80	
				2.0	1.9	2.0	_	1.9	_	
High-level output	V <sub>ОН</sub>	V <sub>IN</sub> = V <sub>IH</sub> or V <sub>IL</sub>	I <sub>OH</sub> = -20 μA	4.5	4.4	4.5	_	4.4	_	
voltage				6.0	5.9	6.0 - 5.9		$\rightarrow$	V	
(Q, $\overline{Q}$ )			I <sub>OH</sub> = -4 mA	4.5	4.18	4.31	-6	4.13	> -	
			I <sub>OH</sub> = −5.2 mA	6.0	5.68	5.80	1	5.63	) —	
			Ι <sub>ΟL</sub> = 20 μΑ	2.0	_	0.0	0.1		0.1	
Low-level output		VIN		4.5	_	0.0	0.1	<i>&gt;</i> −	0.1	
voltage	$V_{OL}$	= V <sub>IH</sub> or V <sub>IL</sub>		6.0	_	0.0	0.1	_	0.1	V
(Q, $\overline{Q}$ )		VIL.	I <sub>OL</sub> = 4 mA	4.5	- (	(0,17)	0.26	_	0.33	
			I <sub>OL</sub> = 5.2 mA	6.0		0.18	0.26	_	0.33	
Input leakage current	I <sub>IN</sub>	V <sub>IN</sub> = V <sub>C</sub>	C or GND	6.0	_	_	±0.1	-	±1.0	μΑ
T2 terminal input leakage current	I <sub>IN</sub>	V <sub>IN</sub> = V <sub>CC</sub> or GND		6.0	1		±0.5	1	±5.0	μΑ
Quiescent supply current	Icc	VIN = VCC or GND				_	4.0	_	40.0	μA
Active-state supply		VIKI = No.	c or GND	2.0	> _	40	120	_	160	
current	lcc	V <sub>IN</sub> = V <sub>CC</sub> or GND T2-ext = 0.5 V <sub>CC</sub>			_	200	300	_	400	μΑ
(Note)	// ) \_	<u> </u>		6.0	_	300	600	_	800	





## Timing Requirements (input: $t_r = t_f = 6 \text{ ns}$ )

Characteristics	Symbol	Test Condition		Ta = 25°C		Ta = -40 to 85°C	Unit
			V <sub>CC</sub> (V)	Тур.	Limit	Limit	
Minimum pulse width	<b>.</b>		2.0	_	75	95	
(A, $\overline{B}$ )	t <sub>w (L)</sub>	_	4.5 <	_	15	19	ns
(A, B)	t <sub>w (H)</sub>		6.0	Á	13	16	
Minimum clear width			2.0	$\overline{)}$	75	95	
(CD)	t <sub>w (L)</sub>	_	4.5		15	19	ns
(GD)		<	6.0	$\langle \cdot \rangle$	13	16	
	t <sub>rem</sub>		2.0		15	15	
Minimum clear removal time		_	(4.5)	· —	5	5	ns
			6.0	_	5	5	
		R <sub>X</sub> = 1 kΩ	2:0	380	4	$\supset$	
	t <sub>rr</sub>	C <sub>X</sub> = 100 pF	4.5	92	> -	_	ns
Minimum retrigger time		ОХ = 100 Ы	6.0	72		_	
		R <sub>X</sub> = 1 kΩ	2.0	(6.0	( <del>(</del>	) —	
		$C_X = 0.01  \mu F$	4.5	<del></del>	$\supset$	_	μs
		ο <sub>λ</sub> – σ.στ μι	6.0	1.2	_	_	

# AC Characteristics (C<sub>L</sub> = 15 pF, $V_{CC}$ = 5 $V_r$ Ta = 25°C, input: $t_r = (t_f \neq 6 \text{ ns})$

Characteristics	Symbol	Test Condition	Min	Тур.	Max	Unit
Output transition time	t <sub>TLH</sub>	-	_	6	12	ns
Propagation delay time $ (A, \ \overline{B} \ -Q, \ \overline{Q} \ ) $	t <sub>p</sub> LH		_	25	44	ns
Propagation delay time ( $\overline{\text{CD}}$ -Q, $\overline{\overline{\text{Q}}}$ )	t <sub>pHL</sub>		_	21	34	ns

#### AC Characteristics ( $C_L = 50 \text{ pF}$ , input: $t_r = t_f = 6 \text{ ns}$ )

Characteristics	Symbol	Test Condition		Ta = 25°C			Ta = −40 to 85°C		Unit
	,		V <sub>CC</sub> (V)	Min	Тур.	Max	Min	Max	
			2.0	_	30	75	_	95	
Output transition time	t <sub>TLH</sub>	_	4.5	_	8	15	_	19	ns
	t <sub>THL</sub>		6.0	_	7	13	_	16	
Propagation delay time	t <sub>pLH</sub>		2.0	_	120	250	7	315	
(A, $\overline{B}$ -Q, $\overline{Q}$ )	t <sub>pHL</sub>	_	4.5	_	30	50_	<i>9</i> –	63	ns
(A, B-Q, Q)	P		6.0	_	25	/43	_	54	
Propagation delay	$t_{pLH}$		2.0		100 <	195	_	245	
time	t <sub>pHL</sub>	_	4.5	-((	25	39	_	49	ns
$(\overline{CD}-Q,\ \overline{Q})$	-prii		6.0	_\	20)	33	_	42	
	t <sub>w</sub> out	C <sub>X</sub> = 0 F	2.0		540	1200	(	1500	
		$R_X = 5 k\Omega (V_{CC} = 2 V)$	4.5	1	180	250 /	>~	320	ns
		$R_X = 1 k\Omega (V_{CC} = 4.5 V, 6 V)$	6.0	? <u>`</u>	150	200		260	
		C <sub>X</sub> = 0.01 μF R <sub>X</sub> = 10 kΩ	2.0	70	83	<b>(96)</b>	70	96	
Output pulse width			4.5	69	77/	85	> 69	85	μs
			6.0	69	77	85	69	85	
		2 21 5	2.0	0.67	0.75	0.83	0.67	0.83	
		$C_X = 0.1  \mu F$	4.5	0.67	0,73	0.77	0.67	0.77	ms
		R <sub>X</sub> = 10 kΩ	6.0	0.67	0.73	0.77	0.67	0.77	
Output pulse width error between circuits	At				)) <sub>±1</sub>				%
(in same package)	Δt <sub>wOUT</sub>		_	$\sim$	/ ±1			_	70
Input capacitance	C <sub>IN</sub>	(( \ -		_	5	10	_	10	pF
Power dissipation capacitance	C <sub>PD</sub>	77	(Note)	> -	70	_	_	_	pF

Note: C<sub>PD</sub> is defined as the value of the internal equivalent capacitance which is calculated from the operating current consumption without load.

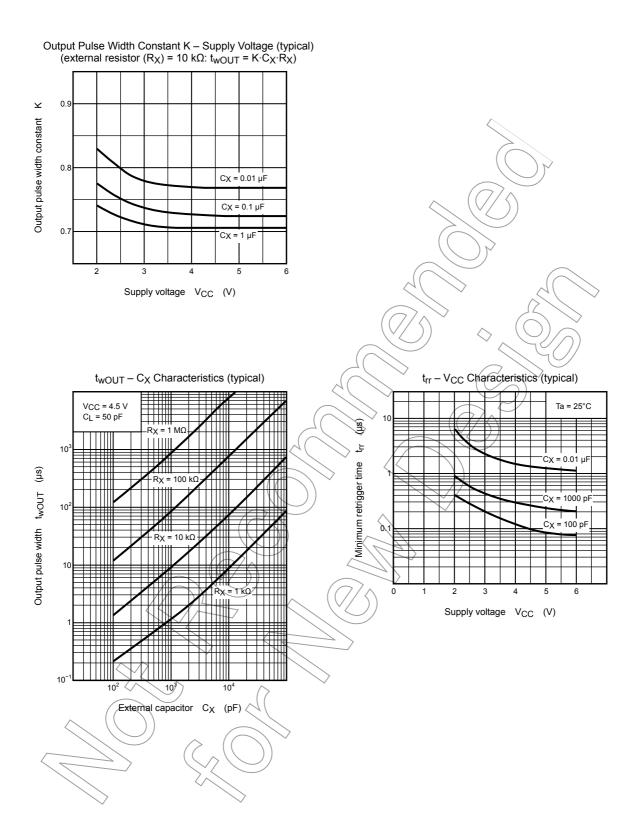
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Average operating current can be obtained by the equation:

 $I_{CC (opr)} = C_{PD} \cdot V_{CC} \cdot f_{IN} + I_{CC} \cdot Duty/100 + I_{CC}/2$  (per circuit)

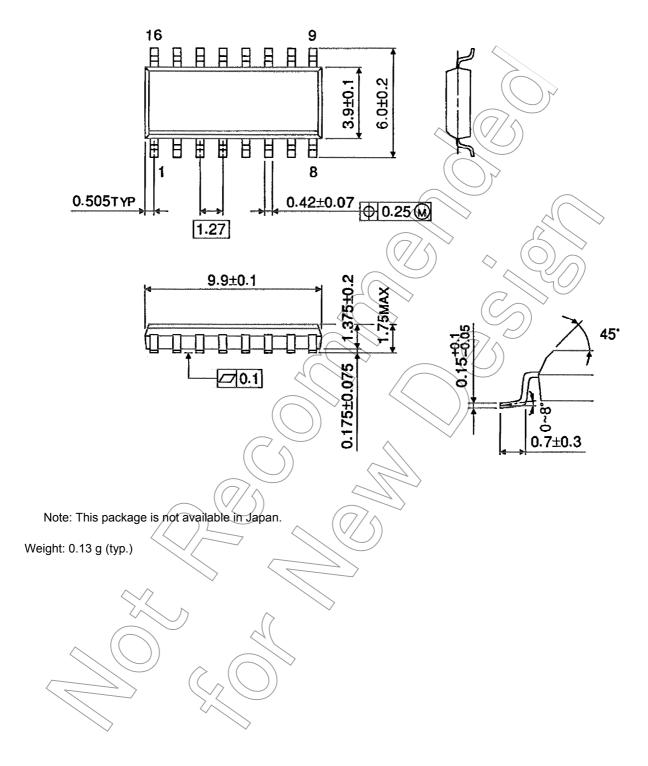
(Icc :: active supply current)

(duty: %)



## **Package Dimensions (Note)**

SOL16-P-150-1.27 Unit: mm



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